19-1759; Rev 1; 8/03

EVALUATION KIT AVAILABLE

# **MAXIMI** 3.4GHz to 3.8GHz SiGe Low-Noise Amplifier/PA Predriver

## **General Description**

The MAX2645 is a versatile, high-linearity, low-noise amplifier designed for 3.4GHz to 3.8GHz wireless local loop (WLL), wireless broadband access, and digital microwave radio applications. The device features an externally adjustable bias control, set with a single resistor, that allows the user to meet minimum linearity requirements while minimizing current consumption. The amplifier's high-gain, low-noise performance and adjustable input third-order intercept (IP3) allow it to be used as a low-noise amplifier (LNA) in the receive path, a PA predriver in the transmit path, or as an LO buffer.

The MAX2645 features a logic-level gain control that provides a 25dB step reduction in gain, which improves IP3 performance for operation during high input signal level conditions. Supply current is reduced from 9mA in high-gain mode to 3mA in low-gain mode. The device also includes a logic-controlled shutdown mode, which reduces supply current to 0.1 $\mu$ A. The MAX2645 operates from a +3V to +5.5V supply and is offered in the miniature 10-pin  $\mu$ MAX package (5mm × 3mm) with an exposed paddle. Its performance has been optimized for use with the MAX2683/MAX2684 3.5GHz SiGe mixers to provide a complete high-performance, front-end solution for 3.5GHz applications.

### Applications

Wireless Local Loop Wireless Broadband Access Digital Microwave Radios

### Features

- 3.4GHz to 3.8GHz Frequency Range
- LNA Performance (High/Low-Gain Modes) Gain: +14.4dB/-9.7dB NF: 2.3dB/15.5dB Input IP3: +4dBm/+13dBm Supply Current: 9.2mA/2.7mA
- Highly Versatile Application Receive Path 1st and 2nd Stage LNA Transmit PA Predriver LO Buffer
- ♦ Adjustable IP3 and Supply Current
- 0.1µA Supply Current in Shutdown Mode
- ♦ +3.0V to +5.5V Single-Supply Operation
- 10-Pin µMAX-EP Package (5.0mm x 3.0mm)

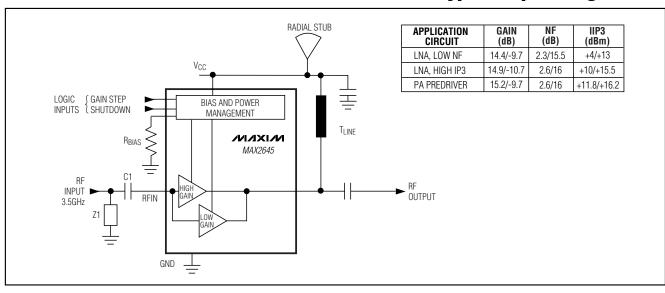
## \_Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	
MAX2645EUB	-40°C to +85°C	10 µMAX-EP*	

\*EP = exposed paddle.

Pin Configuration appears at end of data sheet.

# Typical Operating Circuit



## 

\_ Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

### **ABSOLUTE MAXIMUM RATINGS**

0.3V to +6.0V
0.3V to (Vcc + 0.3V)
16dBm
10kΩ
C)
825mW

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

**CAUTION!** ESD SENSITIVE DEVICE

## DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +3.0V \text{ to } +5.5V, \text{ GAIN} = \overline{SHDN} = V_{CC}, R_{BIAS} = 20k\Omega$ , no RF signals applied,  $T_A = -40^{\circ}C$  to  $+85^{\circ}C$ . Typical values are at V\_{CC} = +3.3V,  $T_A = +25^{\circ}C$ , unless otherwise indicated.) (Note 1)

PARAMETER	CON	MIN	TYP	MAX	UNITS		
Supply Voltage			3.0		5.5	V	
	$R_{BIAS} = 20k\Omega$ ,	$GAIN = V_{CC}$		9.2	10.9		
	$T_A = +25^{\circ}C$	GAIN = GND		2.7	3.9	mA	
Operating Supply Current	$\begin{aligned} R_{B AS} &= 20k\Omega, \\ T_A &= -40^\circC \text{ to } +85^\circC \end{aligned}$	$GAIN = V_{CC}$			11.6		
Operating Supply Current		GAIN = GND			4.0		
	$R_{BIAS} = 15k\Omega$ ,	$GAIN = V_{CC}$		12			
	$T_A = +25^{\circ}C$	GAIN = GND		3.6			
Shutdown Supply Current	SHDN = GND	·		0.1	2	μΑ	
Input Logic Voltage High	GAIN, SHDN		2.0			V	
Input Logic Voltage Low	GAIN, SHDN				0.6	V	
Input Logic Bias Current	$GAIN = \overline{SHDN} = V_{CC}$				1		
	$GAIN = \overline{SHDN} = GND$		-10			μΑ	

