

### ISL28233, ISL28433

Dual and Quad Micropower Chopper Stabilized, RRIO Operational Amplifiers

FN7692 Rev 3.00 July 26, 2011

The ISL28233 and ISL28433 are dual and quad micropower, chopper stabilized operational amplifiers that are optimized for single and dual supply operation from 1.8V to 6.0V and  $\pm 0.825 V$  to  $\pm 3.0 V$ . Their low supply current of  $18 \mu A$  and wide input range enable the ISL28233, ISL28433 to be excellent general purpose op amps for a wide range of applications. The ISL28233 and ISL28433 are ideal for handheld devices that operate off 2 AA or single Li-ion batteries.

The ISL28233 is available in 8 Ld MSOP, 8 Ld SOIC and 8 Ld DFN packages. The ISL28433 is available in 14 Ld TSSOP, 14 Ld SOIC and 14 Ld 3mmx4mm TDFN packages. All devices operate over the temperature range of -40°C to +125°C.

#### **Related Literature**

- See <u>AN1596</u>, "ISL28233SOICEVAL1Z Evaluation Board User's Guide"
- See <u>AN1575</u>, "ISL28433SOICEVAL1Z, ISL28433TSSOPEVAL1Z Evaluation Board User's Guide"

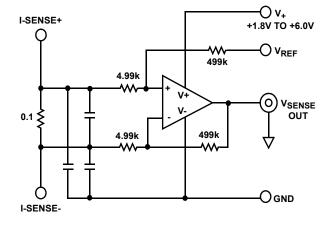
#### **Features**

Low Input Offset Voltage	6μV, Max.
Low Offset Drift	0.05μV/°C, Max.
• Quiescent Current (Per Amplifier)	18μΑ, Typ.
Single Supply Range	+1.8V to +6.0V
Dual Supply Range	±0.825V to ±3.0V
• Low Noise (0.01Hz to 10Hz)	1.0μV <sub>P-P</sub> , Typ.
Rail-to-Rail Inputs and Output	
Input Bias Current	180pA, Max.
Operating Temperature Range	40°C to +125°C

### **Applications**

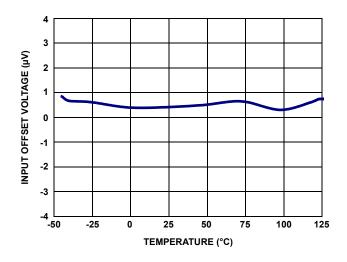
- Bi-Directional Current Sense
- Temperature Measurement
- Medical Equipment
- Electronic Weigh Scales
- Precision/Strain Gauge Sensor
- · Precision Regulation
- Low Ohmic Current Sense
- . High Gain Analog Front Ends

## **Typical Application**



**BI-DIRECTIONAL CURRENT SENSE AMPLIFIER** 

## **V<sub>OS</sub>** vs Temperature

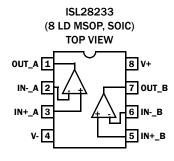


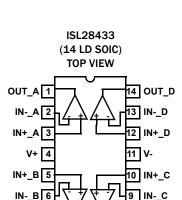
## **Ordering Information**

PART NUMBER (Notes 1, 2, 3)	PART MARKING	PACKAGE (Pb-Free)	PKG. DWG. #			
ISL28233FUZ	233FZ	8 Ld MSOP	M8.118A			
ISL28233FRZ	233Z	8 Ld 3mmx3mm DFN	L8.3x3J			
ISL28233FBZ	28233 FBZ	8 Ld SOIC	M8.15E			
ISL28433FBZ	28433 FBZ	14 Ld SOIC	MDP0027			
ISL28433FVZ	28433 FVZ	14 Ld TSSOP	MDP0044			
Coming Soon ISL28433FRTZ	TBD	14 Ld 3x4 mm TDFN	TBD			
ISL28233SOICEVAL1Z	Evaluation Board					
ISL28433TSSOPEVAL1Z	Evaluation Board					
ISL28433SOICEVAL1Z	Evaluation Board					

- 1. Add "-T\*" suffix for tape and reel. Please refer to TB347 for details on reel specifications.
- These Intersil Pb-free plastic packaged products employ special Pb-free material sets, molding compounds/die attach materials, and 100% matte tin plate
  plus anneal (e3 termination finish, which is RoHS compliant and compatible with both SnPb and Pb-free soldering operations). Intersil Pb-free products are
  MSL classified at Pb-free peak reflow temperatures that meet or exceed the Pb-free requirements of IPC/JEDEC J STD-020.
- 3. For Moisture Sensitivity Level (MSL), please see device information page for <u>ISL28233</u>, <u>ISL28433</u>. For more information on MSL please see techbrief <u>TB363</u>.

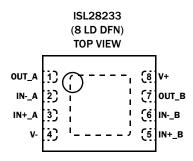
## **Pin Configurations**

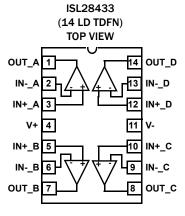


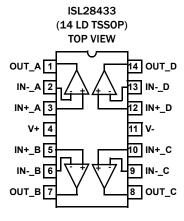


8 OUT\_C

OUT\_B







## **Pin Descriptions**

ISL28233 (8 LD MSOP, SOIC, DFN)	ISL28433 (14 LD TSSOP, SOIC, TDFN)	PIN NAME	FUNCTION	EQUIVALENT CIRCUIT
3	3	IN+_A	Non-inverting input	V+ + <b>&gt;</b>
5	5	IN+_B		<b>1</b>
-	10	IN+_C		IN+ D
-	12	IN+_D		IN- CLOCK GEN + DRIVERS  Circuit 1
4	11	V-	Negative supply	Circuit 1
2	2	INA	Inverting input	(See Circuit 1)
6	6	INB		
-	9	INC		
-	13	IND		
1	1	OUT_A	Output	V+
7	7	OUT_B		│ ···
-	8	OUT_C		OUT
-	14	OUT_D		Circuit 2
8	4	V+	Positive supply	
-	-	PAD	Thermal Pad	Thermal Pad. Connect to most negative supply. TDFN and DFN packages only.

#### **Absolute Maximum Ratings**

Max Supply Voltage V+ to V	6.5V
Max Voltage VIN to GND(V 0.3V) to (V+ +	0.3V)V
Max Input Differential Voltage	. 6.5V
Max Input Current	20mA
Max Voltage VOUT to GND (10s)	.±3.0V
ESD Tolerance	
Human Body Model (Tested per JESD22-A114F)	4000V
Machine Model (Tested per JESD22-A115B)	. 400V
Charged Device Model (Tested per JESD22-C110D)	2000V
Latch-Up (Tested per JESD78B) +1	125°C

#### **Thermal Information**

Thermal Resistance (Typical)	θ <sub>JA</sub> (°C/W)	θ <sub>JC</sub> (°C/W)
14 Ld TSSOP (Notes 4, 7)	110	40
14 Ld SOIC (Notes 4, 7)	75	47
14 Ld TDFN (Notes 5, 6)	TBD	TBD
8 Ld MSOP (Notes 4, 7)	180	65
8 Ld SOIC (Notes 4, 7)	125	90
8 Ld DFN (Notes 5, 6)	53	12
Maximum Storage Temperature Range	6!	5°C to +150°C
Pb-Free Reflow Profile		see link below
http://www.intersil.com/pbfree/Pb-FreeRe	eflow.asp	

#### **Operating Conditions**

Temperature Range ......-40°C to +125°C

CAUTION: Do not operate at or near the maximum ratings listed for extended periods of time. Exposure to such conditions may adversely impact product reliability and result in failures not covered by warranty.

