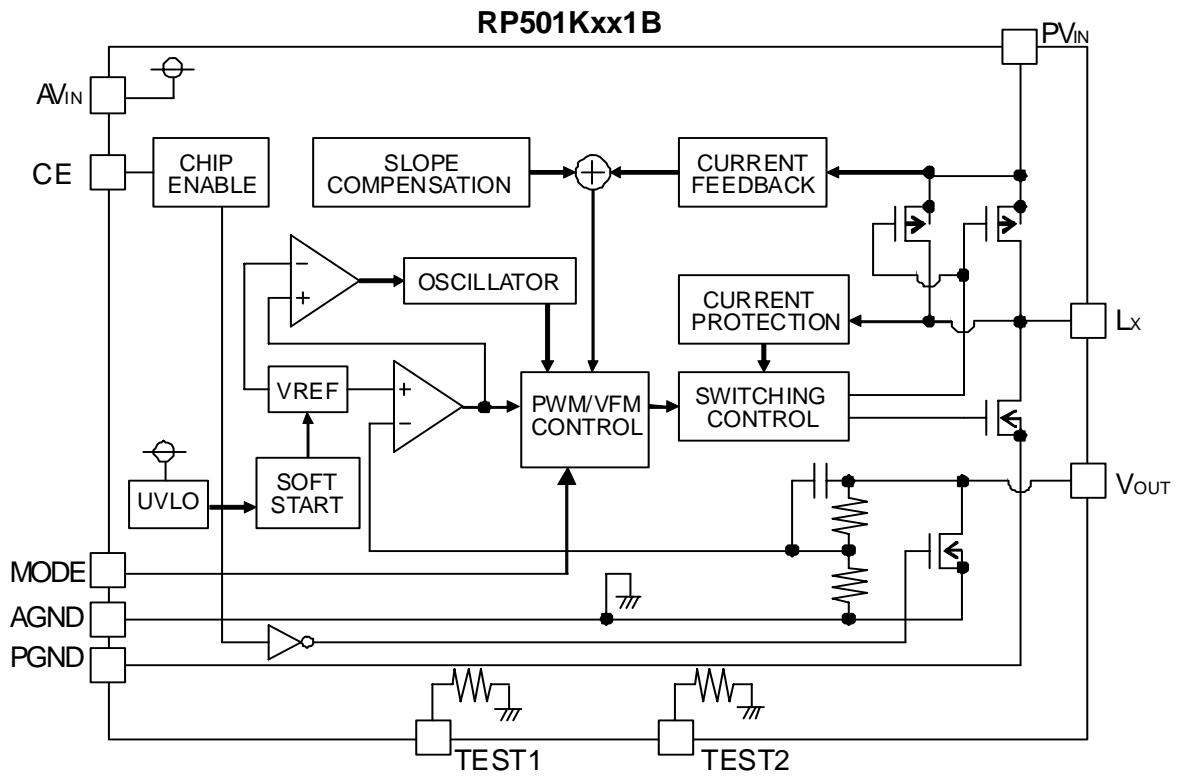
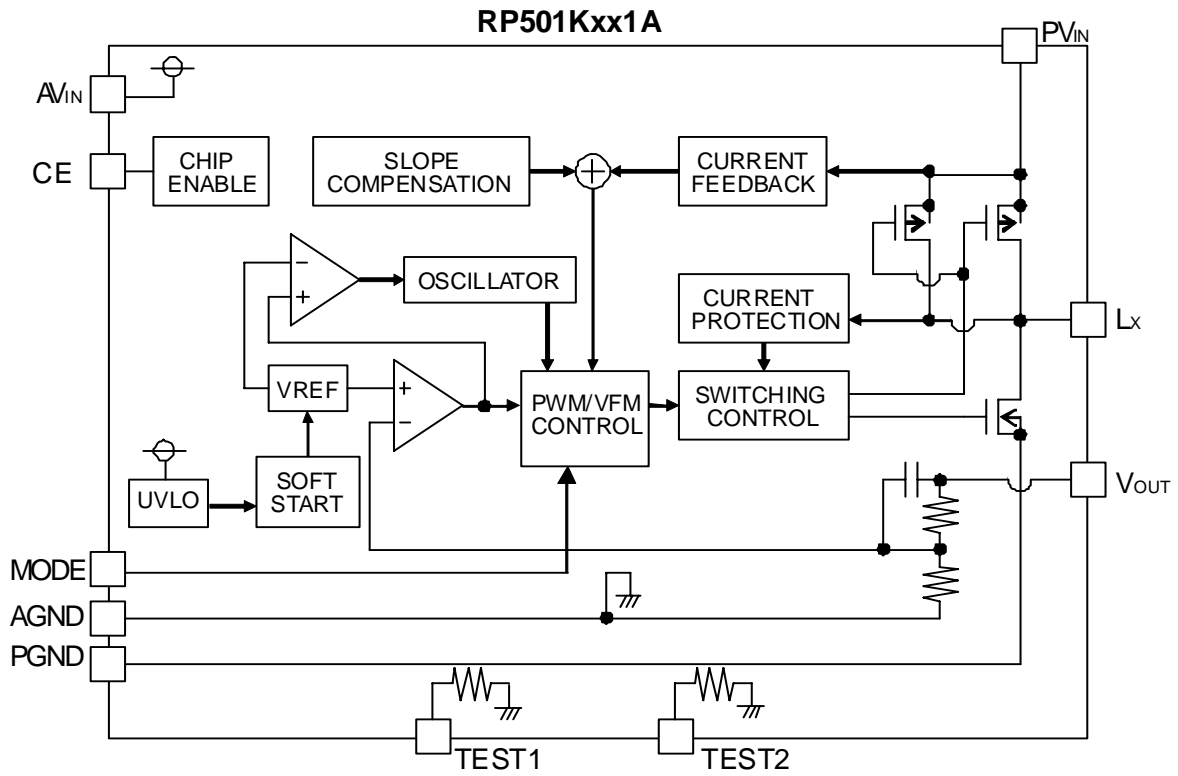




**BLOCK DIAGRAM**



Mode="H": PWM/VFM alternative , Mode="L": PWM fix control  
 Test Pin must be connected to GND.

## Selection Guide

In the RP501K series, output voltage and version can be designated with user's request.  
Part number can be designated as follows:

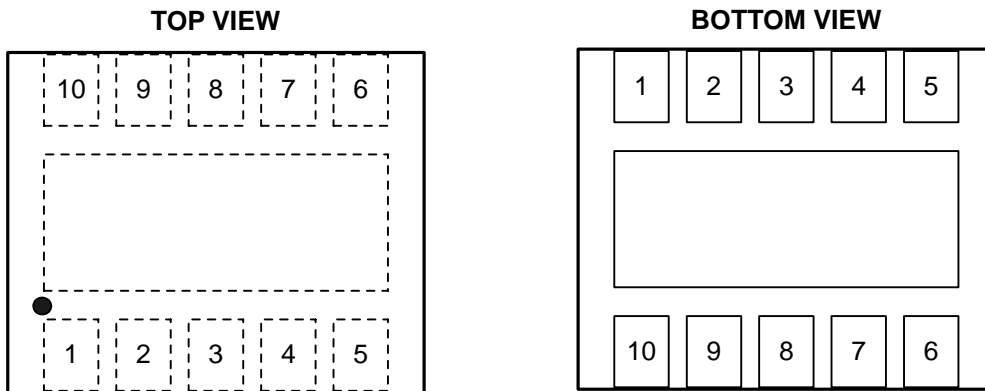
Product Name	Package	Quantity per Reel	Pb Free	Halogen Free
RP501Kxx1*-TR	DFN(PLP)2527-10	5,000 pcs	Yes	Yes
xx : Designation of output voltage( $V_{OUT}$ ): Designation is possible in the range from 1.0V to 3.3V with a step of 0.1V				
* : Designation of Active Type (A) without auto-discharge function (B) with auto-discharge function				

When the mode is into standby with CE signal, auto-discharge transistor turns on, and it makes the turn-off speed faster than normal type.

\* Stepwise setting with 0.05V is possible by custom-made.

## PIN CONFIGURATION

DFN(PLP)2527-10



## PIN DESCRIPTIONS

Pin No.	Symbol	Pin Description
1	CE	Chip Enable Pin ("H" Active)
2	TEST1	Test Pin 1 (Connect this pin to the GND)
3	AGND	Ground Pin
4	PGND	Ground Pin
5	L <sub>x</sub>	Switch Output Pin
6	PV <sub>IN</sub>	Input Pin
7	TEST2	Test Pin 2 (Connect this pin to the GND)
8	MODE	Mode Control Pin
9	AV <sub>IN</sub>	Input Pin
10	V <sub>OUT</sub>	Output Pin

\* Tab is GND level. (They are connected to the reverse side of this IC.) The tab is better to be connected to the GND, but leaving it open is also acceptable.

\* All V<sub>IN</sub> terminals must be connected.

\* All GND terminals must be connected.

## ABSOLUTE MAXIMUM RATINGS

(AGND=PGND=0V)

Symbol	Item	Rating	Unit
$V_{IN}$	$V_{IN}$ Supply Voltage	-0.3 to 6.5	V
$PV_{IN}$	$PV_{IN}$ Supply Voltage	-0.3 to 6.5	V
$V_{LX}$	Lx Pin Voltage	-0.3 to $A/PV_{IN}+0.3$	V
$V_{CE}$	CE Pin Input Voltage	-0.3 to 6.5	V
$V_{MODE}$	MODE Pin Input Voltage	-0.3 to 6.5	V
$V_{OUT}$	$V_{OUT}$ Pin Voltage	-0.3 to 6.5	V
$I_{LX}$	Lx Pin Output Current	1.5	A
$P_D$	Power Dissipation *	910	mW
$T_a$	Operating Temp. Range	-40 to 85	°C
$T_{stg}$	Storage Temp. Range	-55 to 125	°C

\*) For Power Dissipation, please refer to PACKAGE INFORMATION.

### ABSOLUTE MAXIMUM RATINGS

Electronic and mechanical stress momentarily exceeded absolute maximum ratings may cause the permanent damages and may degrade the life time and safety for both device and system using the device in the field. The functional operation at or over these absolute maximum ratings is not assured.

### RECOMMENDED OPERATING CONDITIONS (ELECTRICAL CHARACTERISTICS)

All of electronic equipment should be designed that the mounted semiconductor devices operate within the recommended operating conditions. The semiconductor devices cannot operate normally over the recommended operating conditions, even if when they are used over such conditions by momentary electronic noise or surge. And the semiconductor devices may receive serious damage when they continue to operate over the recommended operating conditions.

## ELECTRICAL CHARACTERISTICS

RP501Kxx1x

Ta=25°C

Symbol	Item	Condition	Min.	Typ.	Max.	Unit
A/PVIN	Operating Input Voltage		2.5		5.5	V
VOUT	Step-down Output Voltage	A/PVIN=VCE=3.6V, VMODE=0V VOUT≥1.6V VOUT<1.6V	-1.5% -0.024		+1.5% 0.024	V
$\Delta V_{OUT}/\Delta T_a$	Step-down Output Voltage Temperature Coefficient	-40°C≤Ta≤85°C		±100		ppm/°C
fosc	Oscillator Frequency	A/PVIN=VCE=3.6V	2.025	2.25	2.475	MHz
IDD1	Supply Current 1	A/PVIN=VCE=5.5V, VOUT=0		450	650	μA
IDD2	Supply Current 2	A/PVIN=VCE=VOUT=5.5V		140	210	μA
Istandby	Standby Current	A/PVIN=5.5V, VCE=0V		0	5	μA
ILxleak	Lx Leakage Current	A/PVIN=5.5V, VLX=A/PVIN/GND, VCE=0V	-5	0	5	μA
RONP	On Resistance of Pch Tr.	A/PVIN=5.0V, ILX=-100mA		0.25	0.35	Ω
RONN	On Resistance of Nch Tr.	A/PVIN=5.0V, ILX=-100mA		0.25	0.35	Ω
Maxduty	Max Duty Ratio		100			%
tstart	Soft-start Time	A/PVIN=VCE=3.6V		0.14	0.2	ms
tprot	Protection Delay Time	A/PVIN=VCE=3.6V	0.5	2.0	6.0	ms
ILxlim	Lx Current Limit	A/PVIN=VCE=3.6V	1.2	1.5		A
VUVLO1	UVLO Detector Voltage	A/PVIN=VCE	2.1	2.2	2.3	V
VUVLO2	UVLO Released Voltage	A/PVIN=VCE	2.2	VUVLO1 +0.1	2.4	V
ICE	CE Input Current	A/PVIN=5.5V, VCE=5.5V/0V	-0.1	0.0	0.1	μA
IVOUT	VOUT Input Current	A/PVIN=5.5V, VCE=0V *1	-0.1	0.0	0.1	μA
RLOW	Nch On Resistance for Auto Discharge	A/PVIN=5.5V, VCE=0V *2		150		Ω
VCEH	CE "H" Input Voltage	A/PVIN=5.5V	1.0			V
VCEL	CE "L" Input Voltage	A/PVIN=3.0V			0.4	V
IMODE	Mode Input Current	A/PVIN= 5.5V, VMODE=5.5V/0V	-0.1	0.0	0.1	μA
VMODEH	Mode "H" Input Voltage	A/PVIN=VCE=5.5V	1.0			V
VMODEL	Mode "L" Input Voltage	A/PVIN=VCE=3.0V			0.4	V

Test circuit is "OPEN LOOP" and A/PVIN=VCE=3.6V, AGND=PGND=0V unless otherwise noted.

