

LTC1928-5

Doubler Charge Pump with Low Noise Linear Regulator in ThinSOT

### FEATURES

- Low Output Noise: 90µV<sub>RMS</sub> (100kHz BW)
- Fixed Output Voltage: 5V
- Input Voltage Range: 2.7V to 4.4V
- No Inductors Required
- Uses Small Ceramic Capacitors
- Output Current Up to 30mA
- 550kHz Switching Frequency
- Low Operating Current: 190µA
- Low Shutdown Current: 4µA
- Internal Thermal Shutdown and Current Limiting
- Low Profile (1mm) ThinSOT<sup>TM</sup> Package

### **APPLICATIONS**

- VCO Power Supplies for Cellular Phones
- 2-Way Pagers
- Wireless PCMCIA Cards
- Portable Medical Instruments
- Low Power Data Acquisition
- Remote Transmitters
- White LED Drivers
- GaAs Switches

## TYPICAL APPLICATION

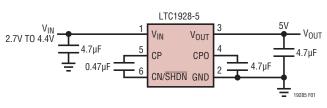


Figure 1. Low Noise 5V Power Supply

### DESCRIPTION

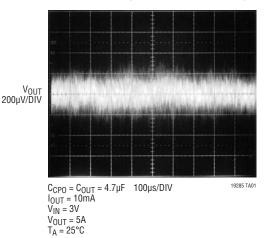
The LTC®1928-5 is a doubler charge pump with an internal low noise, low dropout (LDO) linear regulator. The part is designed to provide a low noise boosted supply voltage for powering noise sensitive devices such as high frequency VCOs in wireless applications.

An internal charge pump converts a 2.7V to 4.4V input to a boosted output, while the internal LDO regulator converts the boosted voltage to a low noise regulated output. The regulator is capable of supplying up to 30mA of output current. Shutdown reduces the supply current to <8 $\mu$ A, removes the load from V<sub>IN</sub> by disabling the regulator and discharges V<sub>OUT</sub> to ground through a 200 $\Omega$  switch.

The LTC1928-5 LDO regulator is stable with only  $2\mu$ F on the output. Small ceramic capacitors can be used, reducing PC board area.

The LTC1928-5 is short-circuit and overtemperature protected. The part is available in a 6-pin low profile (1mm) ThinSOT package.

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#### Output Noise (BW = 10Hz to 2.5MHz)

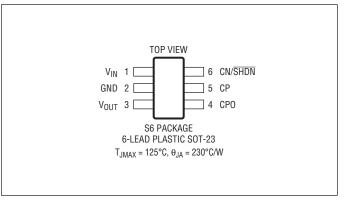
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### **ABSOLUTE MAXIMUM RATINGS**

(Note 1)

V <sub>IN</sub> to Ground
CPO to Ground 10V
$CN/SHDN$ to Ground0.3V to $(V_{IN} + 0.3V)$
V <sub>OUT</sub> Short-Circuit Duration Indefinite
I <sub>OUT</sub> 40mA
Operating Temperature Range (Note 2)–40°C to 85°C
Maximum Junction Temperature 125°C
Storage Temperature Range –65°C to 150°C
Lead Temperature (Soldering, 10 sec)

## PIN CONFIGURATION



## **ORDER INFORMATION**

LEAD FREE FINISH	TAPE AND REEL	PART MARKING	RT MARKING PACKAGE DESCRIPTION TEMPERATURI	
LTC1928ES6-5#PBF	LTC1928ES6-5#TRPBF	LTKT	6-Lead Plastic SOT-23 -40°C to 85°C	
				1

Consult LTC Marketing for parts specified with wider operating temperature ranges.

Consult LTC Marketing for information on nonstandard lead based finish parts.

For more information on lead free part marking, go to: http://www.linear.com/leadfree/

For more information on tape and reel specifications, go to: http://www.linear.com/tapeandreel/

**ELECTRICAL CHARACTERISTICS** The  $\bullet$  denotes the specifications which apply over the full operating temperature range, otherwise specifications are at T<sub>A</sub> = 25°C. V<sub>IN</sub> = 3V, C<sub>FLY</sub> = 0.47µF, C<sub>OUT</sub>, C<sub>CPO</sub>, C<sub>IN</sub> = 4.7µF unless otherwise specified.