#### NOTES:

- 4. θ<sub>JA</sub> is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief TB379 for details.
- 5. θ<sub>JA</sub> is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief TB379.
- 6. For  $\theta_{JC}$ , the "case temp" location is the center of the exposed metal pad on the package underside.
- 7. For  $\theta_{\mbox{\scriptsize JC}},$  the "case temp" location is taken at the package top center.

**Electrical Specifications** V+ = 5V, V- = 0V,  $V_{CM}$  = 2.5V,  $T_A$  = +25°C,  $R_L$  = 10k $\Omega$ , unless otherwise specified. **Boldface limits apply over** the operating temperature range,-40°C to +125°C.

PARAMETER	DESCRIPTION	CONDITIONS	MIN (Note 8)	ТҮР	MAX (Note 8)	UNIT
DC SPECIFICATIO	NS	·	·			
V <sub>OS</sub>	Input Offset Voltage		-6	±2	6	μV
		T = -40°C to +125°C	-11	-	11	μV
TCV <sub>OS</sub>	Input Offset Voltage Temperature Coefficient	T = -40°C to +125°C	-0.05	0.01	0.05	μV/°C
I <sub>OS</sub>	Input Offset Current		-	10	-	pA
TCI <sub>OS</sub>	Input Offset Current Temperature Coefficient	T = -40°C to +85°C	-	0.11	-	pA/°C
Ι <sub>Β</sub>	Input Bias Current	T = -40 °C to +85 °C	-180	-	180	pA
		T = -40°C to +125°C	-600	-	600	pA
TCIB	Input Bias Current Temperature Coefficient	T = -40°C to +85°C	-	0.49	-	pA/°C
CMIR		V+ = 5.0V, V- = 0V Guaranteed by CMRR	-0.1	-	5.1	V
CMRR	Common Mode Rejection Ratio	VCM = -0.1V to 5.1V	118	125	-	dB
			115	-	-	dB
PSRR	Power Supply Rejection Ratio	Vs = 1.8V to 6.0V	110	138	-	dB
			110	-	-	dB
V <sub>OH</sub>	Output Voltage, High		4.965	4.981	-	٧
V <sub>OL</sub>	Output Voltage, Low		-	18	35	m۷
A <sub>OL</sub>	Open Loop Gain	$R_L = 1M\Omega$	-	174	-	dB
V+	Supply Voltage	Guaranteed by PSRR	1.8	-	6.0	٧
Is	Supply Current, Per Amplifier	R <sub>L</sub> = OPEN	-	18	25	μΑ
			-	-	35	μA



# **Electrical Specifications** V+ = 5V, V- = 0V, $V_{CM}$ = 2.5V, $T_A$ = +25°C, $R_L$ = 10k $\Omega$ , unless otherwise specified. Boldface limits apply over the operating temperature range,-40°C to +125°C. (Continued)

PARAMETER	DESCRIPTION	conditions	MIN (Note 8)	TYP	MAX (Note 8)	UNIT
I <sub>SC+</sub>	Output Source Short Circuit Current	R <sub>L</sub> = Short to V-	13	17	26	mA
I <sub>SC-</sub>	Output Sink Short Circuit Current	R <sub>L</sub> = Short to V+	-26	-19	-13	mA
AC SPECIFICATION	S		l	I	Ш	J.
GBWP	Gain Bandwidth Product	$\begin{aligned} & \textbf{A}_{\text{V}} = \textbf{100}, \textbf{R}_{\text{F}} = \textbf{100} \textbf{k} \boldsymbol{\Omega}, \textbf{R}_{\text{G}} = \textbf{1} \textbf{k} \boldsymbol{\Omega}, \\ & \textbf{R}_{\text{L}} = \textbf{10} \textbf{k} \boldsymbol{\Omega} \text{ to } \textbf{V}_{\text{CM}} \end{aligned}$	-	400	-	kHz
e <sub>N</sub> V <sub>P-P</sub>	Peak-to-Peak Input Noise Voltage	f = 0.01Hz to 10Hz	-	1.0	-	μV <sub>P-P</sub>
e <sub>N</sub>	Input Noise Voltage Density	f = 1kHz	-	65	-	nV/√(Hz)
iN	Input Noise Current Density	f = 1kHz	-	72	-	fA/√(Hz)
		f = 10Hz	-	79	-	fA/√(Hz)
C <sub>in</sub>	Differential Input Capacitance	f = 1MHz	-	1.6	-	pF
	Common Mode Input Capacitance		-	1.12	-	pF
TRANSIENT RESPO	DNSE				1	
SR	Positive Slew Rate	$V_{OUT}$ = 1V to 4V, $R_L$ = 10k $\Omega$	-	0.2	-	V/µs
	Negative Slew Rate		-	0.1	-	V/µs
t <sub>r</sub> , t <sub>f</sub> , Small Signal	Rise Time, t <sub>r</sub> 10% to 90%	$A_V = +1, V_{OUT} = 0.1V_{P-P}, R_F = 0\Omega,$	-	1.1	-	μs
	Fall Time, t <sub>f</sub> 10% to 90%	$R_L = 10k\Omega$ , $C_L = 1.2pF$	-	1.1	-	μs
t <sub>r</sub> , t <sub>f</sub> Large Signal	Rise Time, t <sub>r</sub> 10% to 90%	$A_V = +1$ , $V_{OUT} = 2V_{P-P}$ , $R_F = 0\Omega$ ,	-	20	-	μs
	Fall Time, t <sub>f</sub> 10% to 90%	$R_L = 10k\Omega$ , $C_L = 1.2pF$	-	30	-	μs
t <sub>s</sub>	Settling Time to 0.1%, 2V <sub>P-P</sub> Step	$A_V = +1, R_F = 0\Omega, R_L = 10k\Omega, C_L = 1.2pF$	-	35	-	μs
t <sub>recover</sub>	Output Overload Recovery Time, Recovery to 90% of output saturation	$A_V = +2$ , $R_F = 10$ k $\Omega$ , $R_L = 0$ pen, $C_L = 3.7$ pF	-	10.5	-	μs

<sup>8.</sup> Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or design.