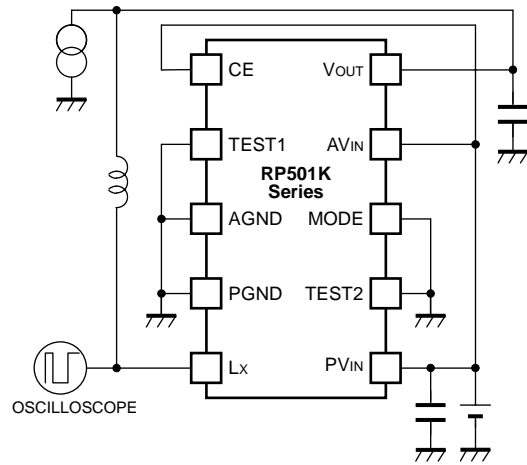
\*1) without auto-discharge type

\*2) with auto-discharge type

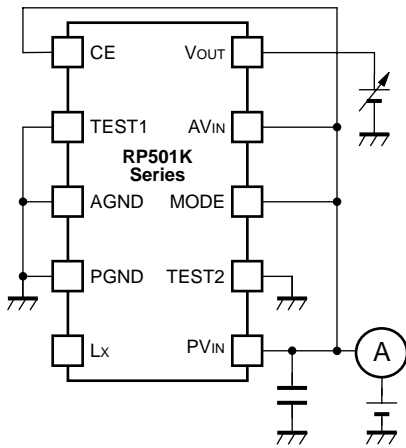
TEST CIRCUITS



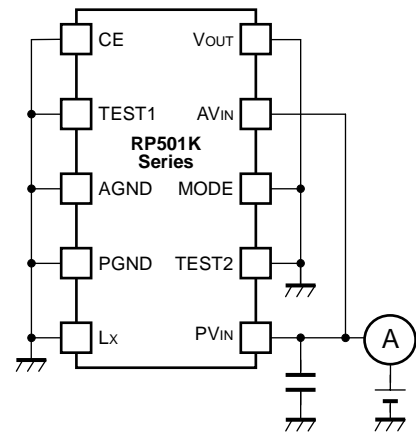
Output Voltage



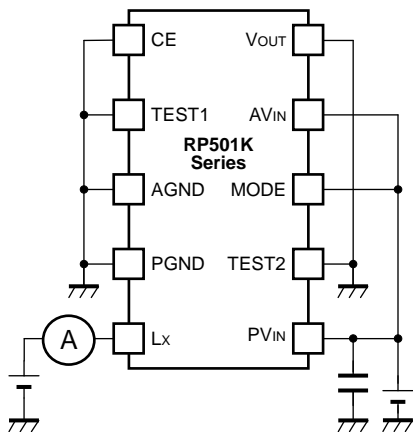
Oscillator Frequency



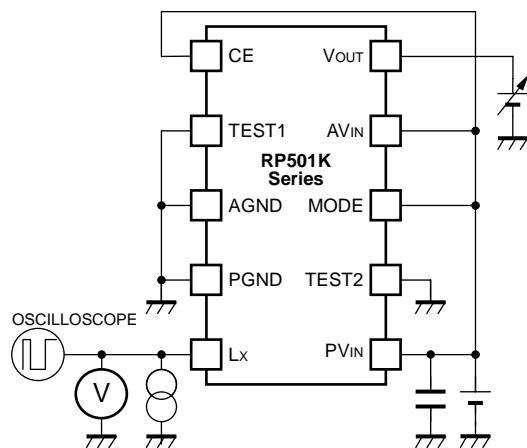
Supply Current 1 / 2



Standby Current



Lx Leakage Current



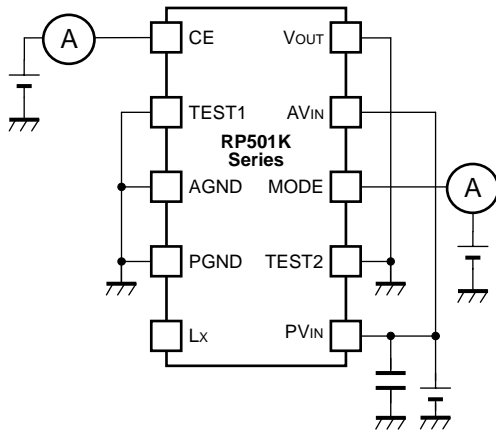
Pch·Nch ON Resistance /  
Protection Delay Time / Lx Current Limit



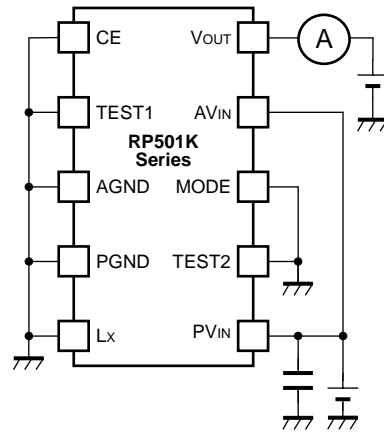
Soft-start Time



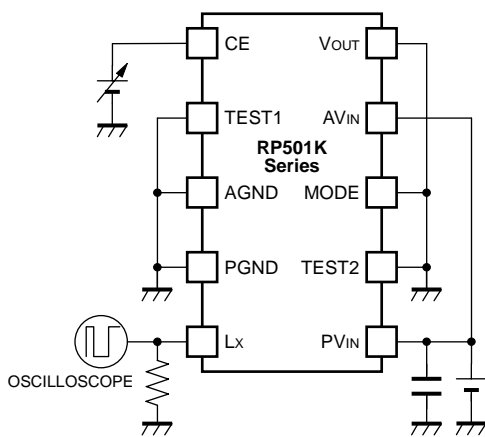
UVLO Detector Voltage / Released Voltage



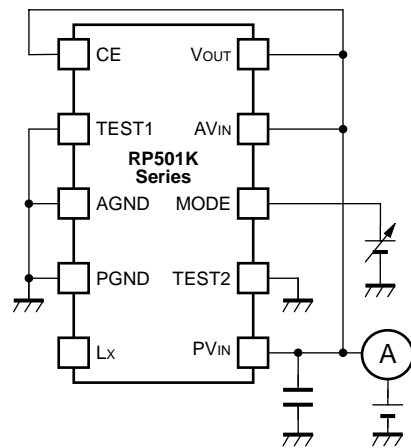
CE-MODE Input Current "H"/"L"



V<sub>out</sub> Input Current /  
Nch On Resistance for Auto Discharge



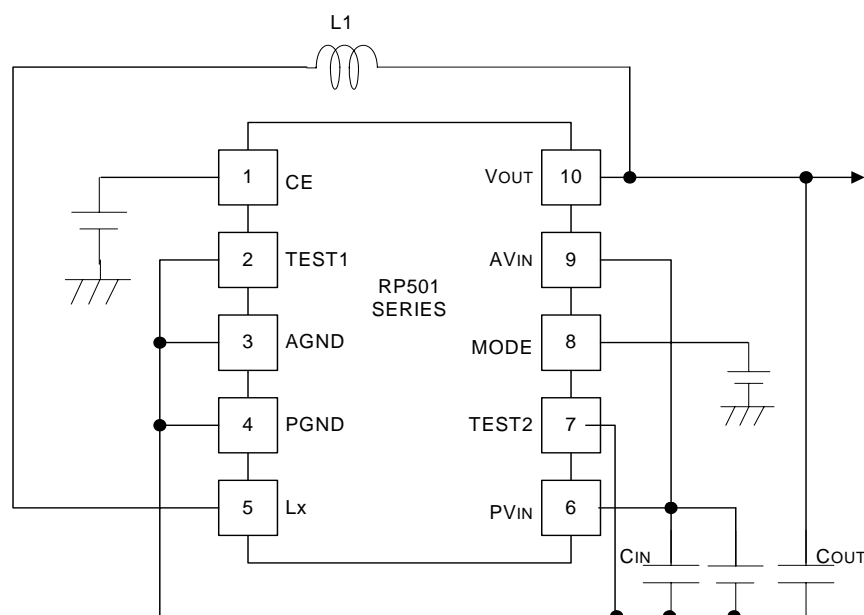
CE Input Voltage



MODE Input Voltage



## TYPICAL APPLICATION



### Parts Recommendation

$C_{IN}$ : 10 $\mu$ F Ceramic (C1608JB0J106K , TDK)

$C_{OUT}$ : 4.7 $\mu$ F Ceramic (C1608JB1A475K , TDK)

L: 2.2 $\mu$ H (NR3010T2R2M , Taiyo Yuden)

## TECHNICAL NOTES

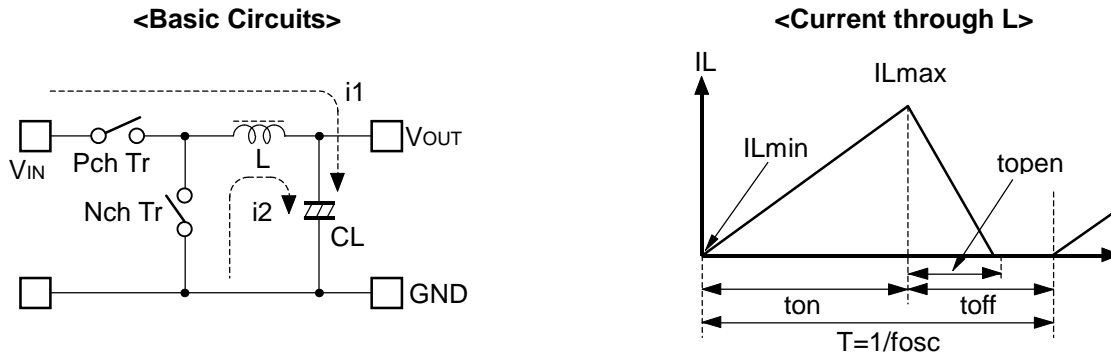
When you use these ICs, consider the following issues:

- Set the same level as AGND and PGND.
- Set external components such as an inductor,  $C_{IN}$ ,  $C_{OUT}$  as close as possible to the IC, in particular, minimize the wiring to  $V_{IN}$  pin and PGND pin. The wiring between  $V_{OUT}$  and load and between L and  $V_{OUT}$  should be separated.
- Use an external capacitor  $C_{IN}$  between  $V_{IN}$  and GND, and  $C_{OUT}$  with a capacity of 4.7 $\mu$ F or more ceramic type.
- Choose an inductor with appropriate inductance range. The phase compensation has been made by these values with output capacitors. The recommendation characteristics of the inductor are low DC resistance, large enough permissible current, and strong against the magnetic saturation. Inductance value may shift depending on an inductor. If the inductance value at an actual load current is low,  $L_x$  peak current may increase and may overlap the  $L_x$  current limit. As a result, over current protection may work.
- Over current protection circuit may be affected by self-heating and heat radiation environment.
- Reinforce the  $V_{IN}$ , PGND, and  $V_{OUT}$  lines sufficiently. Large switching current may flow in these lines. If the impedance of  $V_{IN}$  and PGND lines is too large, the internal voltage level in this IC may shift caused by the switching current, and the operation might be unstable.

The performance of power source circuits using these ICs extremely depends upon the peripheral circuits. Pay attention in the selection of the peripheral circuits. In particular, design the peripheral circuits in a way that the values such as voltage, current, and power of each component, PCB patterns and the IC do not exceed their respected rated values.( such as the voltage, current, and power)

## Operation of step-down DC/DC converter and Output Current

The DC/DC converter charges energy in the inductor when Lx transistor is ON, and discharges the energy from the inductor when Lx transistor is OFF and controls with less energy loss, so that a lower output voltage than the input voltage is obtained. The operation will be explained with reference to the following diagrams:



- Step 1: P-channel Tr. turns on and current IL (=i1) flows, and energy is charged into CL. At this moment, IL increases from ILmin (=0) to reach ILmax in proportion to the on-time period (ton) of P-channel Tr.
- Step 2: When P-channel Tr. turns off, Synchronous rectifier N-channel Tr. turns on in order that L maintains IL at ILmax, and current IL (=i2) flows.
- Step 3: IL (=i2) decreases gradually and reaches IL=ILmin=0 after a time period of topen, and N-channel Tr. Turns off. Provided that in the continuous mode, next cycle starts before IL becomes to 0 because toff time is not enough. In this case, IL value increases from this ILmin (>0).

In the case of PWM control system, the output voltage is maintained by controlling the on-time period (ton), with the oscillator frequency (fosc) being maintained constant.

The maximum value (ILmax) and the minimum value (ILmin) of the current flowing through the inductor are the same as those when P-channel Tr. turns on and off.

The difference between ILmax and ILmin, which is represented by ΔI:

$$\Delta I = I_{Lmax} - I_{Lmin} = V_{OUT} \times t_{open} / L = (V_{IN} - V_{OUT}) \times t_{on} / L \dots\dots\dots \text{Equation 1}$$

wherein,

$$T = 1 / f_{osc} = t_{on} + t_{off}$$

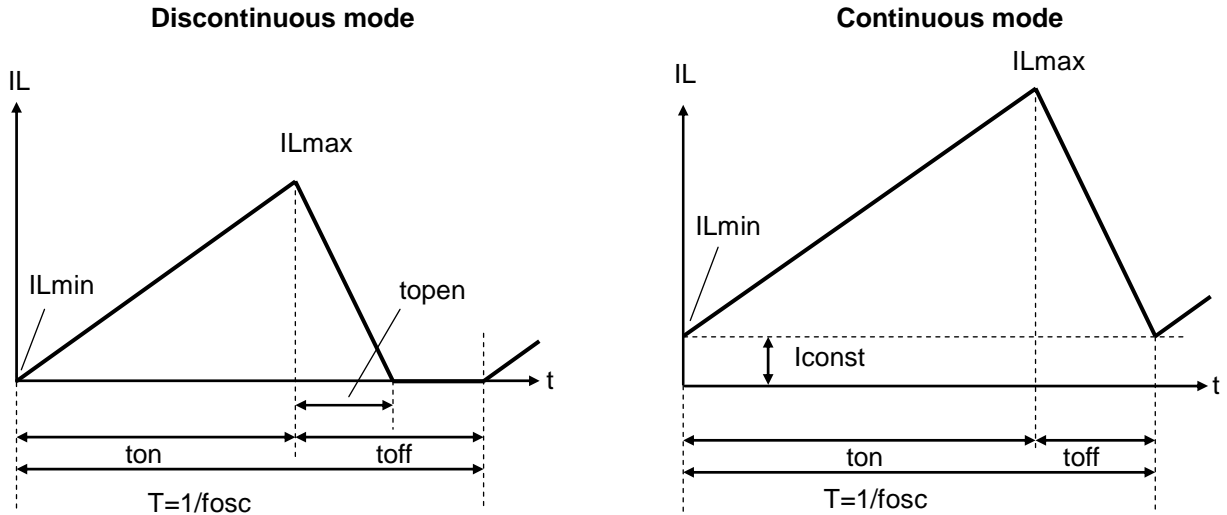
$$\text{duty (\%)} = t_{on} / T \times 100 = t_{on} \times f_{osc} \times 100$$

$$t_{open} \leq t_{off}$$

In Equation 1, VOUT x topen/L and (VIN - VOUT) x ton/L respectively show the change of the current at "ON", and the change of the current at "OFF".

## Discontinuous mode and Continuous mode

When the output current ( $I_{OUT}$ ) is relatively small,  $t_{open} < t_{off}$  as illustrated in the following diagram. In this case, the energy is charged in the inductor during the time period of  $t_{on}$  and is discharged in its entirety during the time period of  $t_{off}$ , therefore  $I_{Lmin}$  becomes to zero ( $I_{Lmin}=0$ ). When  $I_{OUT}$  is gradually increased, eventually,  $t_{open}$  becomes to  $t_{off}$  ( $t_{open}=t_{off}$ ), and when  $I_{OUT}$  is further increased,  $I_{Lmin}$  becomes larger than zero ( $I_{Lmin}>0$ ). The former mode is referred to as the discontinuous mode and the latter mode is referred to as continuous mode.



In the continuous mode, when Equation 1 is solved for  $t_{on}$  and assumed that the solution is  $t_{onc}$ ,

$$t_{onc} = T \times V_{OUT} / V_{IN} \dots\dots\dots \text{Equation 2}$$

When  $t_{on} < t_{onc}$ , the mode is the discontinuous mode, and when  $t_{on} = t_{onc}$ , the mode is the continuous mode.