## AC ELECTRICAL CHARACTERISTICS—LNA (Low-Noise Figure Application Circuit)

(MAX2645 EV kit, V<sub>CC</sub> = GAIN =  $\overline{SHDN}$  = +3.3V,R<sub>BIAS</sub> = 20k $\Omega$  ±1%, P<sub>RFIN</sub> = -20dBm, f<sub>RFIN</sub> = 3550MHz, Z<sub>0</sub> = 50 $\Omega$ , T<sub>A</sub> = +25°C, unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS	
Frequency Range	(Note 2)	3400		3800	MHz	
Coin (Nate 2)	$GAIN = V_{CC}$	12.9	14.4	15.4	dD	
Gain (Note 3)	GAIN = GND	-11.8	-9.7	-8.0	dB	
Gain Variation over Temperature	$T_A = -40^{\circ}C$ to $+85^{\circ}C$ , GAIN = V <sub>CC</sub> or GND (Note 4)		±0.3	±0.7	dB	
Gain Step			±24.1		dB	
Jacob Thind Orden Internet	GAIN = V <sub>CC</sub> (Note 5)	= V <sub>CC</sub> (Note 5) +4			alDuas	
Input Third-Order Intercept	GAIN = GND (Note 6)		+13		dBm	
Jacout 1dD Compression Daint	$GAIN = V_{CC}$	-5			alDura	
Input 1dB Compression Point	GAIN = GND		0		dBm	
	GAIN = V <sub>CC</sub> (Notes 4, 7)		2.3 3.0		-10	
Noise Figure	GAIN = GND		15.5		dB	
Deveres lesister	GAIN = V <sub>CC</sub>		25		dD	
Reverse Isolation	GAIN = GND		19		dB	
Gain Step Transition Time	(Note 8)		1		μs	
Turn-On/Turn-Off Time	(Note 9)		0.5		μs	

## AC ELECTRICAL SPECIFICATIONS—LNA (High-Input IP3 Application Circuit)

(MAX2645 EV kit, V<sub>CC</sub> = GAIN =  $\overline{SHDN}$  = +3.3V,R<sub>BIAS</sub> = 20k $\Omega$  ±1%, P<sub>RFIN</sub> = -20dBm, f<sub>RFIN</sub> = 3550MHz, Z<sub>0</sub> = 50 $\Omega$ , T<sub>A</sub> = +25°C, unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS	
Frequency Range	(Note 2)	3400		3800	MHz	
	$GAIN = V_{CC}$		14.9		dB	
Gain	GAIN = GND		-10.7			
Gain Variation over Temperature	$T_A = -40^{\circ}C$ to +85°C, GAIN = V <sub>CC</sub> or GND		±0.3		dB	
Gain Step			25.6		dB	
Input Third-Order Intercept	GAIN = V <sub>CC</sub> (Note 6)		+10.0		dBm	
input mird-Order intercept	GAIN = GND (Note 7)		+15.5			
Input 1dB Compression Point	$GAIN = V_{CC}$		-4		dBm	
Input Tub Compression Form	GAIN = GND	0			UDIT	
Noigo Figuro	$GAIN = V_{CC}$	2.6			dD	
Noise Figure	GAIN = GND		16		dB	
Reverse Isolation	GAIN = V <sub>CC</sub> 25		25		dB	
	GAIN = GND		19		uВ	

### AC ELECTRICAL SPECIFICATIONS—PA Predriver Application Circuit

 $(MAX2645 \text{ EV kit}, V_{CC} = GAIN = \overline{SHDN} = +3.3V, R_{BIAS} = 20k\Omega \pm 1\%, P_{REIN} = -20dBm, f_{REIN} = 3550MHz, Z_{0} = 50\Omega$  $T_A = +25^{\circ}C$ , unless otherwise noted.)