# Typical Performance Curves V+=5V, V-=0V, $V_{CM}=2.5V$ , $R_L=Open$ , $T=+25\,^{\circ}C$ , unless otherwise specified.

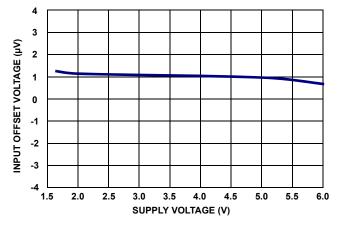


FIGURE 1. VOS vs SUPPLY VOLTAGE

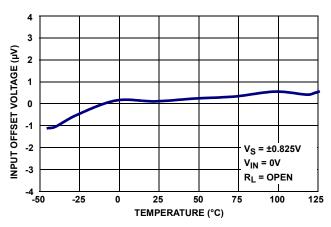


FIGURE 2. V<sub>OS</sub> vs TEMPERATURE

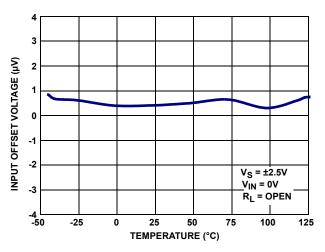


FIGURE 3. VOS VS TEMPERATURE

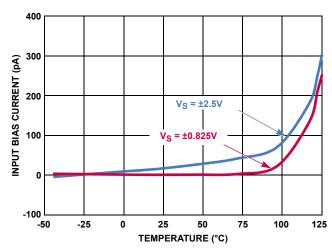


FIGURE 4. I<sub>B+</sub> vs TEMPERATURE

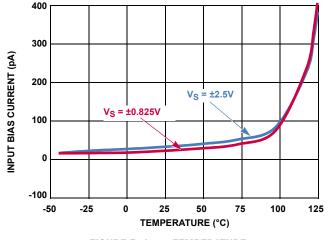


FIGURE 5.  $I_{B-}$  vs TEMPERATURE

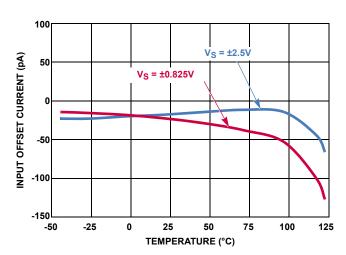


FIGURE 6. I<sub>OS</sub> vs TEMPERATURE

## Typical Performance Curves v+ = 5V, V- = 0V, V<sub>CM</sub> = 2.5V, R<sub>L</sub> = Open, T = +25 °C, unless otherwise specified. (Continued)

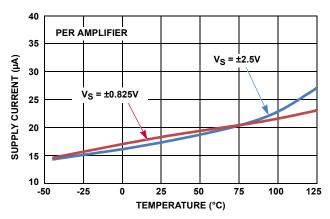


FIGURE 7. SUPPLY CURRENT vs TEMPERATURE

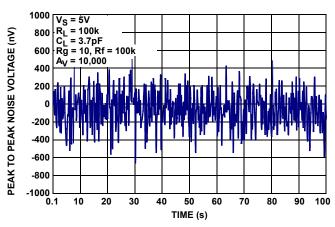


FIGURE 8. INPUT NOISE VOLTAGE 0.01Hz TO 10Hz

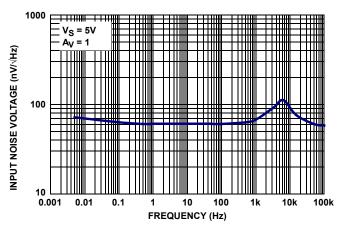


FIGURE 9. INPUT NOISE VOLTAGE DENSITY vs FREQUENCY

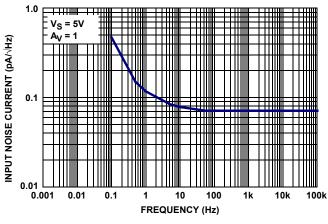


FIGURE 10. INPUT NOISE CURRENT DENSITY vs FREQUENCY

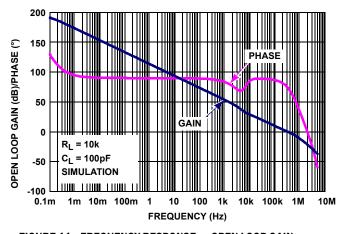


FIGURE 11. FREQUENCY RESPONSE vs OPEN LOOP GAIN,  $R_L = 10 k \Omega \label{eq:RL}$ 

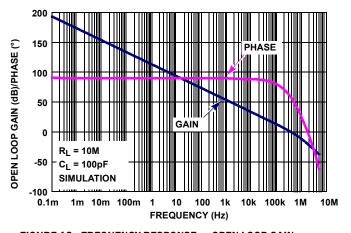


FIGURE 12. FREQUENCY RESPONSE vs OPEN LOOP GAIN,  $R_L = 10 M \Omega \label{eq:response}$ 

### Typical Performance Curves v+ = 5V, V- = 0V, V<sub>CM</sub> = 2.5V, R<sub>L</sub> = Open, T = +25°C, unless otherwise specified. (Continued)

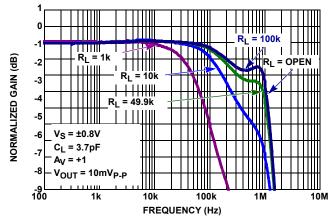


FIGURE 13. GAIN vs FREQUENCY vs  $R_L$ ,  $V_S = \pm 0.8V$ 

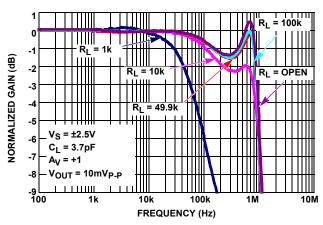


FIGURE 14. GAIN vs FREQUENCY vs  $R_L$ ,  $V_S = \pm 2.5V$ 

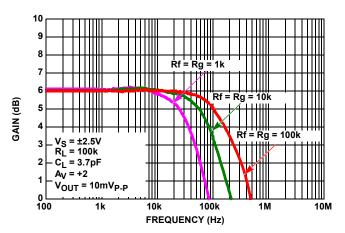


FIGURE 15. GAIN vs FREQUENCY vs FEEDBACK RESISTOR VALUES  $R_{\mbox{\scriptsize f}}/R_{\mbox{\scriptsize g}}$ 