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS
V <sub>IN</sub> Operating Voltage			2.7		4.4	V
I <sub>VIN</sub> Shutdown Current	SHDN = 0V (Note 5)	•		4	8	μA
I <sub>VIN</sub> Operating Current	I <sub>OUT</sub> = 0mA, Burst Mode <sup>™</sup> Operation	•		190	330	μA
Regulated Output Voltage	I <sub>OUT</sub> = 1mA	•	4.9	5	5.1	V
V <sub>OUT</sub> Temperature Coefficient				±50		ppm
Charge Pump Oscillator Frequency	$I_{OUT} > 500 \mu A$ , $V_{IN} = 2.7 V$ to $4.4 V$	•	480	550	620	kHz
CPO Output Resistance	$V_{IN} = 2.7V, I_{OUT} = 10mA$ $V_{IN} = 4.4V, I_{OUT} = 10mA$	•		17 14	30 24	Ω Ω
V <sub>OUT</sub> Dropout Voltage (Note 3)	$I_{OUT} = 10$ mA, $V_{OUT} = 5V$	•			100	mV
V <sub>OUT</sub> Enable Time	R <sub>LOAD</sub> = 2k			0.6		ms
V <sub>OUT</sub> Output Noise Voltage	$I_{OUT}$ = 10mA, 10Hz $\leq$ f $\leq$ 100kHz $I_{OUT}$ = 10mA, 10Hz $\leq$ f $\leq$ 2.5MHz			90 800		μV <sub>RMS</sub> μV <sub>P-P</sub>
V <sub>OUT</sub> Line Regulation	$V_{IN} = 2.7V$ to 4.4V, $I_{OUT} = 0$	•		4	20	mV
V <sub>OUT</sub> Load Regulation	I <sub>OUT</sub> = 1mA to 10mA I <sub>OUT</sub> = 1mA to 30mA (Note 4)	•		2 4	10	mV mV
V <sub>OUT</sub> Shutdown Resistance	$ \begin{array}{l} \mbox{CN/\overline{SHDN}} = 0V \mbox{ (Note 5)} \\ \mbox{V}_{\rm IN} = 2.7V, \mbox{ Resistance Measured to Ground} \\ \mbox{V}_{\rm IN} = 4.4V, \mbox{ Resistance Measured to Ground} \end{array} $	•		160 100	400 300	Ω Ω



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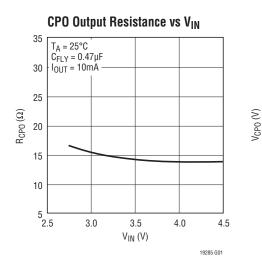
CN/SHDN Input Threshold	V <sub>IN</sub> = 2.7V to 4.4V (Note 5)	•	0.15	0.5	1.6	V
CN/SHDN Input Current	$CN/\overline{SHDN} = OV (Note 5)$		-1	-3	-6	μA

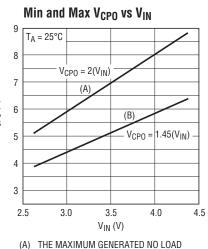
Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

Note 2: The LTC1928ES6-5 is guaranteed to meet performance specifications from 0°C to 70°C. Specifications over the -40°C to 85°C operating temperature range are assured by design, characterization and correlation with statistical process controls.

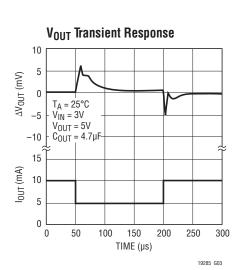
Note 3: Dropout voltage is the minimum input/output voltage required to maintain regulation at the specified output current. In dropout the output voltage will be equal to:  $V_{CPO} - V_{DBOPOUT}$  (see Figure 2).