## Output Current and selection of External components

The relation between the output current and external components is as follows:

When P-channel Tr. of L<sub>x</sub> is ON:

(Wherein, Ripple Current P-P value is described as I<sub>RP</sub>, ON resistance of P-channel Tr. and N-channel Tr. of L<sub>x</sub> are respectively described as R<sub>ONP</sub> and R<sub>ONN</sub>, and the DC resistor of the inductor is described as R<sub>L</sub>.)

$$V_{IN} = V_{OUT} + (R_{ONP} + R_L) \times I_{OUT} + L \times I_{RP} / t_{on} \dots \dots \dots \text{Equation 3}$$

When P-channel Tr. of L<sub>x</sub> is "OFF"(N-channel Tr. is "ON"):

$$L \times I_{RP} / t_{off} = R_{ONN} \times I_{OUT} + V_{OUT} + R_L \times I_{OUT} \dots \dots \dots \text{Equation 4}$$

Put Equation 4 to Equation 3 and solve for ON duty of P-channel transistor,  $t_{on}/(t_{off} + t_{on}) = D_{ON}$ ,

$$D_{ON} = (V_{OUT} + R_{ONN} \times I_{OUT} + R_L \times I_{OUT}) / (V_{IN} + R_{ONN} \times I_{OUT} - R_{ONP} \times I_{OUT}) \dots \dots \dots \text{Equation 5}$$

Ripple Current is as follows:

$$I_{RP} = (V_{IN} - V_{OUT} - R_{ONP} \times I_{OUT} - R_L \times I_{OUT}) \times D_{ON} / f_{osc} / L \dots \dots \dots \text{Equation 6}$$

wherein, peak current that flows through L, and L<sub>x</sub> Tr. is as follows:

$$I_{Lmax} = I_{OUT} + I_{RP} / 2 \dots \dots \dots \text{Equation 7}$$

\*Consider I<sub>Lmax</sub>, condition of input and output and select external components.

\*The above explanation is directed to the calculation in an ideal case in continuous mode.

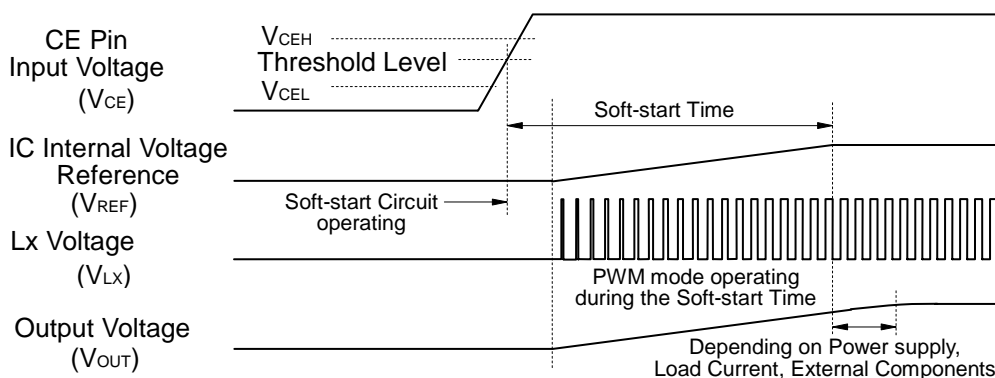
## TIMING CHART

### (1) Soft Start Time

#### In the case of starting this IC with CE

In the case of starting this IC with CE, the operation can be as in the timing chart below.

When the voltage of CE pin ( $V_{CE}$ ) is beyond the threshold level, the operation of the IC starts. The threshold voltage of CE pin is in between CE "H" input voltage ( $V_{CEH}$ ) and CE "L" input voltage ( $V_{CEL}$ ) described in the electrical characteristics table. Soft-start circuit operates, and after the certain time, the reference voltage inside the IC ( $V_{REF}$ ) is rising gradually up to the constant value.



Soft-start time is the time interval from soft start circuit starting point to the reference voltage level reaching point up to this constant level.

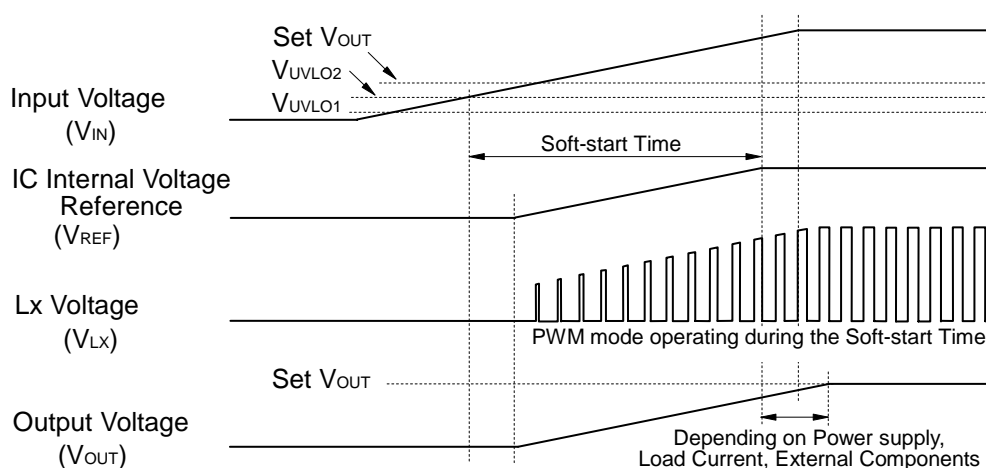
\*Soft start time is not always equal to the turn-on speed of DC/DC converter.

The power supply capacity for this IC, load current, inductance and capacitance values affect the turn-on speed.

#### In the case of starting with power supply

In the case of starting with power supply, when the input voltage ( $V_{IN}$ ) is larger than UVLO released voltage ( $V_{UVLO2}$ ), soft start circuit operates, and after that, the same explanation above is applied to the operation.

Soft-start time is the time interval from soft start circuit starting point to the reference voltage level reaching point up to this constant level.



\*Turn-on speed is affected by next conditions;

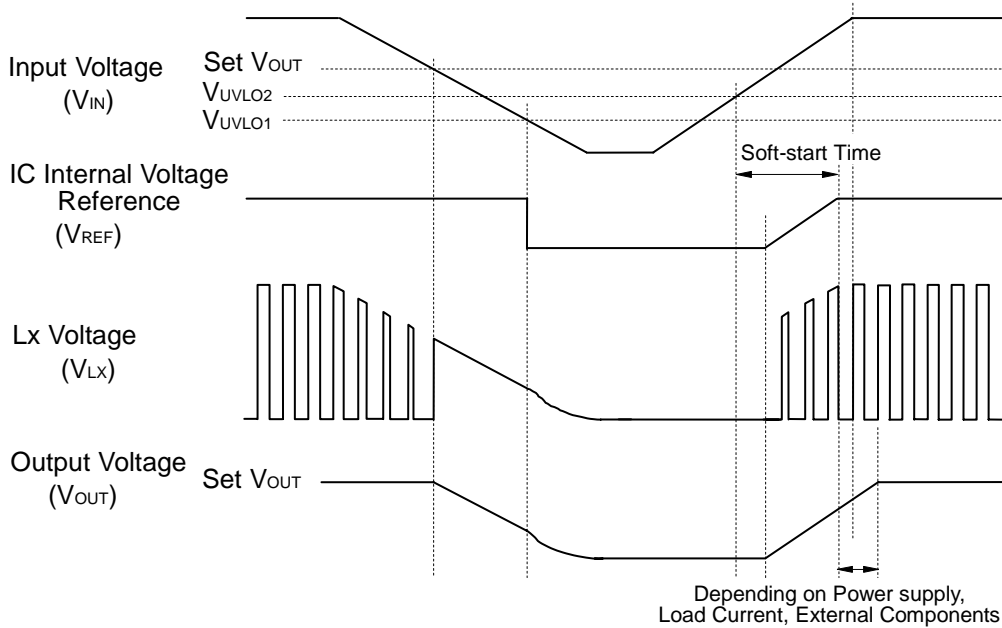
- (a) Input Voltage ( $V_{IN}$ ) rising speed depending on the power supplier to the IC and input capacitor  $C_{IN}$ .
- (b) Output Capacitor  $C_{OUT}$  value and load current value.

**(2) Under Voltage Lockout (UVLO) Circuit**

The step-down DC/DC converter stops and ON duty becomes 100%, if input voltage ( $V_{IN}$ ) becomes less than the set output voltage (Set  $V_{OUT}$ ), the output voltage ( $V_{OUT}$ ) gradually drops according to the input voltage ( $V_{IN}$ ).

If the input voltage drops more and becomes less than UVLO detector threshold ( $V_{UVLO1}$ ), the under voltage lockout circuit (UVLO) operates, the IC internal reference voltage ( $V_{REF}$ ) stops, switching transistors turn off and the output voltage drops according to the load and output capacitor  $C_{OUT}$  value.

To restart the normal operation, the input voltage ( $V_{IN}$ ) must be more than the UVLO released voltage ( $V_{UVLO2}$ ). The timing chart below describes the operation with varying the input voltage ( $V_{IN}$ ).



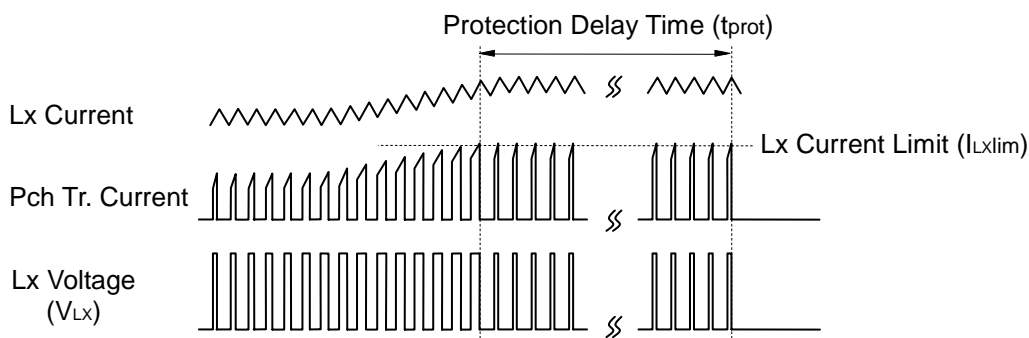
\*Actually, the waveform of  $V_{OUT}$  at UVLO working and releasing varies depending on the initial voltage of  $C_{OUT}$  and load current situation.

### (3) Over Current Protection Circuit, Latch Type Protection Circuit

Over current protection circuit supervises the coil peak current (the current flowing Pch transistor) at each switching cycle, and if the current beyond the Lx current limit ( $I_{Lxlim}$ ), Pch transistor is turned off. The Lx current limit of RP501K is Typ.1.5A.

Further, if the over current status continues equal or longer than protection delay time, or, when the Lx limit current is exceeded even once when the driver operates by duty 100%, a built-in driver is latched in the OFF state and the operation of DC/DC converter stops.

\*Lx current limit and protection delay time is affected by self-heating and ambient environment. If the output is short and the input voltage ( $V_{IN}$ ) is drastically dropped or becomes unstable, the protection operation and delay time may vary.



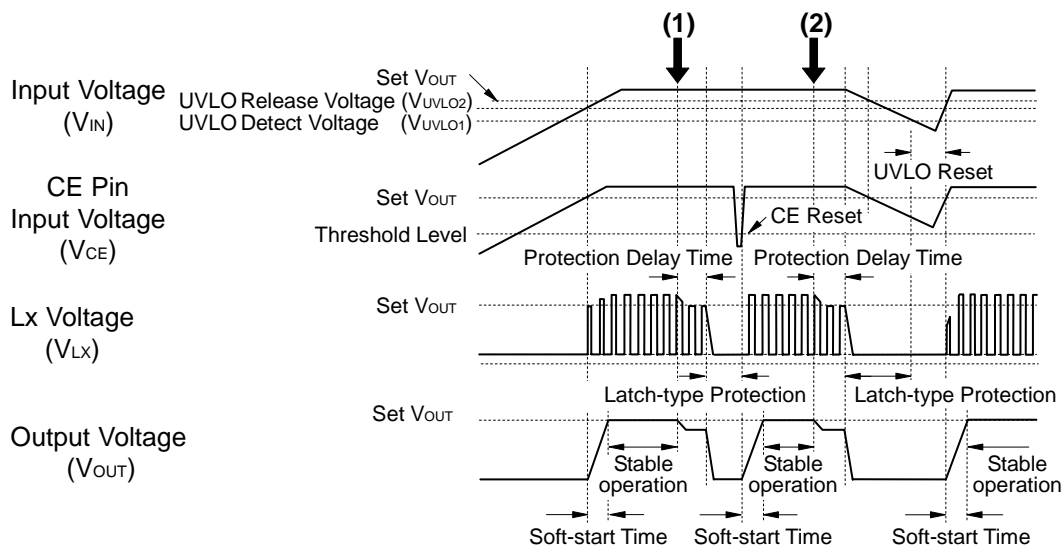
To release the condition of latch type protection, restart this IC by inputting "L" signal to CE pin, or restart this IC with power-on or make the supply voltage lower than UVLO detector threshold ( $V_{UVLO1}$ ) level.

The timing chart shown below describes the changing process of input voltage rising, stable operating, operating with large current, reset with CE pin, stable operating, input voltage falling, input voltage recovering, and stable operating.

If too large current flows through the circuit because of short or other reasons, after the delay time of latch type protection a built-in driver is latched in the OFF state and  $V_{Lx}$  signal will be "L", then output will turn off.

At the point (1), release the latch type protection is realized with CE reset as changed CE signal from "L" to "H".

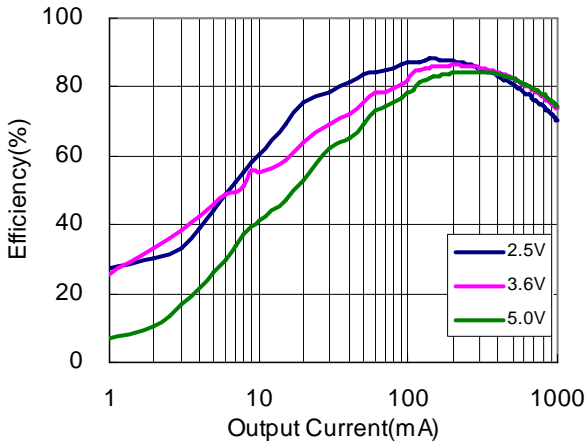
At the point (2), release the latch type protection is realized with UVLO reset as make the supply voltage lower than UVLO detector threshold ( $V_{UVLO1}$ ) level.