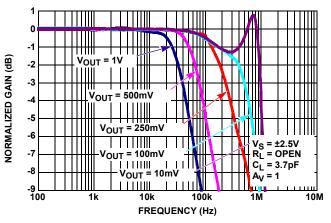


FIGURE 16. GAIN vs FREQUENCY vs V<sub>OUT.</sub> R<sub>L</sub> = OPEN

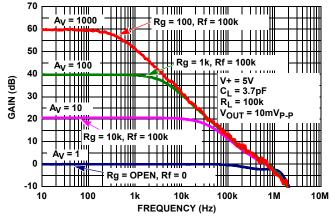


FIGURE 17. FREQUENCY RESPONSE vs CLOSED LOOP GAIN

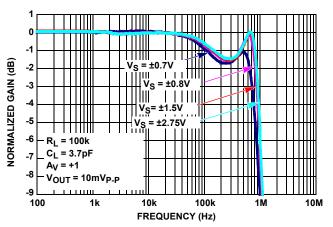


FIGURE 18. GAIN vs FREQUENCY vs SUPPLY VOLTAGE

## Typical Performance Curves v+ = 5V, V- = 0V, V<sub>CM</sub> = 2.5V, R<sub>L</sub> = Open, T = +25 °C, unless otherwise specified. (Continued)

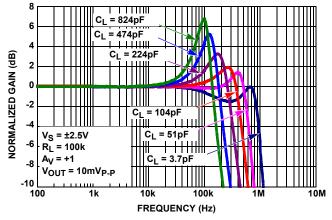


FIGURE 19. GAIN vs FREQUENCY vs CL

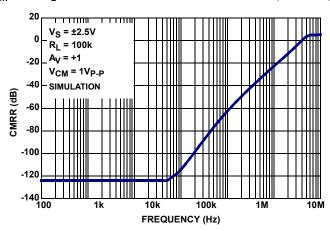


FIGURE 20. CMRR vs FREQUENCY, V<sub>S</sub> = ±2.5V

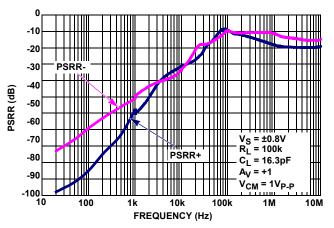


FIGURE 21. PSRR vs FREQUENCY,  $V_S = \pm 0.8V$ 

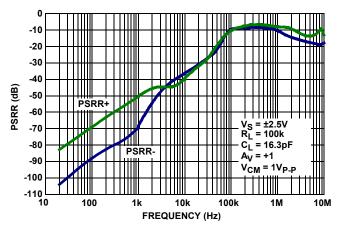
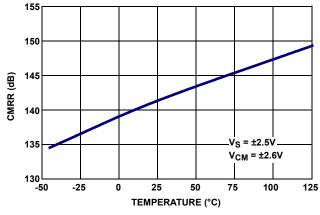


FIGURE 22. PSRR vs FREQUENCY, V<sub>S</sub> = ±2.5V





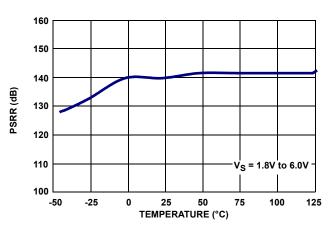


FIGURE 24. PSRR vs TEMPERATURE

## $\textbf{Typical Performance Curves} \ \ \text{V+ = 5V, V- = 0V, V}_{\text{CM}} = 2.5\text{V, R}_{\text{L}} = \text{Open, T = +25}\,^{\circ}\text{C, unless otherwise specified.} \ \ \textbf{(Continued)}$

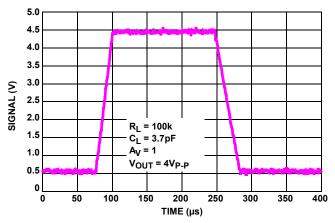


FIGURE 25. LARGE SIGNAL STEP RESPONSE (4V)

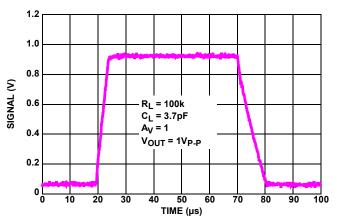


FIGURE 26. LARGE SIGNAL STEP RESPONSE (1V)

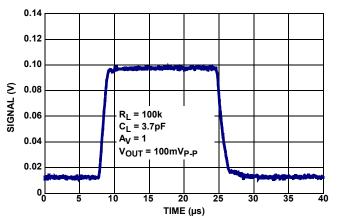
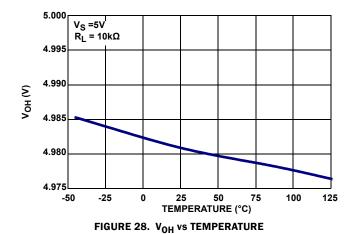


FIGURE 27. SMALL SIGNAL STEP RESPONSE (100mV)



V<sub>S</sub> = 5V R<sub>L</sub> = 10kΩ 35 30 25 20 25 20 15 -50 -25 0 25 50 75 100 125 TEMPERATURE (°C)

FIGURE 29.  $V_{OL}$  vs TEMPERATURE

## $\textbf{Typical Performance Curves} \ \, \text{V+} = 5 \text{V, V-} = 0 \text{V, V}_{\text{CM}} = 2.5 \text{V, R}_{\text{L}} = \text{Open, T} = +25 \, ^{\circ}\text{C, unless otherwise specified.} \ \, \textbf{(Continued)}$

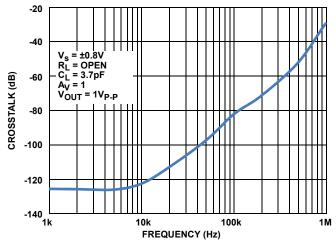


FIGURE 30. CROSSTALK vs FREQUENCY,  $V_S = \pm 0.8V$ 

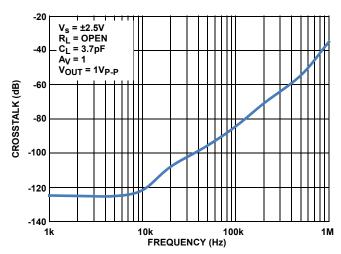


FIGURE 31. CROSSTALK vs FREQUENCY, V<sub>S</sub> = ±2.5V

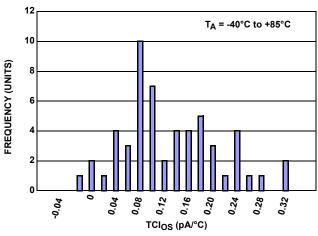


FIGURE 32. TCIOS HISTOGRAM

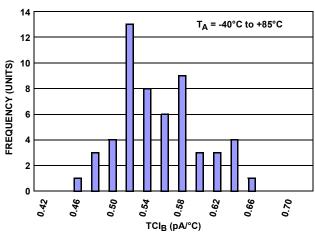


FIGURE 33. TCIB HISTOGRAM

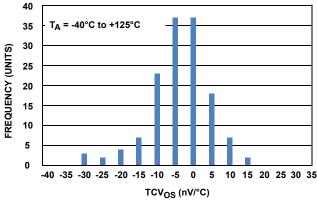


FIGURE 34.  $TCV_{OS}$  HISTOGRAM

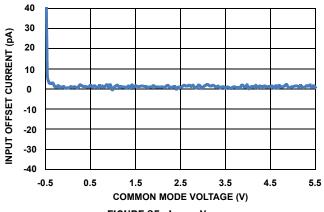
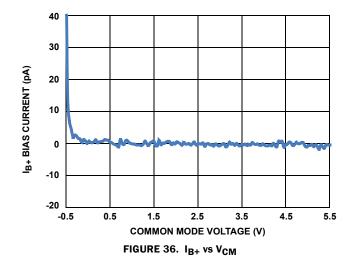
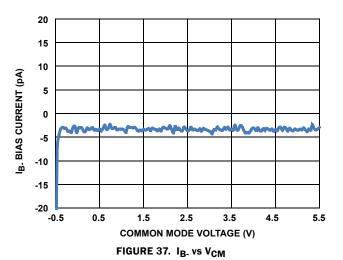
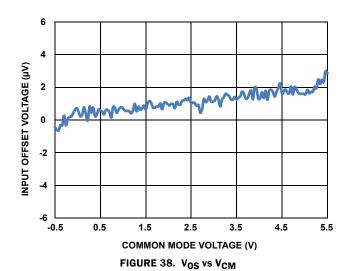