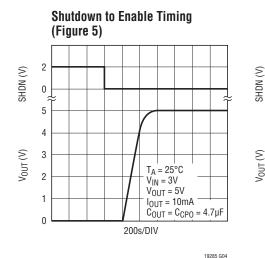
## TYPICAL PERFORMANCE CHARACTERISTICS



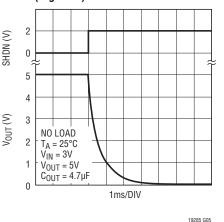


- CPO VOI TAGE (B)
- THE MINIMUM ALLOWABLE CPO VOLTAGE, AT FULL LOAD, TO ENSURE THAT THE LDO IS NOT DISABLED 19285 G02

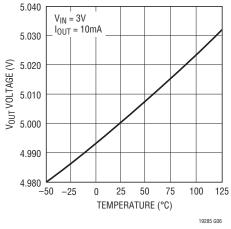




#### **Enable to Shutdown Timing** (Figure 5)



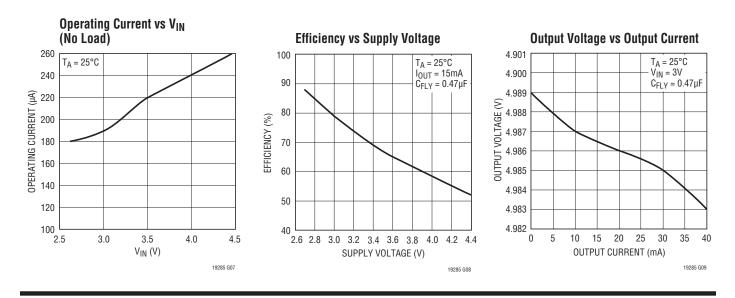
#### V<sub>OUT</sub> Voltage vs Temperature



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## **TYPICAL PERFORMANCE CHARACTERISTICS**



### **PIN FUNCTIONS**

 $V_{IN}$  (Pin 1): Input Voltage, 2.7V to 4.4V.  $V_{IN}$  should be bypassed with a  $\geq 2\mu F$  low ESR capacitor as close to the pin as possible for best performance. A minimum capacitance value of 0.1  $\mu F$  is required.

GND (Pin 2): System Ground.

**V<sub>OUT</sub> (Pin 3):** Low Noise Regulated Output Voltage. V<sub>OUT</sub> should be bypassed with a  $\ge 2\mu$ F low ESR capacitor as close to the pin as possible for best performance. The V<sub>OUT</sub> voltage is internally set to 5V.

**CPO (Pin 4):** Boosted Unregulated Voltage. Approximately  $1.95V_{IN}$  at low loads. Bypass with  $a \ge 2\mu$  Flow ESR capacitor.

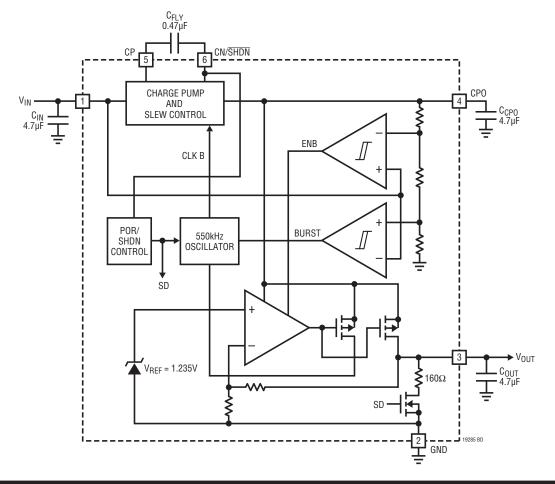
CP (Pin 5): Flying Capacitor Positive Input.

**CN/SHDN** (Pin 6): Flying Capacitor Negative Input and SHDN. When this pin is pulled to ground through a  $100\Omega$  resistor, the part will go into shutdown within approximately  $30\mu$ s.



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### **BLOCK DIAGRAM**



## **APPLICATIONS INFORMATION**

#### Operation

The LTC1928-5 uses a switched-capacitor charge pump to generate a CPO voltage of approximately  $2V_{IN}$ . CPO powers an internal low dropout linear regulator that supplies a regulated output at  $V_{OUT}$ . Internal comparators are used to sense CPO and  $V_{IN}$  voltages for power-up conditioning. The output current is sensed to determine the charge pump operating mode. A trimmed internal bandgap is used as the voltage reference and a trimmed internal oscillator is used to control the charge pump switches.

The charge pump is a doubler configuration that uses one external flying capacitor. When enabled, a 2-phase nonoverlapping clock controls the charge pump switches. At start-up, the LDO is disabled and the load is removed from CPO. When CPO reaches  $1.75V_{IN}$  the LDO is enabled. If CPO falls below  $1.45V_{IN}$  the LDO will be disabled. Gen-

erally, the charge pump runs open loop with continuous clocking for low noise. If CPO is greater than  $1.95V_{IN}$  and  $I_{OUT}$  is less than 200µA, the charge pump will operate in Burst Mode operation for increased efficiency but slightly higher output noise. In Burst Mode operation, the clock is disabled when CPO reaches  $1.95V_{IN}$  and enabled when CPO droops by about 150mV. The switching frequency is precisely controlled to ensure that the frequency is above 455kHz and at the optimum rate to ensure maximum efficiency. The switch edge rates are also controlled to minimize noise. The effective output resistance at CPO is dependent on the voltage at  $V_{IN}$ , CPO, the flying capacitor value  $C_{FLY}$  and the junction temperature. Alow ESR capacitor of  $\ge 2\mu$ F should be used at CPO for minimum noise.