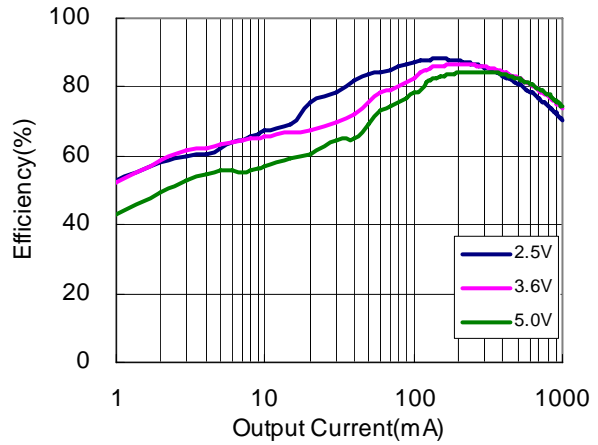
TYPICAL CHARACTERISTICS

1) Efficiency vs. Output Current  
1-1 Input Voltage Dependence

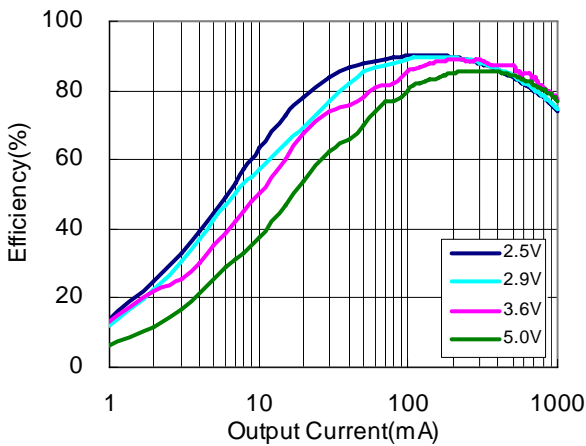
RP501K121x(PWM fixed)  
MODE=0V



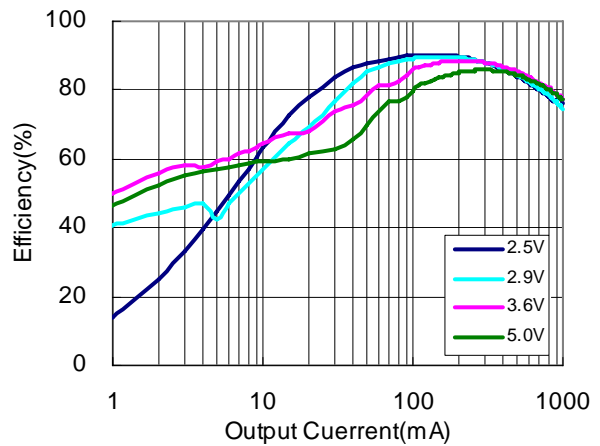
RP501K121x(PWM/VFM mode change)  
MODE=VIN



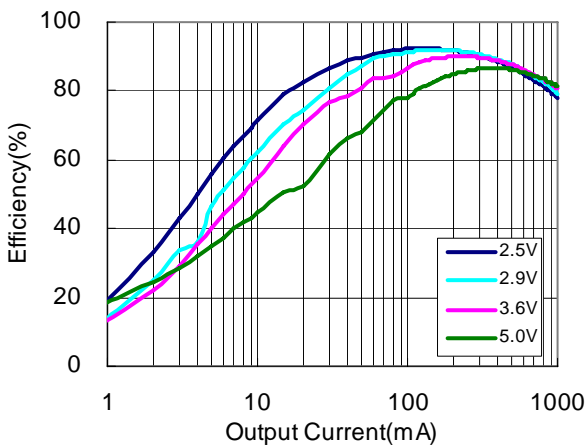
RP501K151x(PWM fixed)  
MODE=0V



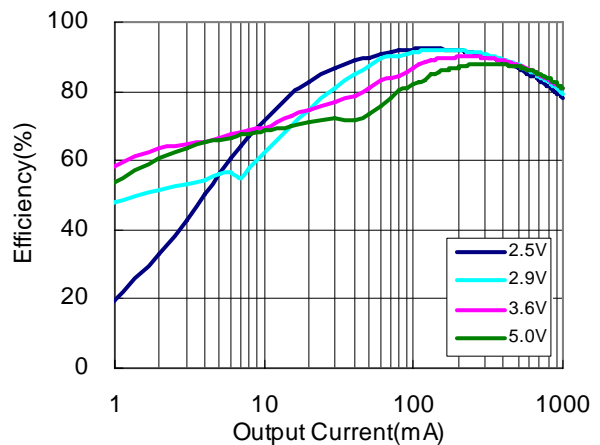
RP501K151x(PWM/VFM mode change)  
MODE=VIN



RP501K181x(PWM fixed)  
MODE=0V

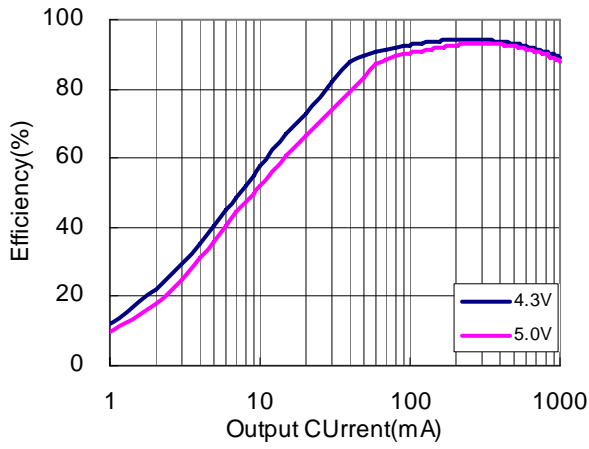


RP501K181x(PWM/VFM mode change)  
MODE=VIN

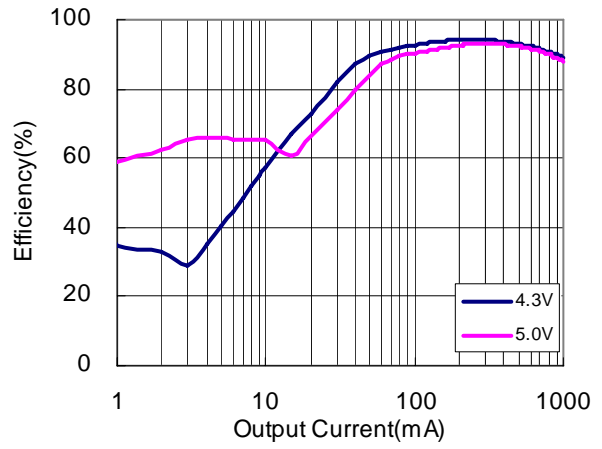




**RP501K331x(PWM fixed)**  
MODE=0V



**RP501K331x(PWM/VFM mode change)**  
MODE=VIN



1-2 Temperature Dependence

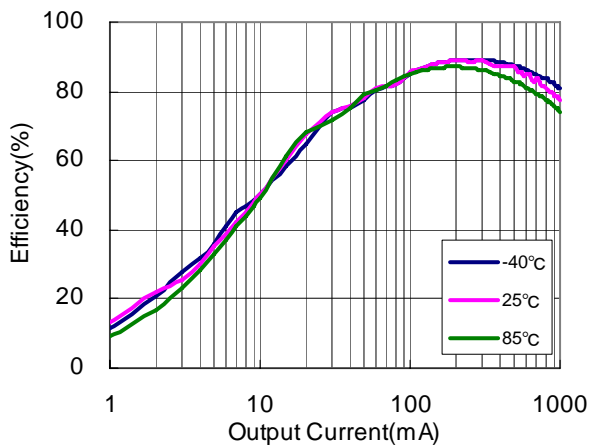
**RP501K121x(PWM fixed)**  
VIN=3.6V, MODE=0V



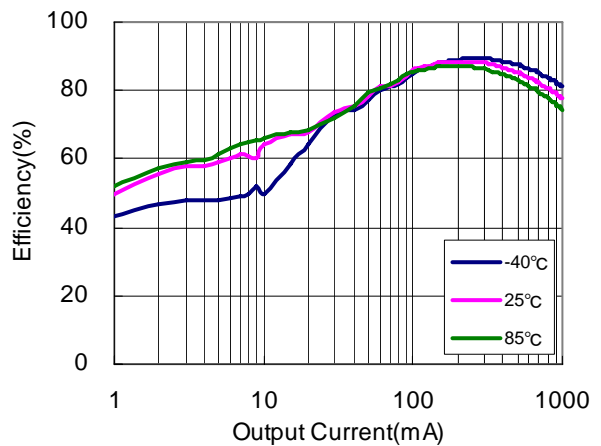
**RP501K121x(PWM/VFM mode change)**  
VIN=MODE=3.6V



**RP501K151x(PWM fixed)**  
VIN=3.6V, MODE=0V

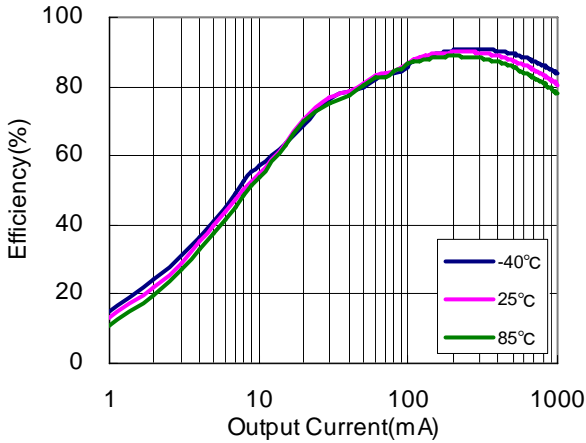


**RP501K151x(PWM/VFM mode change)**  
VIN=MODE=3.6V



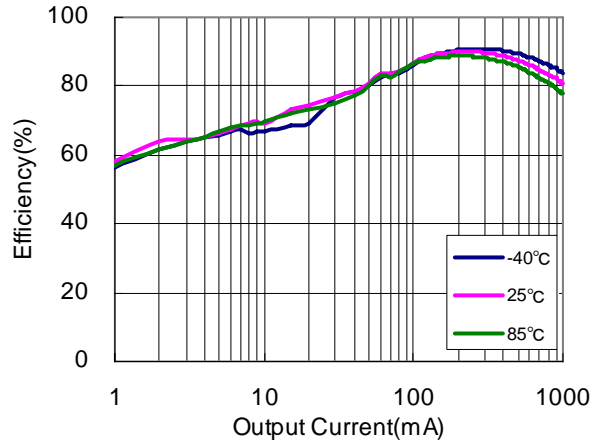
RP501K181x(PWM fixed)

V<sub>IN</sub>=3.6V,MODE=0V



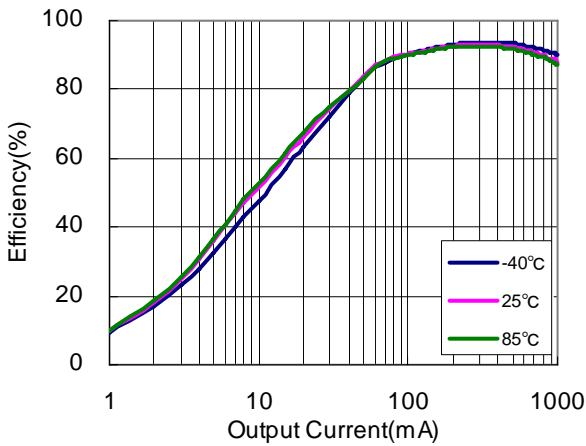
RP501K181x(PWM/VFM mode change)

V<sub>IN</sub>=MODE=3.6V



RP501K331x(PWM fixed)

V<sub>IN</sub>=5.0V,MODE=0V



RP501K331x(PWM/VFM mode change)

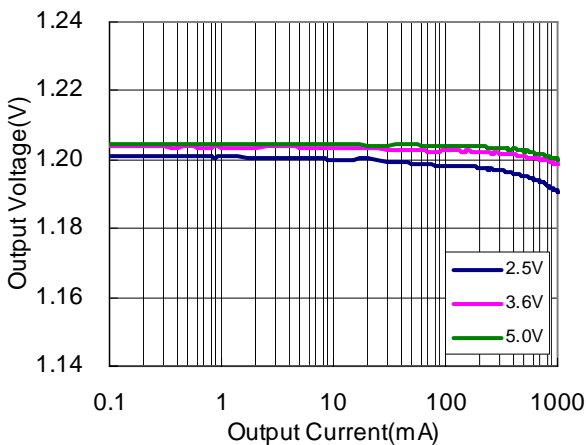
V<sub>IN</sub>=MODE=5.0V



2) Output Voltage vs. Output Current  
2-1 Input Voltage Dependence

RP501K121x(PWM fixed)

MODE=0V

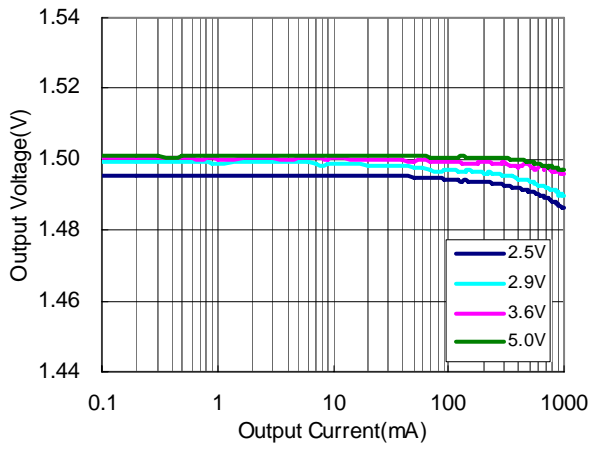


RP501K121x(PWM/VFM mode change)