FIGURE 35. I<sub>OS</sub> vs V<sub>CM</sub>

## $\textbf{Typical Performance Curves} \ \ \text{V+} = 5 \text{V, V-} = 0 \text{V, V}_{\text{CM}} = 2.5 \text{V, R}_{\text{L}} = \text{Open, T} = +25 \, ^{\circ}\text{C, unless otherwise specified.} \ \ \textbf{(Continued)}$







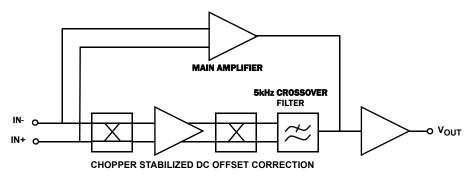


FIGURE 39. ISL28233, ISL28433 FUNCTIONAL BLOCK DIAGRAM

## **Applications Information**

#### **Functional Description**

The ISL28233 and ISL28433 use a proprietary chopper-stabilized technique (see Figure 39) that combines a 400kHz main amplifier with a very high open loop gain (174dB) chopper amplifier to achieve very low offset voltage and drift (2 $\mu$ V, 0.01 $\mu$ V/°C typical) while consuming only 18 $\mu$ A of supply current per channel.

This multi-path amplifier architecture contains a time continuous main amplifier whose input DC offset is corrected by a parallel-connected, high gain chopper stabilized DC correction amplifier operating at 100kHz. From DC to ~5kHz, both amplifiers are active with DC offset correction and most of the low frequency gain is provided by the chopper amplifier. A 5kHz crossover filter cuts off the low frequency amplifier path leaving the main amplifier active out to the 400kHz gain-bandwidth product of the device.

The key benefits of this architecture for precision applications are very high open loop gain, very low DC offset, and low 1/f noise. The noise is virtually flat across the frequency range from a few millihertz out to 100kHz, except for the narrow noise peak at the amplifier crossover frequency (5kHz).

#### Rail-to-rail Input and Output (RRIO)

The RRIO CMOS amplifier uses parallel input PMOS and NMOS that enable the inputs to swing 100mV beyond either supply rail. The inverting and non-inverting inputs do not have back-to-back input clamp diodes and are capable of maintaining high input impedance at high differential input voltages. This is effective in eliminating output distortion caused by high slew-rate input signals.

The output stage uses common source connected PMOS and NMOS devices to achieve rail-to-rail output drive capability with 17mA current limit and the capability to swing to within 20mV of either rail while driving a 10k $\Omega$  load.

#### **IN+ and IN- Protection**

All input terminals have internal ESD protection diodes to both positive and negative supply rails, limiting the input voltage to within one diode beyond the supply rails. For applications where either input is expected to exceed the rails by 0.5V, an external series resistor must be used to ensure the input currents never exceed 20mA (see Figure 40).

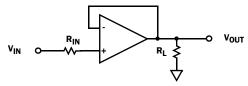


FIGURE 40. INPUT CURRENT LIMITING

#### **Layout Guidelines for High Impedance Inputs**

To achieve the maximum performance of the high input impedance and low offset voltage of the ISL28233 and ISL28433 amplifiers, care should be taken in the circuit board layout. The PC board surface must remain clean and free of moisture to avoid leakage currents between adjacent traces. Surface coating of the circuit board will reduce surface moisture and provide a humidity barrier, reducing parasitic resistance on the board.

#### **High Gain, Precision DC-Coupled Amplifier**

The circuit in Figure 41 implements a single-stage DC-coupled amplifier with an input DC sensitivity of under 100nV that is only possible using a low VOS amplifier with high open loop gain. High gain DC amplifiers operating from low voltage supplies are not practical using typical low offset precision op amps. For example, a typical precision amplifier in a gain of 10kV/V with a  $\pm 100\mu\text{V}$  Vos and offset drift  $0.5\mu\text{V/°C}$  of a low offset op amp would produce a DC error of >1V with an additional 5mV/°C of temperature dependent error making it difficult to resolve DC input voltage changes in the mV range.

The  $\pm 6\mu V$  max  $V_{OS}$  and  $0.05\mu V/^{\circ}C$  max temperature drift of the ISL28233, ISL28433 produces a temperature stable maximum DC output error of only  $\pm 60mV$  with a maximum output temperature drift of  $0.5mV/^{\circ}C$ . The additional benefit of a very low 1/f noise corner frequency and some feedback filtering enables DC voltages and voltage fluctuations well below 100mV to be easily detected with a simple single stage amplifier.

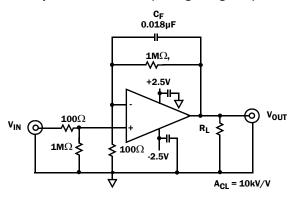


FIGURE 41. HIGH GAIN, PRECISION DC-COUPLED AMPLIFIER

#### ISL28233, ISL28433 SPICE Model

Figure 42 shows the SPICE model schematic and Figure 43 shows the net list for the ISL28233, ISL28433 SPICE model. The model is a simplified version of the actual device and simulates important parameters such as noise, Slew Rate, Gain and Phase. The model uses typical parameters from the "Electrical Specifications Table" on page 5. The poles and zeroes in the model were determined from the actual open and closed-loop gain and phase response. This enables the model to present an accurate AC representation of the actual device. The model is configured for ambient temperature of +25°C.

Figures 44 through 51 show the characterization vs simulation results for the Noise Density, Frequency Response vs Close Loop Gain, Gain vs Frequency vs  $C_L$  and Large Signal Step Response (4V).