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
Frequency Range	(Note 2)	3400		3800	MHz
Cain	$GAIN = V_{CC}$		15.2		alD
Gain	GAIN = GND		-9.7		dB
Gain Variation over Temperature	$T_A = -40^{\circ}C$ to +85°C, GAIN = V <sub>CC</sub> or GND		±0.3		dB
Gain Step			24.9		dB
Insuit Third Order Intereest	$GAIN = V_{CC}$ (Note 6)	+11.8			dBm
Input Third-Order Intercept	GAIN = GND (Note 7)		+16.2		
Input 1dB Compression Point	GAIN = V <sub>CC</sub> -1.8			dBm	
Input Tub Compression Foint	GAIN = GND	GAIN = GND 0			uBm
Noise Figure	$GAIN = V_{CC}$ 2.6			dB	
Noise Figure	GAIN = GND 16				
Reverse Isolation	$GAIN = V_{CC}$		25		dB
	GAIN = GND		19		uр

Note 1: Limits over temperature guaranteed by correlation to worst-case temperature testing.

Note 2: This is the recommended operating frequency range. Operation outside this frequency range is possible but has not been characterized. The device is characterized and tested at 3550MHz. For optimum performance at a given frequency, the output matching network must be properly designed. See Applications Information section.

Note 3: Specifications are corrected for board losses (0.25dB at input, 0.25dB at output).

Note 4: Guaranteed by design and characterization.

**Note 5:** Input IP3 measured with two tones,  $f_1 = 3550$ MHz and  $f_2 = 3551$ MHz, at -20dBm per tone.

**Note 6:** Input IP3 measured with two tones,  $f_1 = 3550$ MHz and  $f_2 = 3551$ MHz, at -12dBm per tone.

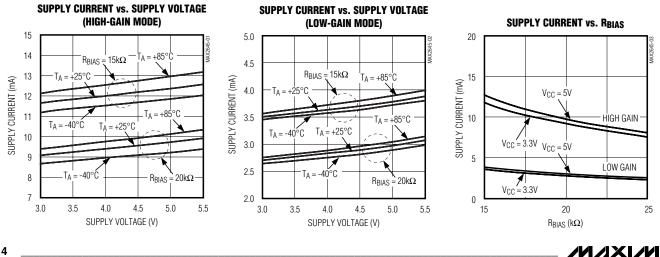
Note 7: Specifications are corrected for board losses (0.25dB at input).

Note 8: Time from when GAIN changes state to when output power reaches 1dB of its final value.

Note 9: Time from when SHDN changes state to when output power reaches 1dB of its final value.

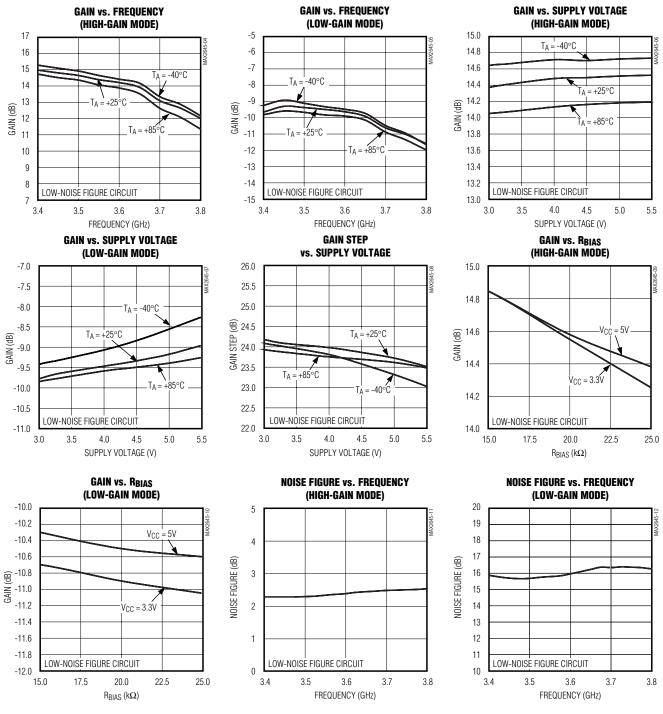
## **Typical Operating Characteristics**

(MAX2645 EV kit, V<sub>CC</sub> = +3.3V, R<sub>BIAS</sub> =  $20k\Omega$ , f<sub>RFIN</sub> = 3550MHz, T<sub>A</sub> = + $25^{\circ}$ C, unless otherwise noted.)