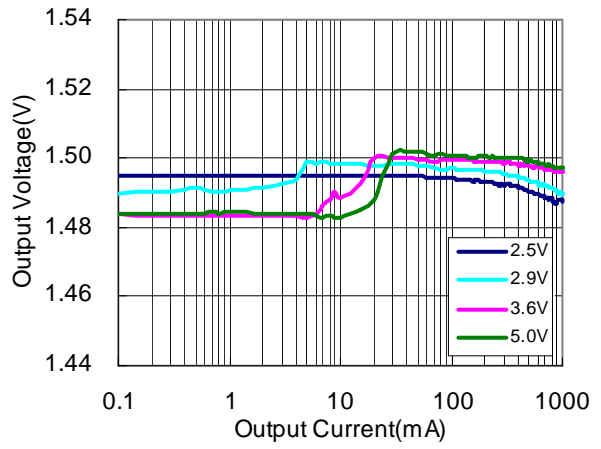
MODE=V<sub>IN</sub>



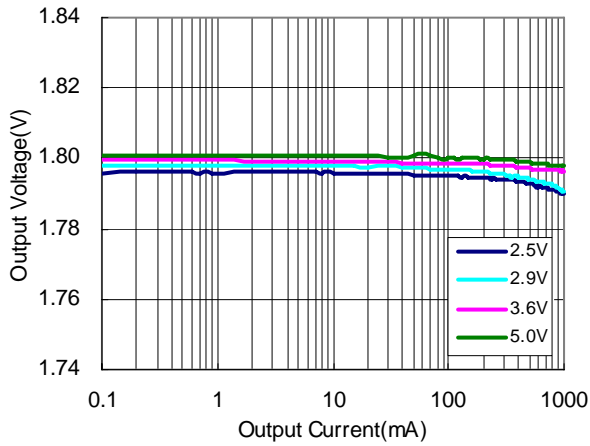
**RP501K151x(PWM fixed)**  
MODE=0V



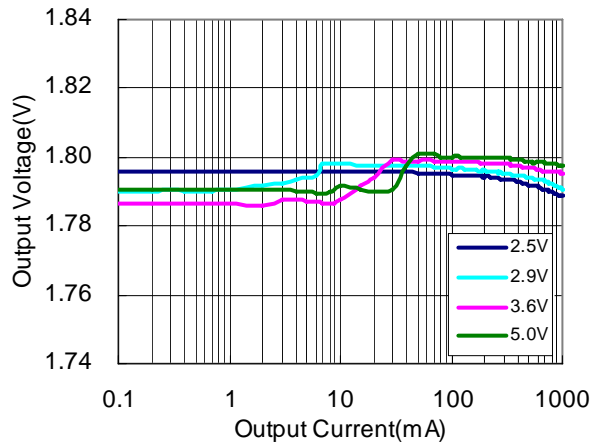
**RP501K151x(PWM/VFM mode change)**  
MODE=V<sub>IN</sub>



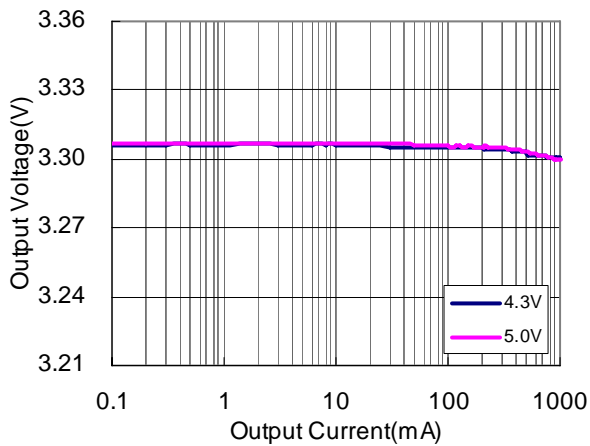
**RP501K181x(PWM fixed)**  
MODE=0V



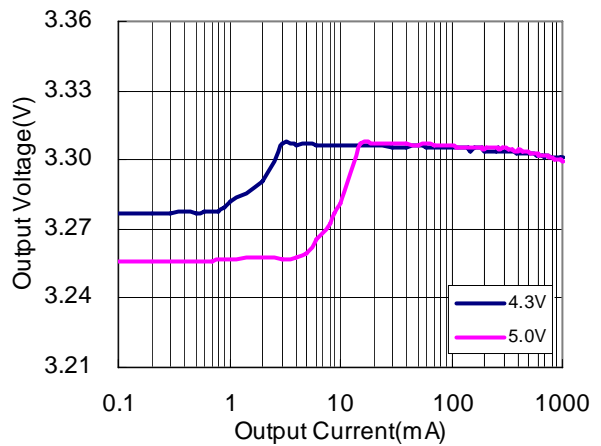
**RP501K181x(PWM/VFM mode change)**  
MODE=V<sub>IN</sub>



**RP501K331x(PWM fixed)**  
MODE=0V



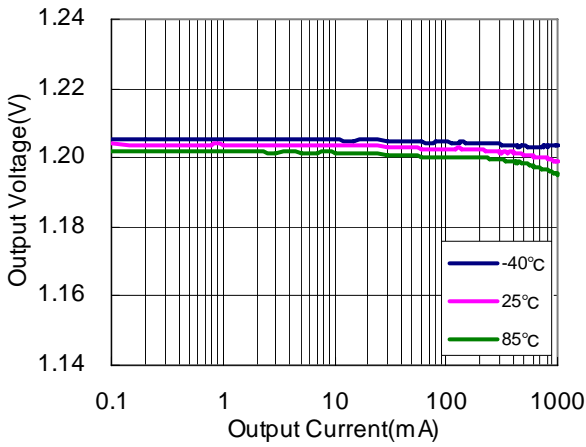
**RP501K331x(PWM/VFM mode change)**  
MODE=V<sub>IN</sub>



2-2 Temperature Dependence

RP501K121x(PWM fixed)

V<sub>IN</sub>=3.6V,MODE=0V



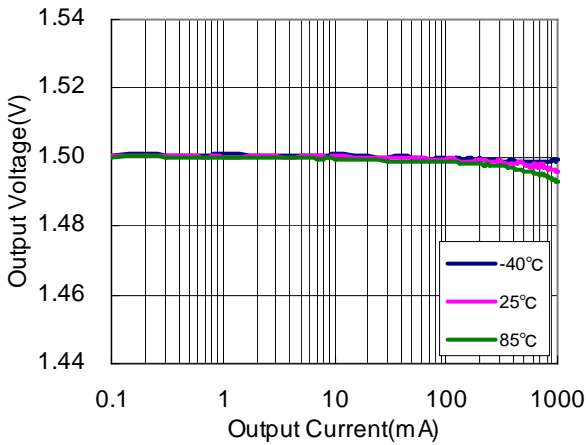
RP501K121x(PWM/VFM mode change)

V<sub>IN</sub>=MODE=3.6V



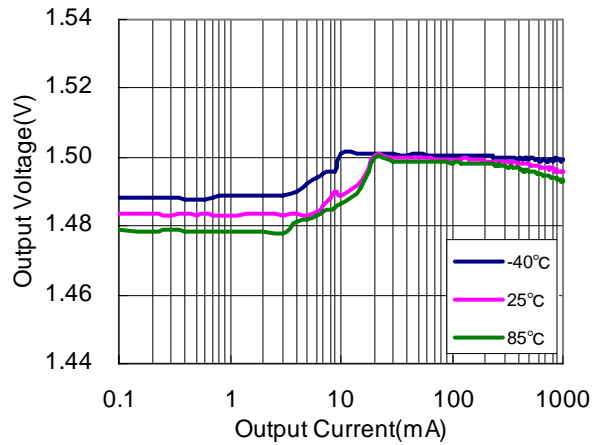
RP501K151x(PWM fixed)

V<sub>IN</sub>=3.6V,MODE=0V



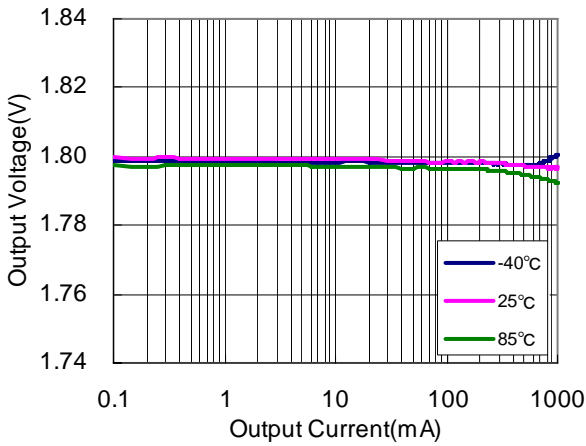
RP501K151x(PWM/VFM mode change)

V<sub>IN</sub>=MODE=3.6V



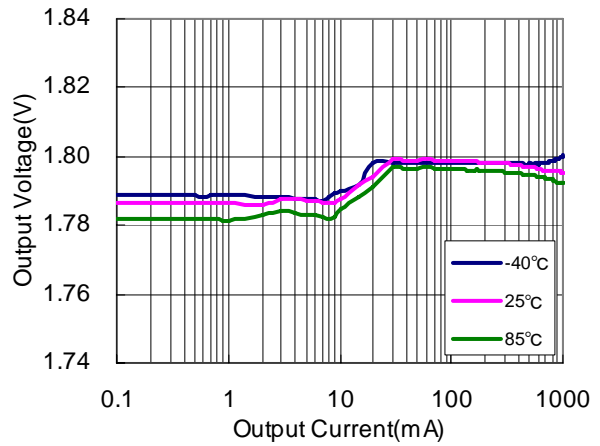
RP501K181x(PWM fixed)

V<sub>IN</sub>=3.6V,MODE=0V



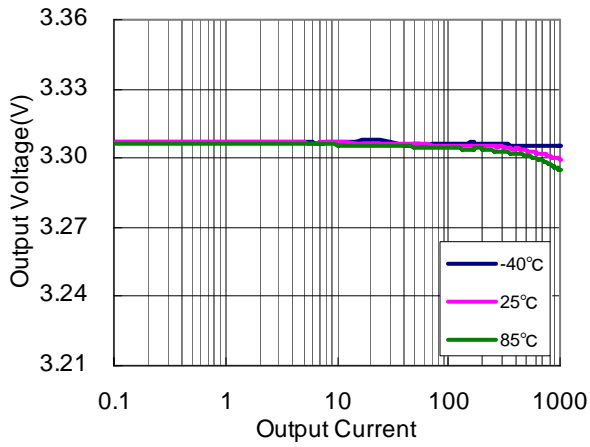
RP501K181x(PWM/VFM mode change)