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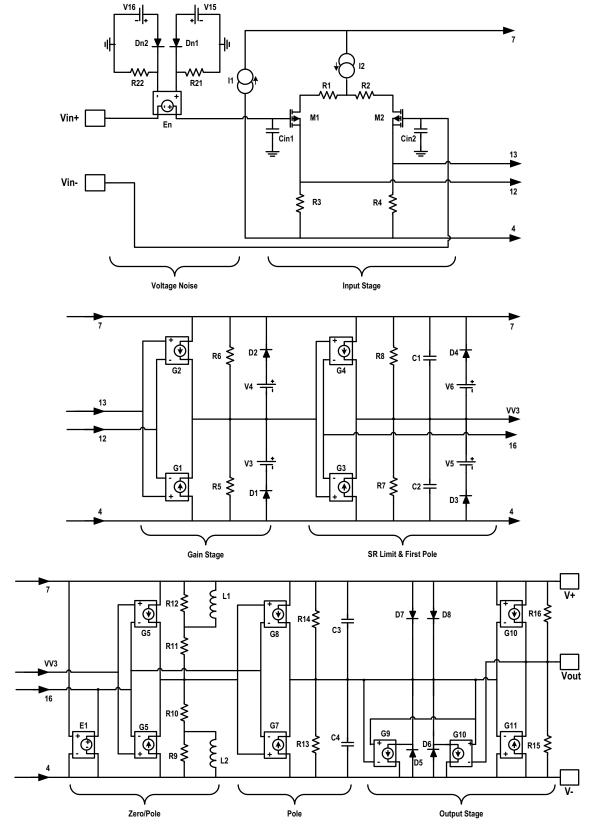


FIGURE 42. SPICE CIRCUIT SCHEMATIC

```
C C2
* ISL28233, ISL28433 Macromodel
                                                                       4 VV3 12u
* Revision B, April 2009
                                                             D D3
                                                                       4 17 DX
* AC characteristics, Voltage Noise
                                                             D D4
                                                                       18 7 DX
                                                             V V5
                                                                       VV3 17 0.7Vdc
*Copyright 2009 by Intersil Corporation
                                                             V_V6
                                                                       18 VV3 0.7Vdc
*Refer to data sheet "LICENSE STATEMENT" Use of
*this model indicates your acceptance with the
                                                             *Zero/Pole
*terms and provisions in the License Statement.
* Connections:
                      +input
                                                             E E1
                                                                       16 4 7 4 0.5
                                                             G G5
                                                                       4 VV4 VV3 16 0.000001
                           -input
                                +Vsupply
                                                             G G6
                                                                       7 VV4 VV3 16 0.000001
                                                                      20 7 0.3H
                                      -Vsupply
                                                             L L1
                                           output
                                                             R R12
                                                                       20 7 2.5meg
                                                             R R11
                                                                       VV4 20 1meg
                      3
                           2
                                7
                                           6
.subckt ISL28233
                                      4
                                                             L L2
                                                                      4 19 0.3H
                                                             R R9
                                                                       4 19 2.5meg
*Voltage Noise
                                                             R R10
                                                                       19 VV4 1meg
D DN1
            102 101 DN
                                                             *Pole
D DN2
            104 103 DN
                                                             G G7
                                                                       4 VV5 VV4 16 0.000001
R R21
                                                                       7 VV5 VV4 16 0.000001
           0 101 120k
                                                             G G8
R R22
           0 103 120k
                                                             C C3
                                                                       VV5 7 0.12p
E EN
           8 3 101 103 1
                                                             C C4
                                                                       4 VV5 0.12p
V V15
           102 0 0.1Vdc
                                                             R R13
                                                                       4 VV5 1meg
V_V16
           104 0 0.1Vdc
                                                                       VV57 1meg
                                                             R R14
*Input Stage
                                                             *Output Stage
C Cin1
            80 0.4p
                                                             G G9
                                                                        21 4 6 VV5 0.0000125
C_Cin2
                                                             G_G10
            20 2.0p
                                                                        22 4 VV5 6 0.0000125
R R1
          9 10 10
                                                             D D5
                                                                       4 21 DY
R<sub>R2</sub>
          10 11 10
                                                             D D6
                                                                       4 22 DY
R R3
          4 12 100
                                                             D D7
                                                                       7 21 DX
                                                             D_ D8
R R4
          4 13 100
                                                                       7 22 DX
M M1
          12899 pmosisil
                                                             R R15
                                                                        46 8k
+ L=50u
                                                             R R16
                                                                        67 8k
+ W=50u
                                                             G G11
                                                                        6 4 VV5 4 -0.000125
M M2
           13 2 11 11 pmosisil
                                                                        7 6 7 VV5 -0.000125
                                                             G G12
+ L=50u
+ W=50u
                                                             .model pmosisil pmos (kp=16e-3 vto=10m)
I_I1
        4 7 DC 92uA
                                                             .model DN D(KF=6.4E-16 AF=1)
1 12
        7 10 DC 100uA
                                                             .MODEL DX D(IS=1E-18 Rs=1)
                                                             .MODEL DY D(IS=1E-15 BV=50 Rs=1)
*Gain stage
                                                             .ends ISL28233
          4 VV2 13 12 0.0002
G G1
G G2
          7 VV2 13 12 0.0002
R R5
          4 VV2 1.3Meg
R R6
          VV2 7 1.3Meg
D D1
          4 14 DX
D<sub>D</sub>2
          15 7 DX
          VV2 14 0.7Vdc
V V3
V_V4
          15 VV2 0.7Vdc
*SR limit first pole
G G3
          4 VV3 VV2 16 1
G G4
          7 VV3 VV2 16 1
R R7
          4 VV3 1meg
R R8
          VV3 7 1meg
C C1
          VV3 7 12u
```