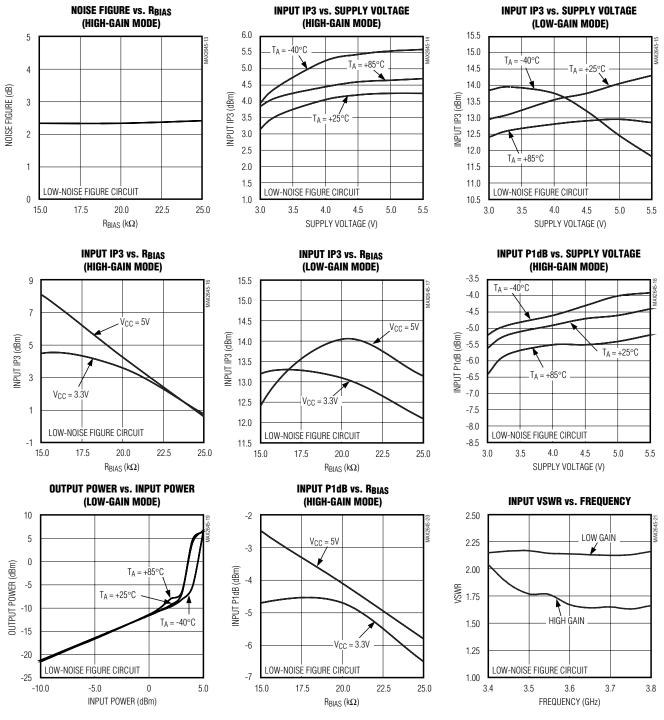
### **Typical Operating Characteristics (continued)**

(MAX2645 EV kit, V<sub>CC</sub> = +3.3V, R<sub>BIAS</sub> =  $20k\Omega$ , f<sub>RFIN</sub> = 3550MHz, T<sub>A</sub> = + $25^{\circ}C$ , unless otherwise noted.)



## **Typical Operating Characteristics (continued)**

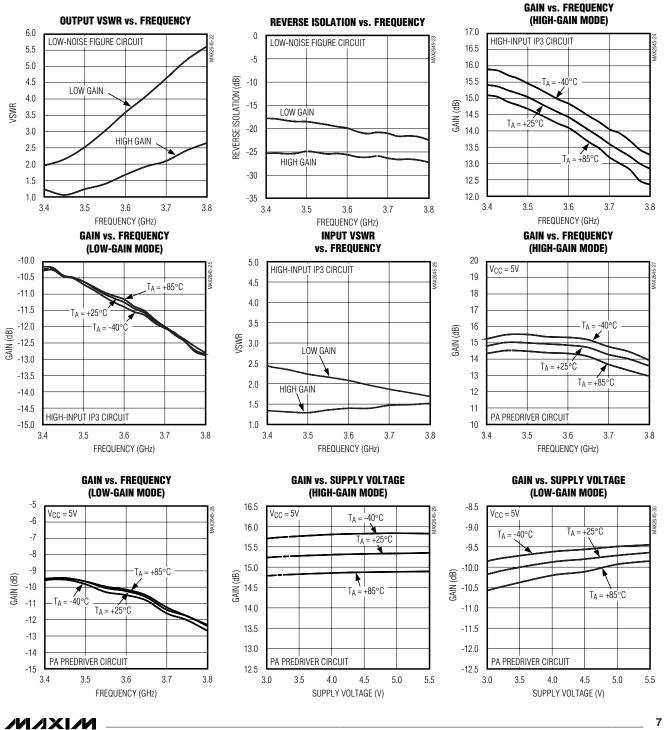
(MAX2645 EV kit, V<sub>CC</sub> = +3.3V, R<sub>BIAS</sub> =  $20k\Omega$ , f<sub>RFIN</sub> = 3550MHz, T<sub>A</sub> = + $25^{\circ}$ C, unless otherwise noted.)