V<sub>IN</sub>=MODE=3.6V



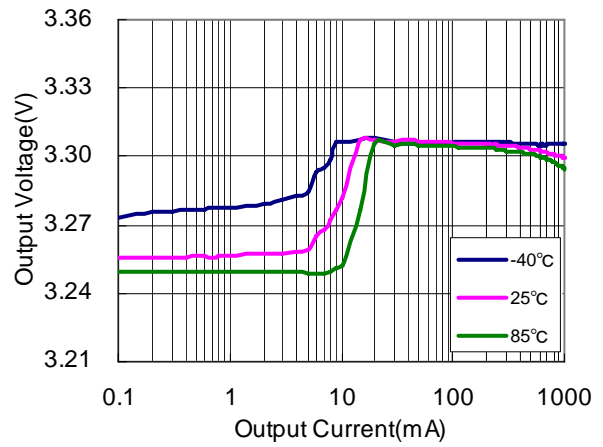
**RP501K331x(PWM fixed)**

$V_{IN}=5.0V, MODE=0V$



**RP501K331x(PWM/VFM mode change)**

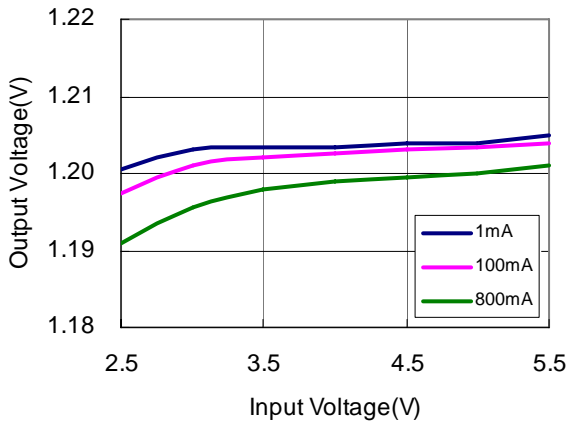
$V_{IN}=MODE=5.0V$



**3) Output Voltage vs. Input Voltage**

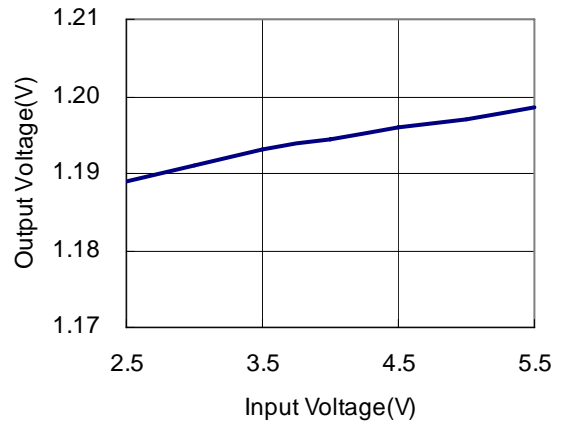
**RP501K121x(PWM fixed)**

$MODE=0V$



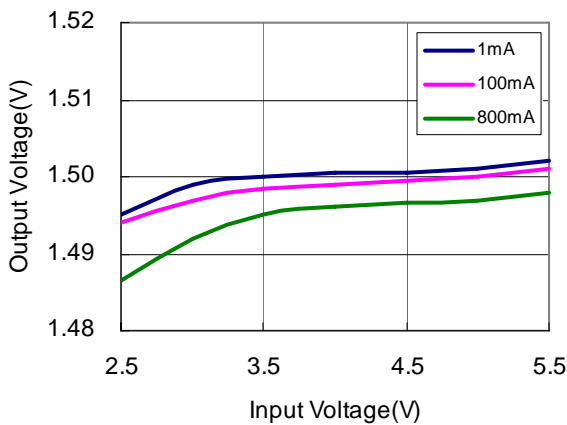
**RP501K121x(PWM/VFM mode change)**

$MODE=V_{IN}, I_{OUT}=1mA$



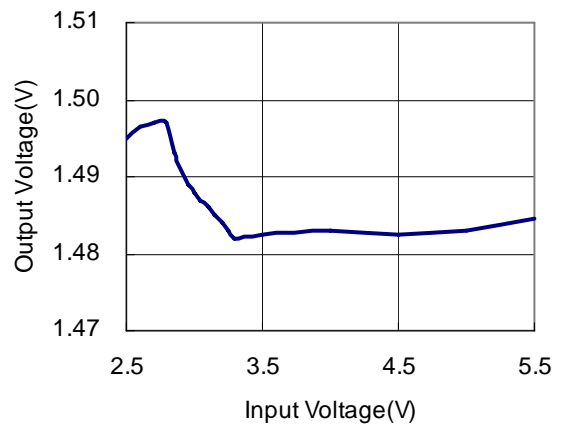
**RP501K151x(PWM fixed)**

$MODE=0V$

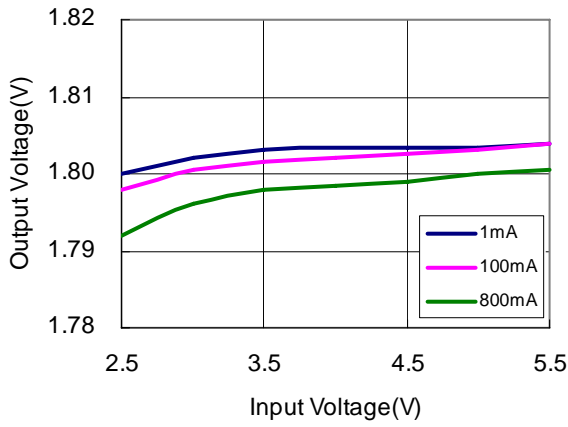


**RP501K151x(PWM/VFM mode change)**

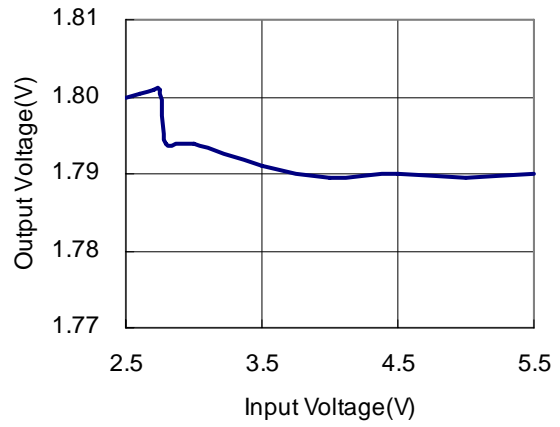
$MODE=V_{IN}, I_{OUT}=1mA$



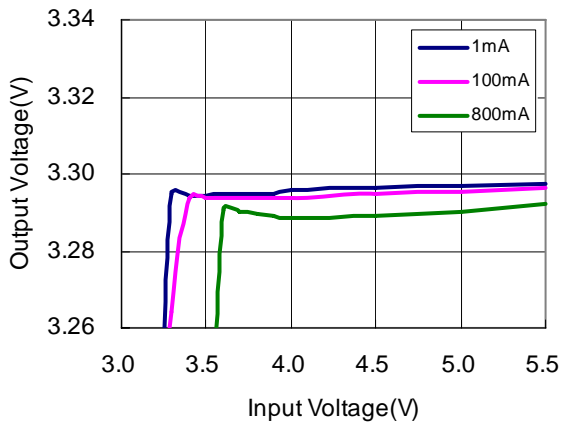
**RP501K181x(PWM fixed)**  
MODE=0V



**RP501K181x(PWM/VFM modechange)**  
MODE=VIN ,I<sub>OUT</sub>=1mA



**RP501K331x(PWM fixed)**  
MODE=0V

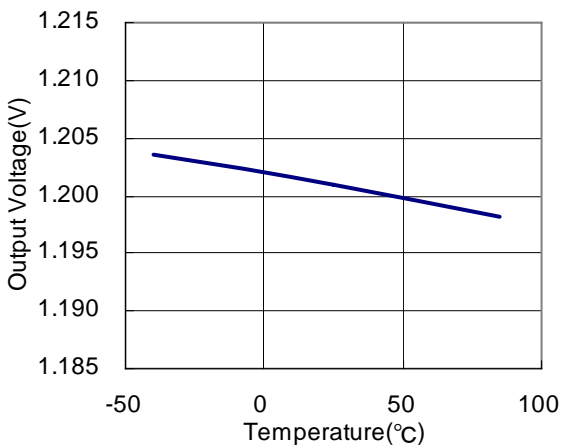


**RP501K331x(PWM/VFM mode change)**  
MODE=VIN ,I<sub>OUT</sub>=1mA

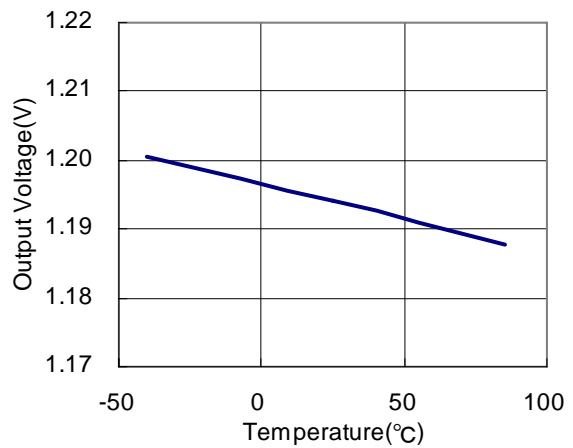


**4) Output Voltage vs. Temperature**

**RP501K121x(PWM fixed)**  
V<sub>IN</sub>=3.6V,MODE=0V,I<sub>OUT</sub>=500mA

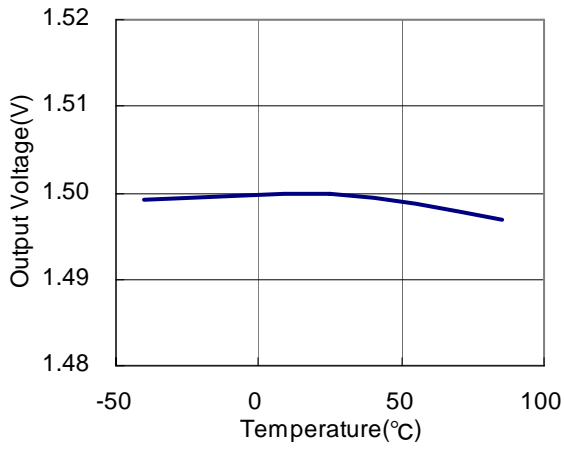


**RP501K121x(PWM/VFM mode change)**  
V<sub>IN</sub>=MODE=3.6V ,I<sub>OUT</sub>=1mA



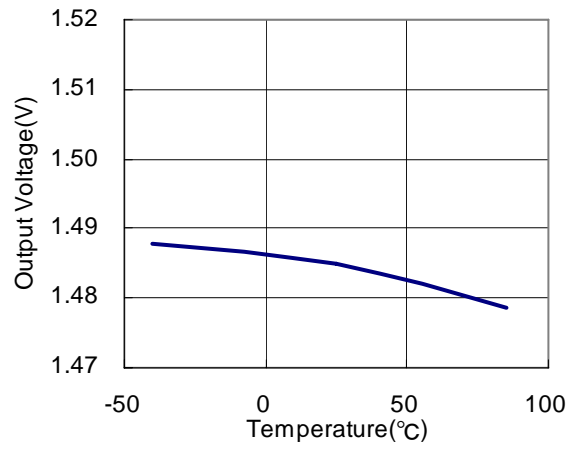
**RP501K151x(PWM fixed)**

$V_{IN}=3.6V, MODE=0V, I_{OUT}=500mA$



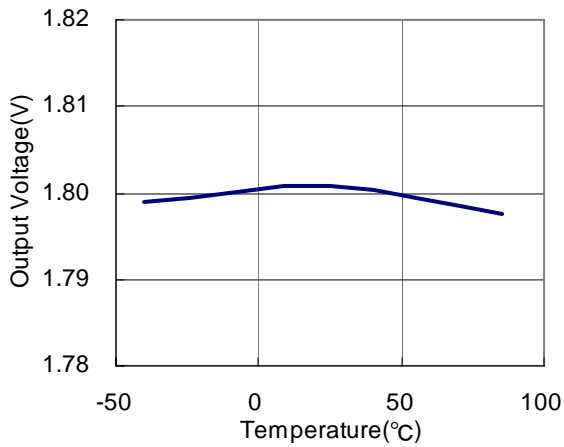
**RP501K151x(PWM/VFM mode change)**

$V_{IN}=MODE=3.6V, I_{OUT}=1mA$



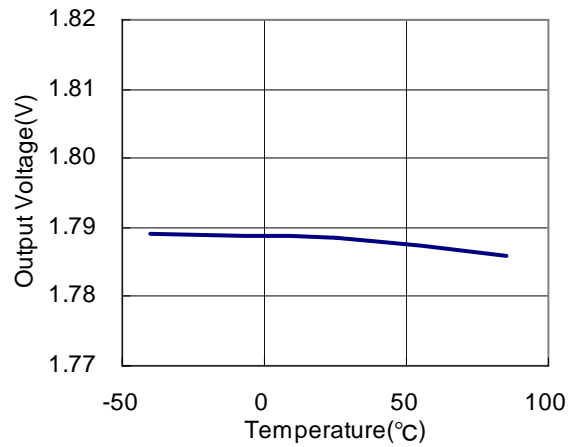
**RP501K181x(PWM fixed)**

$V_{IN}=3.6V, MODE=0V, I_{OUT}=500mA$



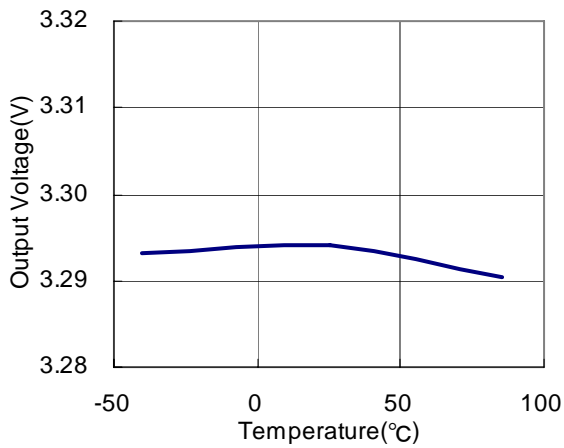
**RP501K181x(PWM/VFM mode change)**

$V_{IN}=MODE=3.6V, I_{OUT}=1mA$



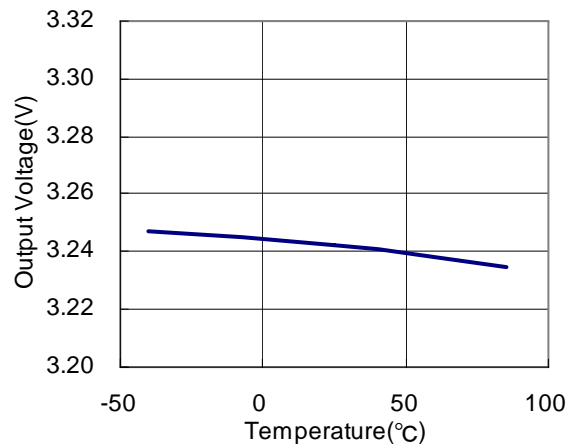
**RP501K331x(PWM fixed)**

$V_{IN}=5.0V, MODE=0V, I_{OUT}=500mA$



**RP501K331x(PWM/VFM mode change)**

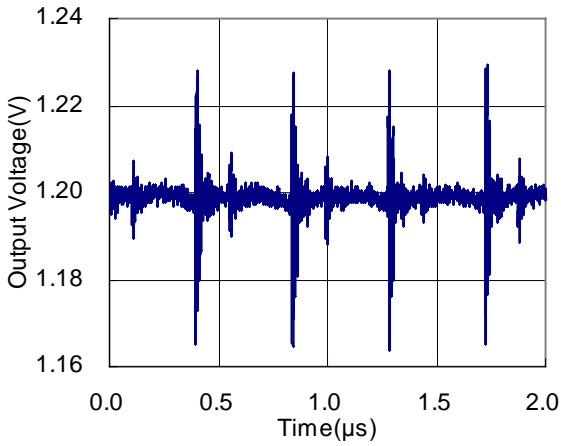
$V_{IN}=MODE=5.0V, I_{OUT}=1mA$



5) DC/DC Output Waveform ( $C_{IN}=10\mu F$ ,  $C_{OUT}=4.7\mu F$ ,  $L=2.2\mu H$ )

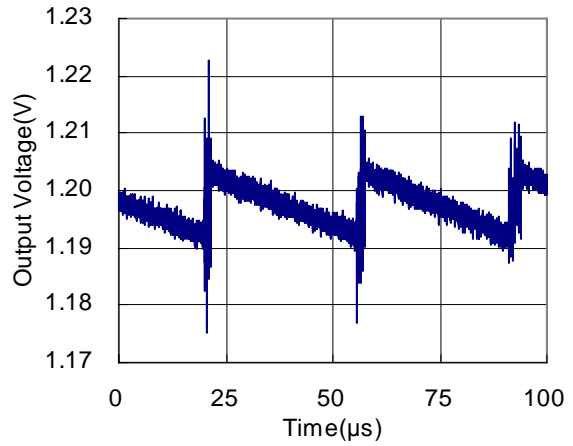
RP501K121x(PWM fixed)

$V_{IN}=3.6V, MODE=0V, R_{OUT}=10ohm$



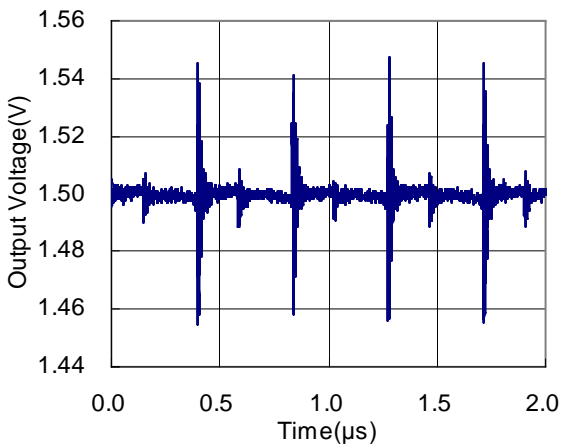
RP501K121x(PWM/VFM mode change)

$V_{IN}=MODE=3.6V, R_{OUT}=1kohm$



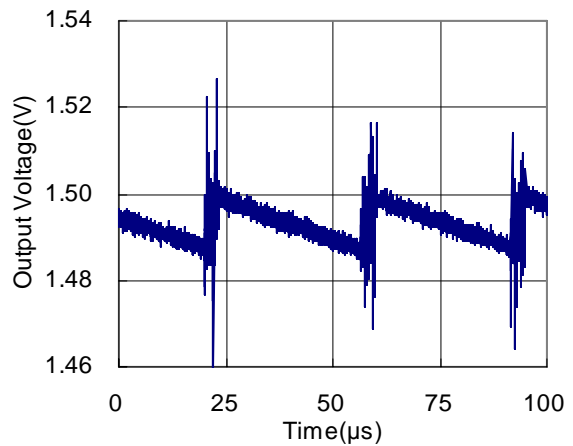
RP501K151x(PWM fixed)

$V_{IN}=3.6V, MODE=0V, R_{OUT}=10ohm$



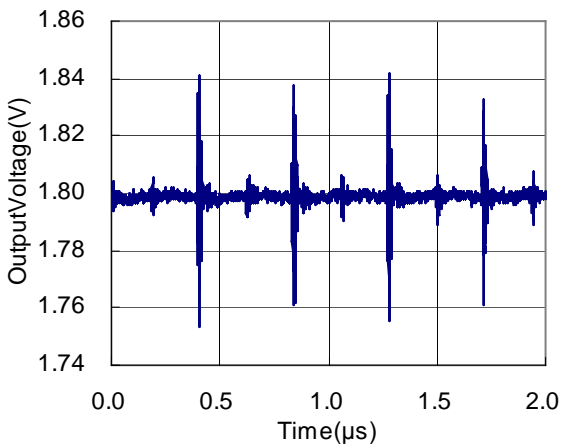
RP501K151x(PWM/VFM mode change)