FIGURE 43. SPICE NET LIST

### **Characterization vs Simulation Results**

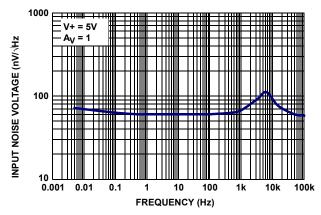


FIGURE 44. CHARACTERIZED INPUT NOISE VOLTAGE DENSITY vs FREQUENCY

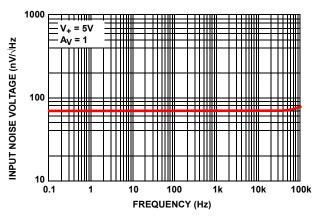


FIGURE 45. SIMULATED INPUT NOISE VOLTAGE DENSITY vs FREQUENCY

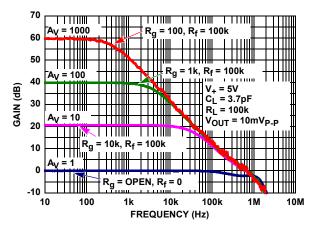


FIGURE 46. CHARACTERIZED FREQUENCY RESPONSE vs CLOSED LOOP GAIN

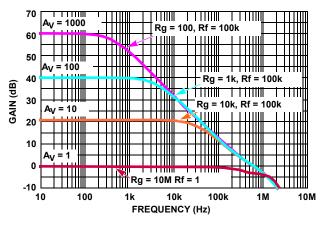


FIGURE 47. SIMULATED FREQUENCY RESPONSE vs CLOSED LOOP GAIN

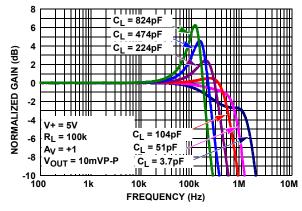


FIGURE 48. CHARACTERIZED GAIN vs FREQUENCY vs CL

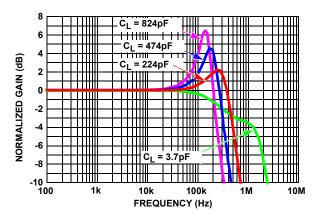


FIGURE 49. SIMULATED GAIN vs FREQUENCY vs CL

## **Characterization vs Simulation Results (Continued)**

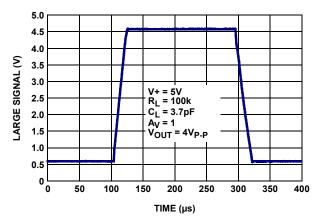


FIGURE 50. CHARACTERIZED LARGE SIGNAL STEP RESPONSE (4V)

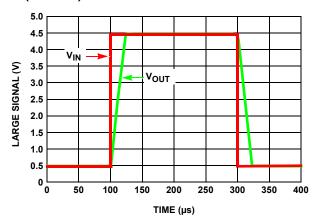


FIGURE 51. SIMULATED LARGE SIGNAL STEP RESPONSE (4V)

## **Revision History**

The revision history provided is for informational purposes only and is believed to be accurate, but not warranted. Please go to web to make sure you have the latest Rev.

DATE	REVISION	CHANGE
5/31/11	FN7692.3	Changed minimum operating supply voltage from +1.65V to +1.8V throughout entire datasheet.
3/24/11		Added to Ordering Information Table on page 2 - ISL28233SOICEVAL1Z, ISL28433TSSOPEVAL1Z, ISL28433SOICEVAL1Z
12/2/10	FN7692.2	Added "Related Literature" on page 1
		Removed "Coming Soon" from ISL28233FRZ device (8 Ld DFN) in "Ordering Information" on page 2.
		Corrected Thermal Pad Pin Name in "Pin Descriptions" on page 4 from "NC" to "PAD"
		Corrected $\theta_{JA}$ note for TDFN package in "Thermal Information" on page 5 from " $\theta_{JA}$ is measured with the component mounted on a high effective thermal conductivity test board in free air. See Tech Brief TB379 for details." To " $\theta_{JA}$ is measured in free air with the component mounted on a high effective thermal conductivity test board with "direct attach" features. See Tech Brief TB379." (since TDFN has thermal pad; TDFN package option not released yet)
10/27/10	FN7692.1	Changed Part Marking for ISL28233FUZ from 8233Z to 233FZ in "Ordering Information" table on page 2 Added ISL28233 in DFN package to Ordering Information" table on page 2. On page 6, removed Note 8. Changed note in MIN MAX columns of "Electrical Specifications" table from: "Parameters with MIN and/or MAX limits are 100% tested at +25°C, unless otherwise specified. Temperature limits established by characterization and are not production tested." To: "Compliance to datasheet limits is assured by one or more methods: production test, characterization and/or
8/25/10	FN7692.0	design."  Initial Release.

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\*For a complete listing of Applications, Related Documentation and Related Parts, please see the respective device information page on intersil.com: <a href="ISL28233">ISL28433</a></a>

To report errors or suggestions for this datasheet, please go to www.intersil.com/askourstaff

FITs are available from our website at <a href="http://rel.intersil.com/reports/search.php">http://rel.intersil.com/reports/search.php</a>

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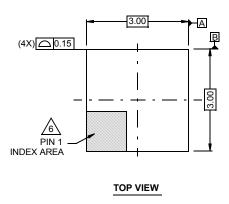
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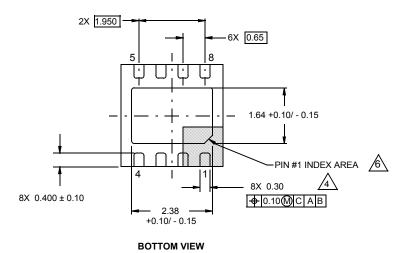
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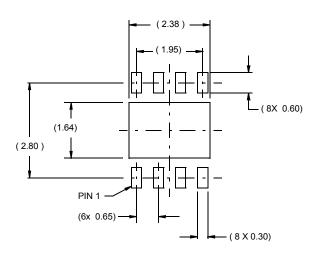


# Package Outline Drawing

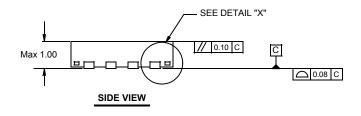
8 LEAD DUAL FLAT NO-LEAD PLASTIC PACKAGE Rev 0 9/09

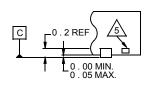






TYPICAL RECOMMENDED LAND PATTERN



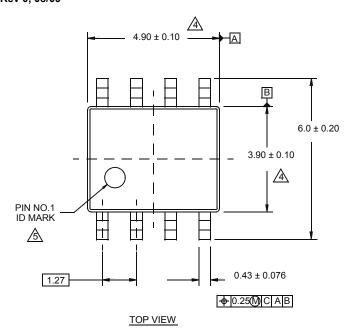


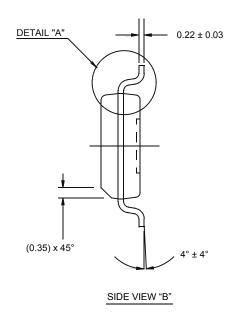
DETAIL "X"

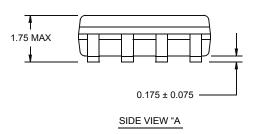
- Dimensions are in millimeters.
   Dimensions in ( ) for Reference Only.
- 2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
- 3. Unless otherwise specified, tolerance : Decimal  $\pm 0.05$
- Dimension applies to the metallized terminal and is measured between 0.15mm and 0.30mm from the terminal tip.
- 5. Tiebar shown (if present) is a non-functional feature.
- The configuration of the pin #1 identifier is optional, but must be located within the zone indicated. The pin #1 identifier may be either a mold or mark feature.