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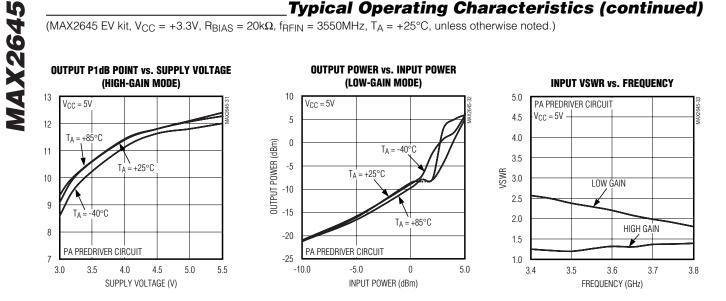
## **Typical Operating Characteristics (continued)**

(MAX2645 EV kit, V<sub>CC</sub> = +3.3V, R<sub>BIAS</sub> =  $20k\Omega$ , f<sub>RFIN</sub> = 3550MHz, T<sub>A</sub> = + $25^{\circ}C$ , unless otherwise noted.)



## **Typical Operating Characteristics (continued)**

(MAX2645 EV kit, V<sub>CC</sub> = +3.3V, R<sub>BIAS</sub> =  $20k\Omega$ , f<sub>RFIN</sub> = 3550MHz, T<sub>A</sub> = + $25^{\circ}$ C, unless otherwise noted.)



## **Pin Description**

PIN	NAME	FUNCTION
1, 2, 4, 7, EP	GND	Ground. Connect to ground plane with a low-inductance connection. Solder exposed paddle evenly to the board ground plane.
3	RFIN	RF Input Port to Amplifier. Requires a matching network and a DC-blocking capacitor that may be part of this network. See Figure 1 for recommended component values.
5	BIAS	Bias-Setting Resistor Connection. A resistor, R <sub>BIAS</sub> , placed from BIAS to ground sets the linearity and supply current of the amplifier.
6	RFOUT	RF Open-Collector Output Port of Amplifier. Requires a matching network composed of an inductance to $V_{CC}$ and a DC-blocking capacitor. See Figure 1 for recommended component values.
8	SHDN	Shutdown Control Logic-Level Input. A logic high enables the device for normal operation. A logic low places the device in low-power shutdown mode.
9	GAIN	Gain Control Logic-Level Input. A logic high places the device in high-gain mode. A logic low places the device in low-gain mode, reducing the gain by 25dB.
10	V <sub>CC</sub>	Power Supply Input. Bypass directly to ground with a capacitor as close to the supply pin as possible. See Figure 1 for recommended component values.

## **Detailed Description**

The MAX2645 is a versatile amplifier with high-gain, high-linearity, and low-noise performance—features that make it suitable for use as an LNA, high-linearity/low-noise amplifier, PA predriver, or LO buffer in the 3.4GHz to 3.8GHz frequency range. See Figure 1, MAX2645 *Typical Application Circuit*, for recommended component values. A single external bias-setting resistor allows the system designer to trade off linearity for reduced supply current. A logic-level control reduces gain by a 25dB step to further improve input IP3 performance. A low-power shutdown mode disables the device and reduces current consumption to 0.1µA.

#### **Bias Circuitry**

The linearity and supply current of the MAX2645 are externally programmable with a single resistor (R<sub>BIAS</sub>) placed from BIAS to GND. Larger resistor values result in lower IP3 performance and lower supply current, while smaller resistor values result in higher IP3 performance and higher supply current. Use resistor values in the 15k $\Omega$  to 25k $\Omega$  range, with a nominal value of 20k $\Omega$  suitable for most applications. See *Typical Operating Characteristics* for performance variation vs. R<sub>BIAS</sub> value.

#### Table 1. MAX2645 S-Parameters

#### **Gain Step Control**

The MAX2645 features a logic-level gain step control input (GAIN) that places the device in high-gain or low-gain mode. A logic-level high places the device in high-gain mode, where the gain is 14.5dB. A logic-level low places the device in low-gain/high-linearity mode, where the gain is reduced to 10dB and the input IP3 performance is increased.

#### **Shutdown Control**

The MAX2645 features a logic-level shutdown control input. A logic high on SHDN enables the device for normal operation. A logic low on SHDN disables all device functions and reduces supply current to 0.1µA.