$V_{IN}=MODE=3.6V, R_{OUT}=1kohm$



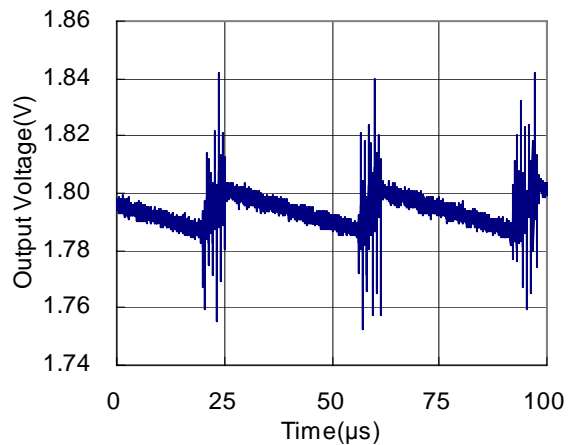
RP501K181x(PWM fixed)

$V_{IN}=3.6V, MODE=0V, R_{OUT}=10ohm$



RP501K181x(PWM/VFM mode change)

$V_{IN}=MODE=3.6V, R_{OUT}=1kohm$





**RP501K331x(PWM fixed)**

V<sub>IN</sub>=5.0V,MODE=0V,R<sub>OUT</sub>=10ohm



**RP501K331x(PWM/VFM mode change)**

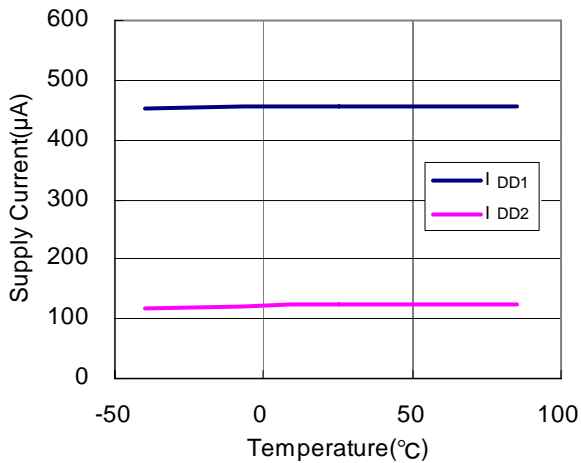
V<sub>IN</sub>=MODE=5.0V ,R<sub>OUT</sub>=1kohm



**6) Supply Current 1, 2 vs. Temperature**

**RP501Kxx1x**

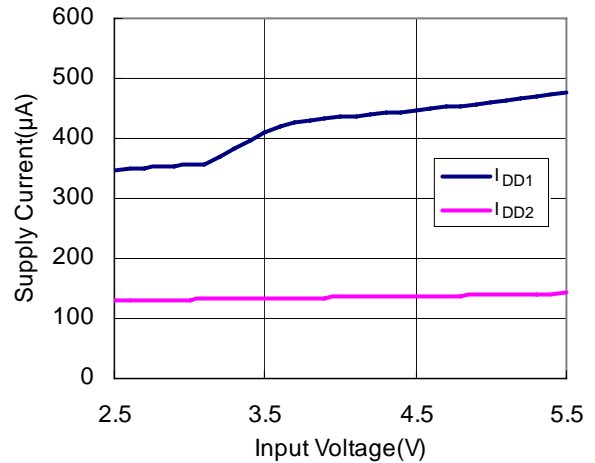
V<sub>IN</sub>=5.5V



**7) Supply Current 1, 2 vs. Input Voltage**

**RP501Kxx1x**

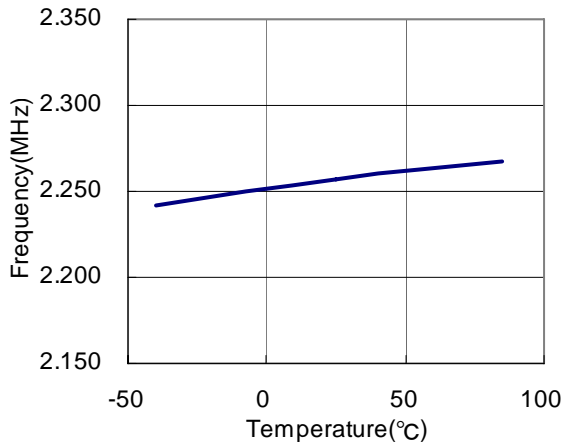
V<sub>IN</sub>=5.5V



**8) Oscillator Frequency vs. Temperature**

**RP501Kxx1x**

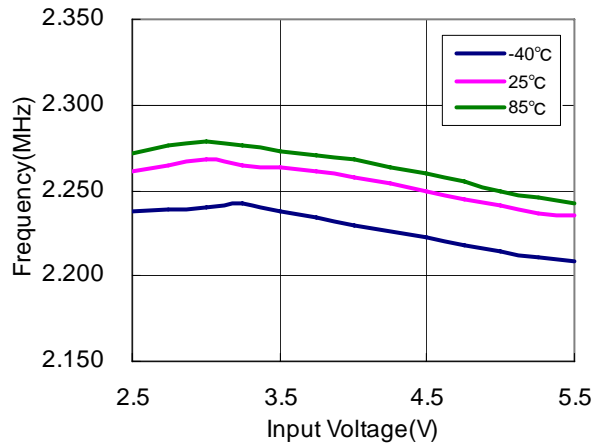
V<sub>IN</sub>=3.6V,R<sub>OUT</sub>=10ohm



**9) Oscillator Frequency vs. Input Voltage**

**RP501Kxx1x**

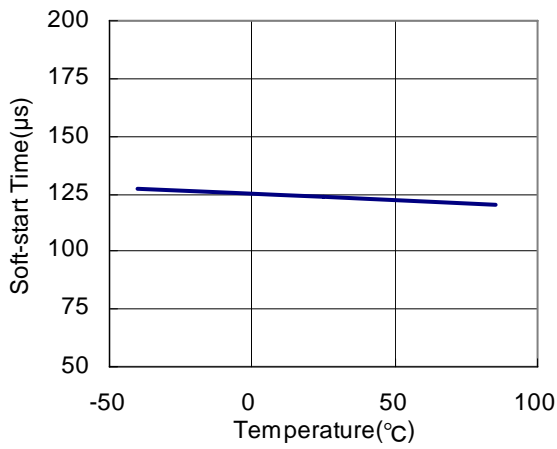
V<sub>IN</sub>=3.6V,R<sub>OUT</sub>=10ohm



10) Soft-start Time vs. Temperature

RP501Kxx1x

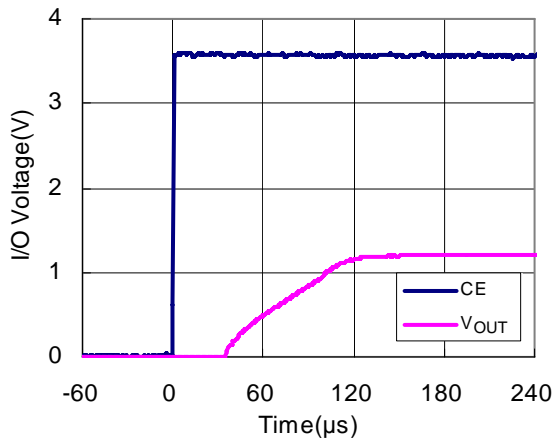
V<sub>IN</sub>=3.6V



11) Turn on waveform with CE pin (C<sub>IN</sub>=10µF, C<sub>OUT</sub>= 4.7µF, L=2.2µH)

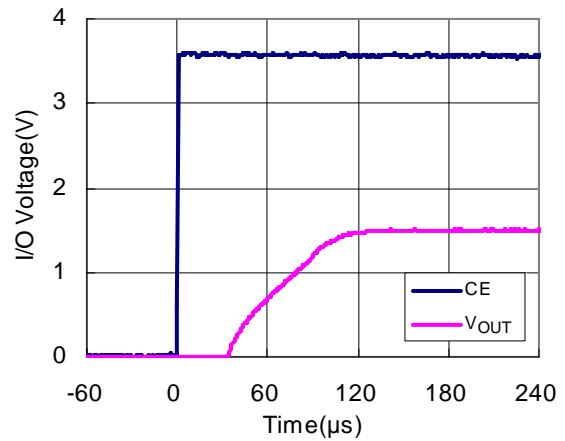
RP501K121x

V<sub>IN</sub>=3.6V, R<sub>OUT</sub>=100ohm



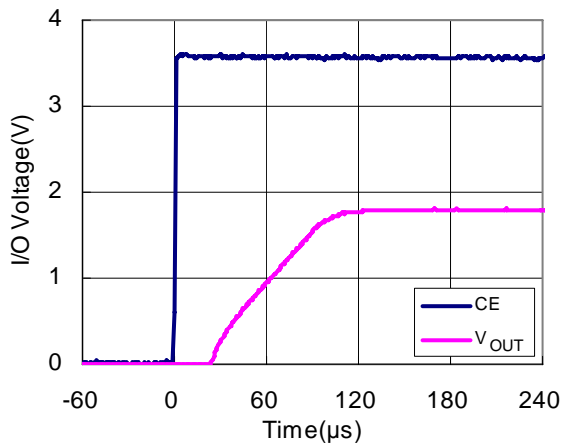
RP501K151x

V<sub>IN</sub>=3.6V, R<sub>OUT</sub>=100ohm



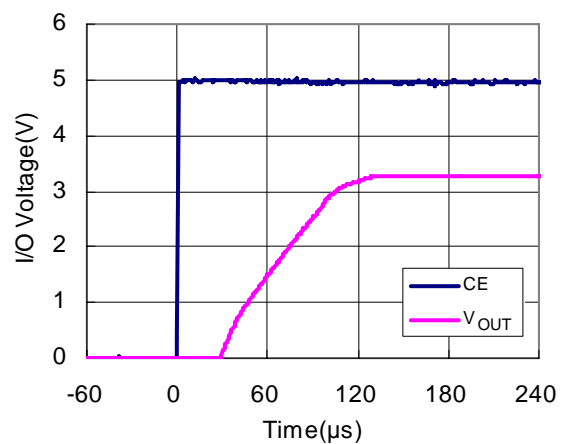
RP501K181x

V<sub>IN</sub>=3.6V, R<sub>OUT</sub>=100ohm



RP501K331x

V<sub>IN</sub>=5.0V, R<sub>OUT</sub>=100ohm



12) Turn off waveform with CE pin ( $C_{IN}=10\mu F$ ,  $C_{OUT}=4.7\mu F$ ,  $L=2.2\mu H$ )

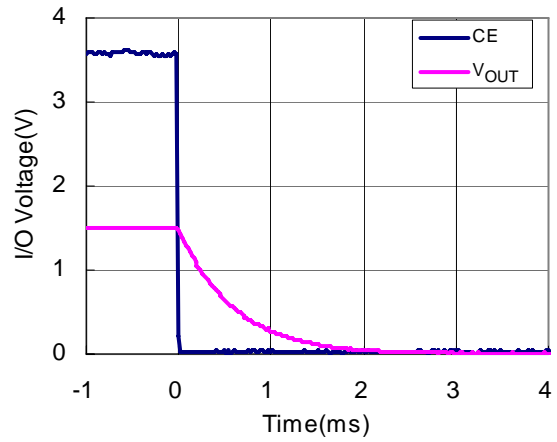
RP501K121B

$V_{IN}=3.6V, R_{OUT}=1k\Omega$



RP501K151B

$V_{IN}=3.6V, R_{OUT}=1k\Omega$



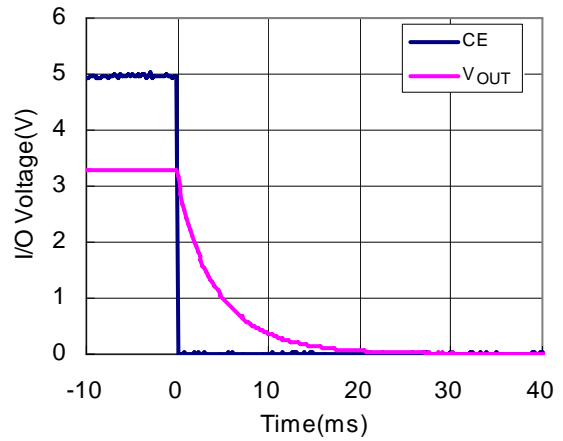
RP501K181A

$V_{IN}=3.6V, R_{OUT}=1k\Omega$



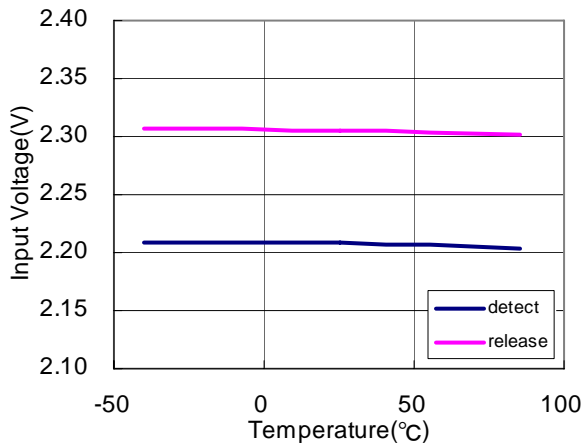
RP501K331A

$V_{IN}=5.0V, R_{OUT}=1k\Omega$



13) UVLO Detector Threshold/ Released Voltage vs. Temperature

RP501Kxx1x



14) CE Input vs. Temperature

RP501Kxx1x

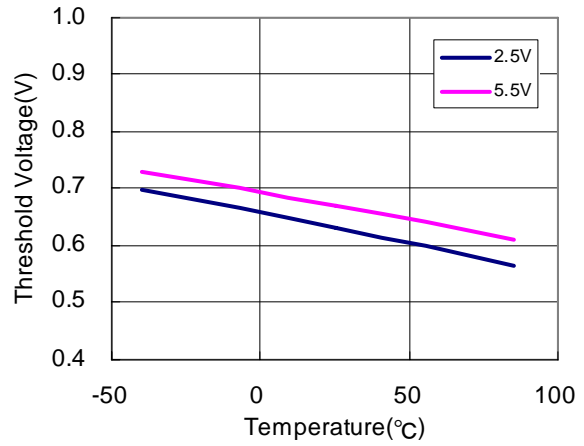
Input Voltage Dependence



15) MODE Input vs. Temperature

RP501Kxx1x

Input Voltage Dependence



16) Lx Current Limit vs. Temperature

RP501Kxx1x

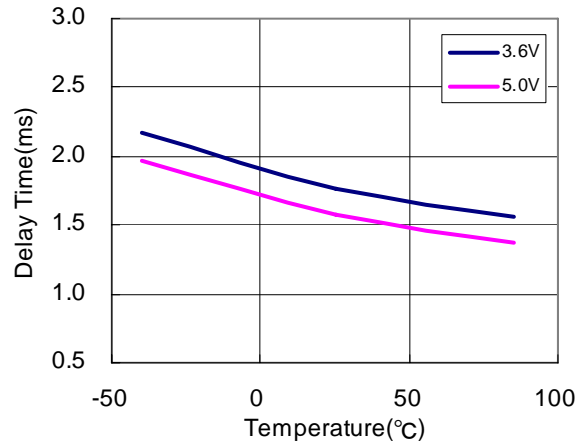
Input Voltage Dependence



17) Protection Delay Time VS. Temperature

RP501Kxx1x

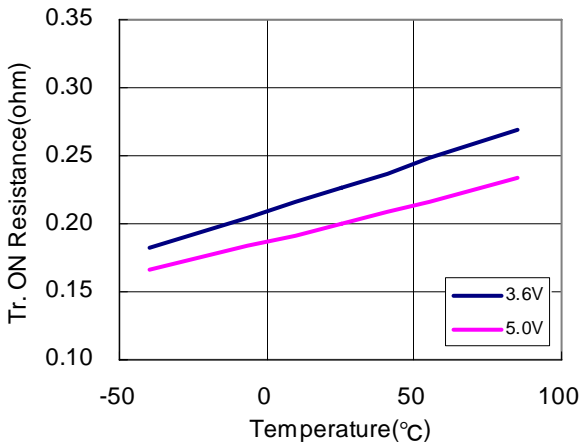
Input Voltage Dependence



18) Nch Tr. ON Resistance vs. Temperature

RP501Kxx1x

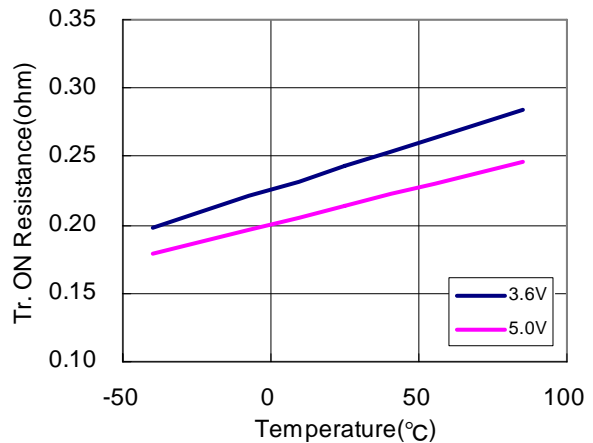
Input Voltage Dependence, I<sub>OUT</sub>=100mA