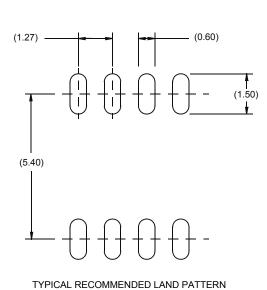
## **Package Outline Drawing**

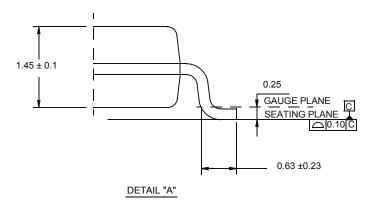
M8.15E **8 LEAD NARROW BODY SMALL OUTLINE PLASTIC PACKAGE** Rev 0, 08/09









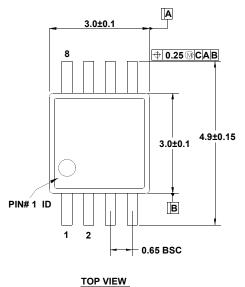


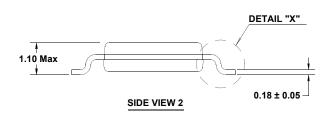
- 1. Dimensions are in millimeters. Dimensions in ( ) for Reference Only.
- 2. Dimensioning and tolerancing conform to AMSE Y14.5m-1994.
- Unless otherwise specified, tolerance: Decimal ± 0.05
- Dimension does not include interlead flash or protrusions. Interlead flash or protrusions shall not exceed 0.25mm per side.
- 5. The pin #1 identifier may be either a mold or mark feature.
- Reference to JEDEC MS-012.

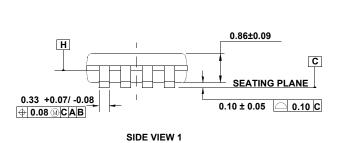
## **Package Outline Drawing**

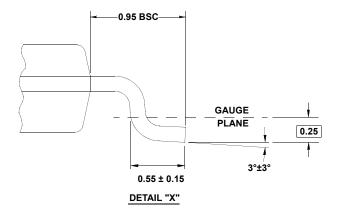
#### M8.118A

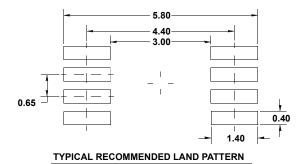
8 LEAD MINI SMALL OUTLINE PLASTIC PACKAGE (MSOP) Rev 0, 9/09





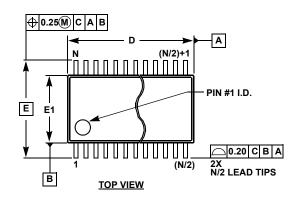


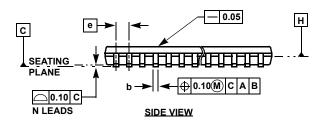


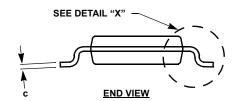


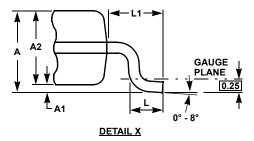
- 1. Dimensions are in millimeters.
- 2. Dimensioning and tolerancing conform to JEDEC MO-187-AA and AMSE Y14.5m-1994.
- Plastic or metal protrusions of 0.15mm max per side are not included.
- Plastic interlead protrusions of 0.25mm max per side are not included.
- 5. Dimensions "D" and "E1" are measured at Datum Plane "H".
- 6. This replaces existing drawing # MDP0043 MSOP 8L.

## Thin Shrink Small Outline Package Family (TSSOP)









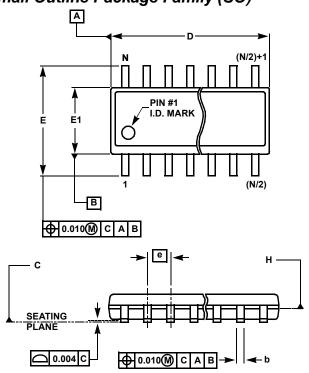
# MDP0044 THIN SHRINK SMALL OUTLINE PACKAGE FAMILY

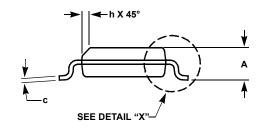
		MIL				
SYMBOL	14 LD	16 LD	20 LD	24 LD	28 LD	TOLERANCE
Α	1.20	1.20	1.20	1.20	1.20	Max
A1	0.10	0.10	0.10	0.10	0.10	±0.05
A2	0.90	0.90	0.90	0.90	0.90	±0.05
b	0.25	0.25	0.25	0.25	0.25	+0.05/-0.06
С	0.15	0.15	0.15	0.15	0.15	+0.05/-0.06
D	5.00	5.00	6.50	7.80	9.70	±0.10
Е	6.40	6.40	6.40	6.40	6.40	Basic
E1	4.40	4.40	4.40	4.40	4.40	±0.10
е	0.65	0.65	0.65	0.65	0.65	Basic
L	0.60	0.60	0.60	0.60	0.60	±0.15
L1	1.00	1.00	1.00	1.00	1.00	Reference

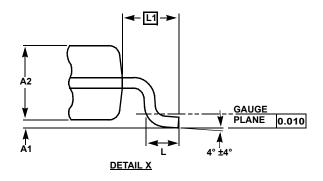
Rev. F 2/07

- Dimension "D" does not include mold flash, protrusions or gate burrs. Mold flash, protrusions or gate burrs shall not exceed 0.15mm per side.
- Dimension "E1" does not include interlead flash or protrusions. Interlead flash and protrusions shall not exceed 0.25mm per side.
- 3. Dimensions "D" and "E1" are measured at dAtum Plane H.
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994.

## Small Outline Package Family (SO)







#### **MDP0027**

#### **SMALL OUTLINE PACKAGE FAMILY (SO)**

	INCHES								
SYMBOL	SO-8	SO-14	SO16 (0.150")	SO16 (0.300") (SOL-16)	SO20 (SOL-20)	SO24 (SOL-24)	SO28 (SOL-28)	TOLERANCE	NOTES
Α	0.068	0.068	0.068	0.104	0.104	0.104	0.104	MAX	-
A1	0.006	0.006	0.006	0.007	0.007	0.007	0.007	±0.003	-
A2	0.057	0.057	0.057	0.092	0.092	0.092	0.092	±0.002	-
b	0.017	0.017	0.017	0.017	0.017	0.017	0.017	±0.003	-
С	0.009	0.009	0.009	0.011	0.011	0.011	0.011	±0.001	-
D	0.193	0.341	0.390	0.406	0.504	0.606	0.704	±0.004	1, 3
Е	0.236	0.236	0.236	0.406	0.406	0.406	0.406	±0.008	-
E1	0.154	0.154	0.154	0.295	0.295	0.295	0.295	±0.004	2, 3
е	0.050	0.050	0.050	0.050	0.050	0.050	0.050	Basic	-
L	0.025	0.025	0.025	0.030	0.030	0.030	0.030	±0.009	-
L1	0.041	0.041	0.041	0.056	0.056	0.056	0.056	Basic	-
h	0.013	0.013	0.013	0.020	0.020	0.020	0.020	Reference	-
N	8	14	16	16	20	24	28	Reference	-

NOTES:

Rev. M 2/07

- 1. Plastic or metal protrusions of 0.006" maximum per side are not included.
- 2. Plastic interlead protrusions of 0.010" maximum per side are not included.
- 3. Dimensions "D" and "E1" are measured at Datum Plane "H".
- 4. Dimensioning and tolerancing per ASME Y14.5M-1994

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