### Applications Information

#### **RF** Input

The RFIN port is internally biased and requires an external DC-blocking capacitor. A matching network is required for best performance. Figure 1 shows component values optimized for best noise-figure performance, low-noise figure, high-input IP3 performance, and highoutput P1dB performance in the 3.4GHz to 3.8GHz frequency range. For matching to other frequencies, see Tables 1 and 2.

FREQ	S	S11		21	S	12	S22			
(MHz)	MAG	PHASE	MAG	PHASE	MAG	PHASE	MAG	PHASE		
	$R_{BIAS} = 20k\Omega$ , $V_{CC} = +3.3V$ , $T_A = +25^{\circ}C$									
3400	0.468	-149.8	5.061	-44.6	0.053	-55.5	0.660	-57.0		
3450	0.466	-150.4	4.975	-46.3	0.058	-60.8	0.658	-58.4		
3500	0.472	-151.6	5.098	-49.9	0.056	-64.6	0.661	-60.6		
3550	0.469	-153.4	4.883	-53.7	0.054	-62.7	0.658	-63.0		
3600	0.471	-154.6	4.814	-53.7	0.056	-64.4	0.647	-64.2		
3650	0.477	-155.0	5.118	-57.4	0.058	-68.9	0.657	-66.2		
3700	0.485	-156.6	4.769	-63.4	0.054	-70.5	0.657	-69.8		
3750	0.484	-156.5	4.780	-62.3	0.058	-72.0	0.654	-70.9		
3800	0.492	-157.0	4.939	-66.6	0.060	-75.4	0.654	-72.3		
			RBIAS = 15k	2, Vcc = +5V, 1	Γ <sub>A</sub> = +25°C					
3400	0.454	-146.6	5.350	-41.8	0.057	-51.3	0.651	-52.3		
3450	0.457	-147.4	5.245	-43.5	0.061	-56.7	0.646	-53.7		
3500	0.465	-147.9	5.375	-46.6	0.060	-61.2	0.654	-55.6		
3550	0.468	-149.7	5.165	-50.3	0.057	-61.0	0.652	-58.3		
3600	0.472	-150.5	5.066	-50.2	0.060	-62.7	0.645	-59.3		
3650	0.481	-150.5	5.386	-53.4	0.063	-67.6	0.652	-60.7		
3700	0.486	-152.2	5.040	-59.4	0.060	-67.8	0.648	-63.9		
3750	0.486	-152.4	5.019	-58.3	0.062	-67.0	0.642	-64.8		
3800	0.499	-152.6	5.207	-62.0	0.065	-73.3	0.643	-66.2		

FREQUENCY (MHz)	FMIN (dB)	Γ <b>opt</b>	Γopt ANGLE	<b>R<sub>N</sub> (</b> Ω)
	RBIAS	= 20kΩ, V <sub>CC</sub> = +3.3V, T <sub>A</sub>	= +25°C	
3400	2.098	0.237	144.1	31.1
3450	2.122	0.235	146.1	31.5
3500	2.148	0.235	148.2	32.0
3550	2.173	0.234	150.3	32.5
3600	2.198	0.233	152.4	32.9
3650	2.225	0.232	154.5	33.5
3700	2.251	0.231	156.5	33.9
3750	2.279	0.230	158.6	34.5
3800	2.306	0.229	160.7	35.0
	R <sub>BIAS</sub>	; = 15kΩ, V <sub>CC</sub> = +5V, T <sub>A</sub> =	= +25°C	
3400	2.103	0.210	146.3	31.1
3450	2.127	0.209	148.4	31.6
3500	2.152	0.208	150.5	32.1
3550	2.177	0.207	152.6	32.5
3600	2.203	0.206	154.7	33.0
3650	2.229	0.206	156.8	33.5
3700	2.256	0.205	158.9	34.0
3750	2.282	0.204	161.0	34.6
3800	2.310	0.204	163.1	35.1

### Table 2. MAX2645 Noise Parameters

#### **RF Output**

The RFOUT port is an open-collector output that must be tied to Vcc through an inductance for proper biasing. The MAX2645 EV kit uses a length of transmission line equivalent to 1.5nH of inductance. A DC-blocking capacitor is required and can be part of the output matching network. See Figure 1 for component values recommended for operation over the 3.4GHz to 3.8GHz frequency range. See Table 1 for matching to other frequencies. This transmission line is terminated at the Vcc node with a radial stub for high-frequency bypassing. This arrangement provides a high-Q, low-loss bias network used to optimize performance. The radial stub can be replaced with an appropriate microwave capacitor.