19) Pch Tr. ON Resistance vs. Temperature

RP501Kxx1x

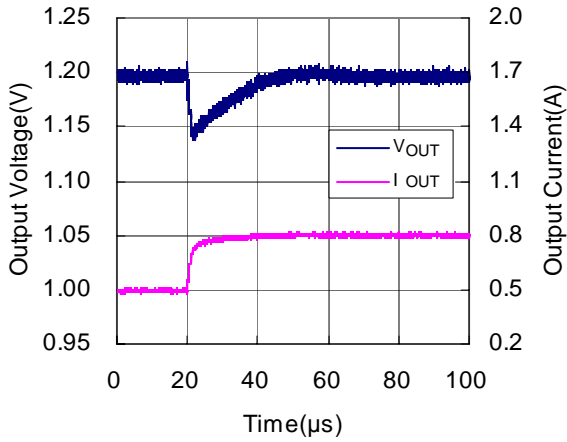
Input Voltage Dependence, I<sub>OUT</sub>=100mA



20) Load Transient Response ( $C_{IN}=10\mu F$ ,  $C_{OUT}=4.7\mu F$ ,  $L=2.2\mu H$ )

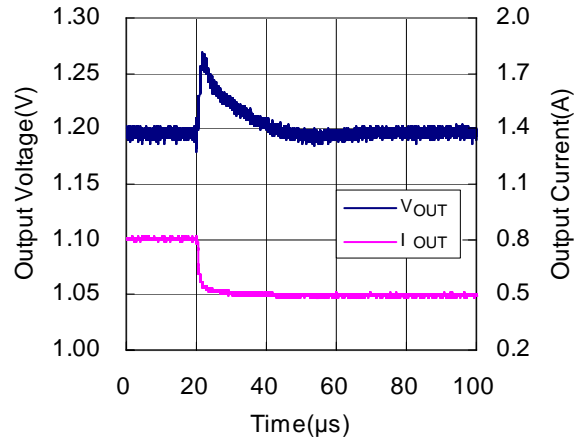
RP501K121x

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.5A$  to  $0.8A$



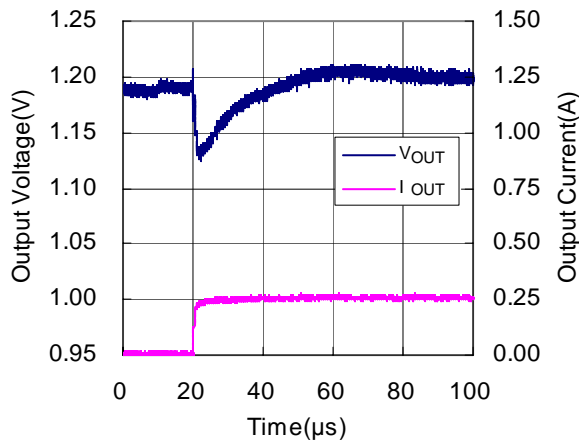
RP501K121x

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.8A$  to  $0.5A$



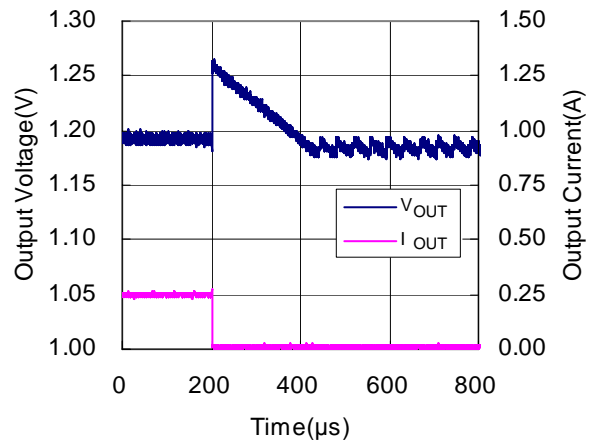
RP501K121x

$V_{IN}=MODE=3.6V, I_{OUT}=0.001A$  to  $0.25A$



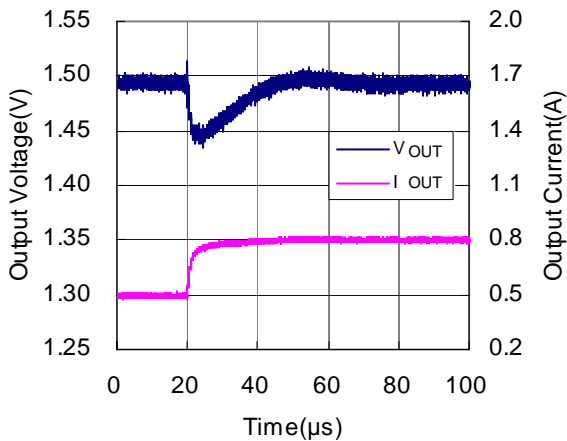
RP501K121x

$V_{IN}=MODE=3.6V, I_{OUT}=0.25A$  to  $0.001A$



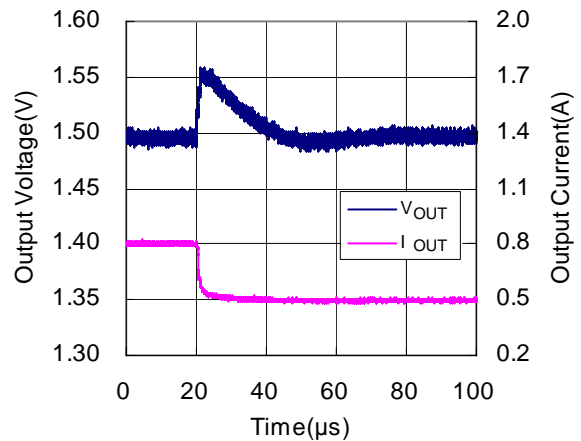
RP501K151x

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.5A$  to  $0.8A$



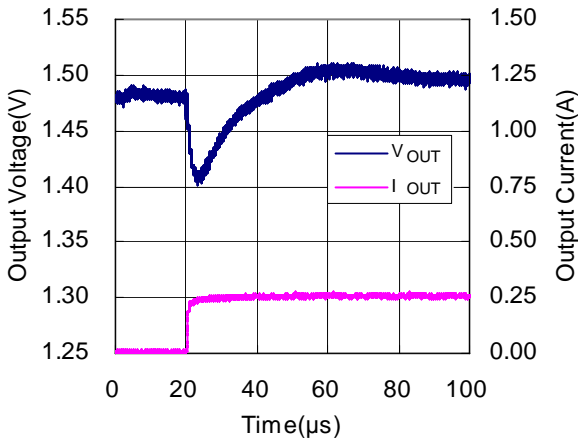
RP501K151x

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.8A$  to  $0.5A$



**RP501K151x**

$V_{IN}=MODE=3.6V, I_{OUT}=0.001A$  to  $0.25A$



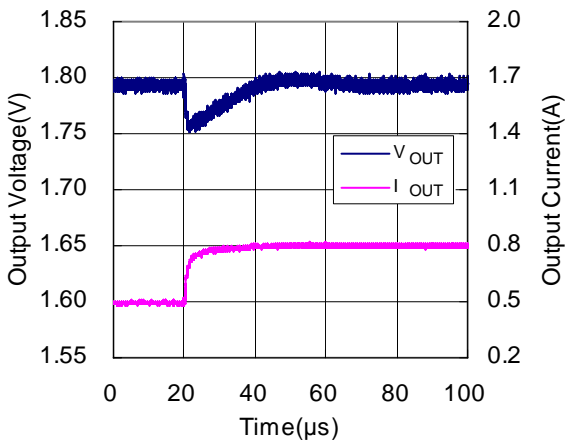
**RP501K151x**

$V_{IN}=MODE=3.6V, I_{OUT}=0.25A$  to  $0.001A$



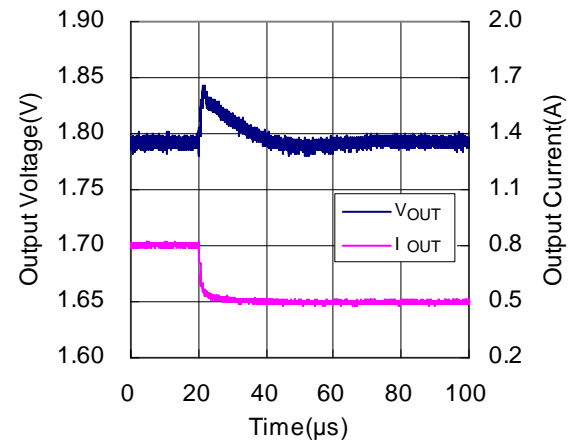
**RP501K181x**

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.5A$  to  $0.8A$



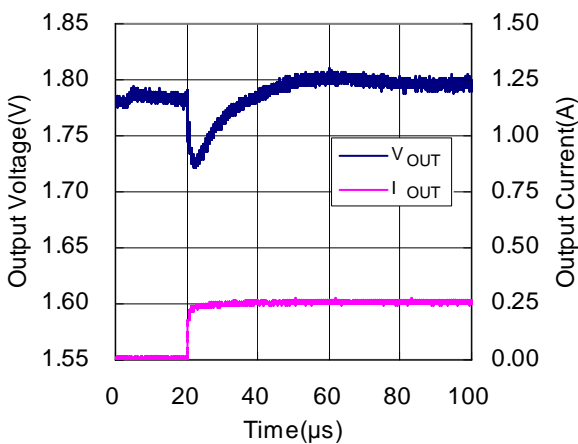
**RP501K181x**

$V_{IN}=3.6V, MODE=0V, I_{OUT}=0.8A$  to  $0.5A$



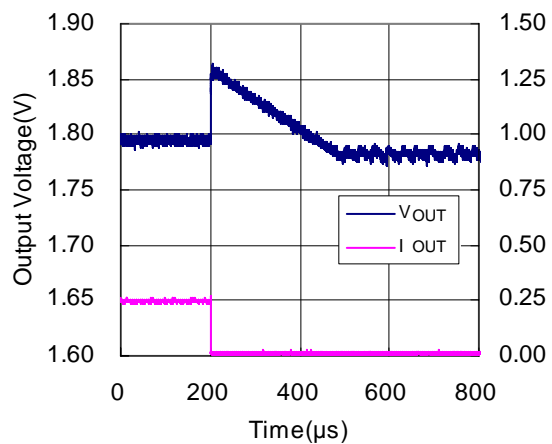
**RP501K181x**

$V_{IN}=MODE=3.6V, I_{OUT}=0.001A$  to  $0.25A$



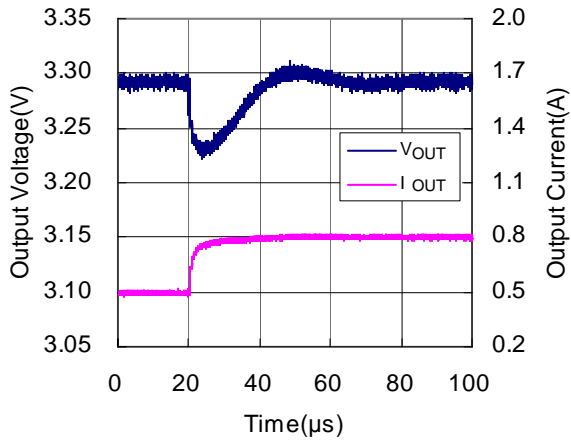
**RP501K181x**

$V_{IN}=MODE=3.6V, I_{OUT}=0.25A$  to  $0.001A$



**RP501K331x**

$V_{IN}=5.0V, MODE=0V, I_{OUT}=0.5A$  to  $0.8A$



**RP501K331x**

$V_{IN}=5.0V, MODE=0V, I_{OUT}=0.8A$  to  $0.5A$



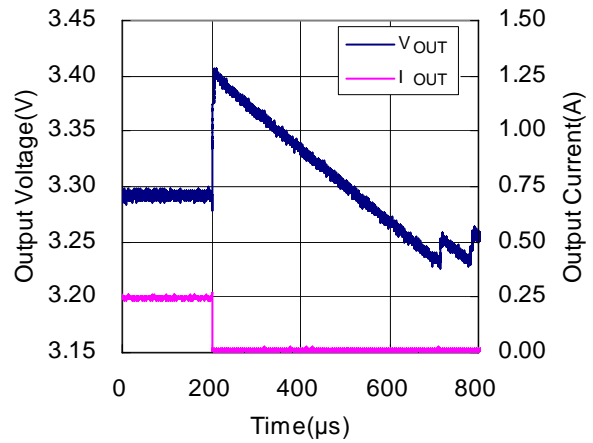
**RP501K331x**

$V_{IN}=MODE=5.0V, I_{OUT}=0.001A$  to  $0.25A$



**RP501K331x**

$V_{IN}=MODE=5.0V, I_{OUT}=0.25A$  to  $0.001A$





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