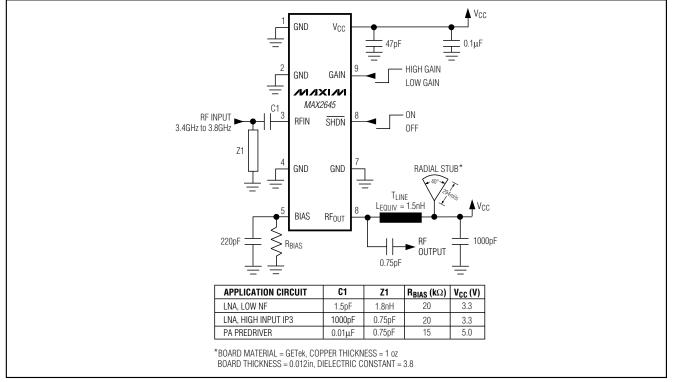
#### Power-Supply, Bias Circuitry, and Logic-Input Bypassing

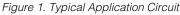
Proper power-supply bypassing is essential for high-frequency circuit stability. Bypass V<sub>CC</sub> with  $10\mu$ F,  $0.1\mu$ F, and 50pF capacitors located as close to the V<sub>CC</sub> pin as possible.

To minimize the amount of noise injected into the bias circuitry and logic inputs, bypass the pins with capacitors located as near to the device pin as possible. For additional isolation on the logic-control pins, place resistors between the logic-control inputs and the bypass capacitors. See Figure 1 for recommended component values; refer to MAX2645 EV kit manual for recommended board layout.

#### **Layout Considerations**

A properly designed PC board is an essential part of any RF/microwave circuit. Keep RF signal lines as short as possible to reduce losses, radiation, and inductance. Use separate, low-inductance vias to the ground plane for each ground pin. For best performance, solder the exposed paddle on the bottom of the device package evenly to the board ground plane.





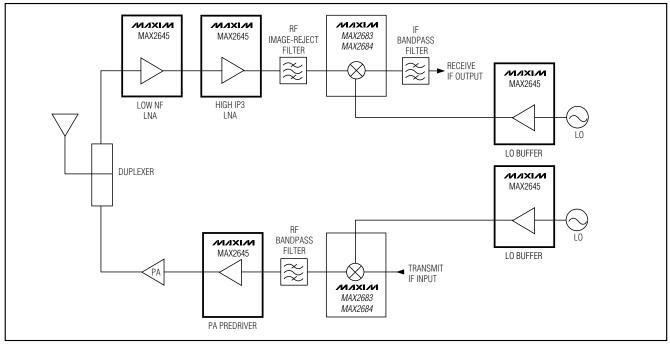
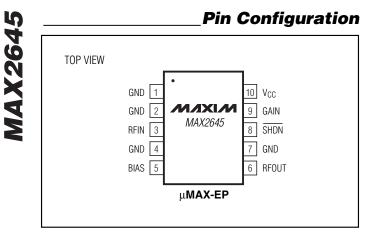


Figure 2. Typical System Application Block Diagram

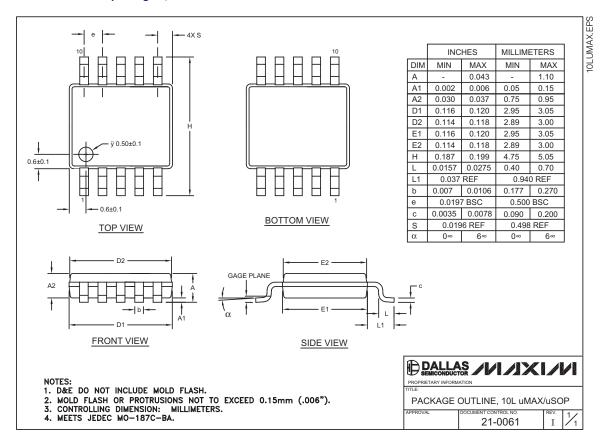


**Chip Information** 

TRANSISTOR COUNT: 271

## **Package Information**

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <u>www.maxim-ic.com/packages</u>.)



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