The LDO is used to filter the ripple on CPO and to set an output voltage independent of CPO.  $V_{OUT}$  is set by an in-



# **APPLICATIONS INFORMATION**

ternal reference and resistor divider. The LDO requires a capacitor on V<sub>OUT</sub> for stability and improved load transient response. A low ESR capacitor of  $\geq 2\mu$ F should be used.

### Maximum I<sub>OUT</sub> Calculations

The maximum available current can be calculated based on the open circuit CPO voltage, the dropout voltage of the LDO and the effective output resistance of the charge pump. The open circuit CPO voltage is approximately  $2V_{IN}$  (see Figure 2).

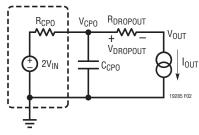


Figure 2. Equivalent Circuit

#### Example:

 $\begin{array}{l} V_{IN} = 3V \\ V_{OUT} = 5V \\ R_{CPO} = 30\Omega \\ Maximum \ unloaded \ CPO \ voltage = 2V_{IN} = 6V \\ V_{DROPOUT(MAX)} = 100mV \\ I_{OUT(MAX)} = (2V_{IN} - V_{DROPOUT(MAX)} - V_{OUT})/R_{CPO} \\ = (6V - 0.1V - 5V)/30\Omega = 30mA \end{array}$ 

 $V_{CPO}$  must be greater than 1.45V\_{IN} = 4.35V. To confirm this, calculate  $V_{CPO}$ :

 $V_{CPO} = 6V - (30 \text{mA} \cdot 30 \Omega) = 5.1 \text{V}$ 

For minimum noise applications the LDO must be kept out of dropout to prevent CPO noise from coupling into  $V_{\text{OUT}}.$ 

### External CPO Loading

The CPO output can drive an external load (for example, an LDO). The current required by this additional load will reduce the available current from  $V_{OUT}$ . If the external load requires 1mA, the available current at  $V_{OUT}$  will be reduced by 1mA.

### Short-Circuit and Thermal Protection

 $V_{\text{OUT}}\,\text{can}\,\text{be}\,\text{shorted}\,\text{to}\,\text{ground}\,\text{indefinitely}.$  Internal circuitry will limit the output current. If the junction temperature

exceeds 150°C the part will shut down. Excessive power dissipation due to heavy loads will also cause the part to shut down when the junction temperature exceeds 150°C. The part will become enabled when the junction temperature drops below 140°C. If the fault conditions remain in place, the part will cycle between the shutdown and enabled states.

#### **Capacitor Selection**

For best performance it is recommended that low ESR ceramic capacitors be used to reduce noise and ripple.  $C_{OUT}$  must be  ${\geq}2\mu$ F and  $C_{CPO}$  must be equal to or greater than  $C_{OUT}$ .  $C_{IN}$  is dependent on the input power supply source impedance. The charge pump demands large instantaneous currents which may induce ripple onto a common voltage rail.  $C_{IN}$  should be  ${\geq}2\mu$ F and a spike reducing resistor of  $2.2\Omega$  may be required between  $V_{IN}$  and the supply.

A low ESR ceramic capacitor is recommended for the flying capacitor  $C_{FLY}$  with a value of  $0.47\mu$ F. At low load or high  $V_{IN}$  a smaller capacitor could be used to reduce ripple on CPO which would reflect as lower ripple on  $V_{OUT}$ .

If a minimum enable time is required, the CPO output filter capacitor should be at least  $2 \times$  the V<sub>OUT</sub> filter capacitor. When the LDO is first enabled, the CPO capacitor will dump a large amount of charge into the V<sub>OUT</sub> capacitor. If the drop in the CPO voltage falls below 1.45V<sub>IN</sub> the LDO will be disabled and the CPO voltage will be required to charge up to 1.75V<sub>IN</sub> to enable the LDO. The resulting cycling extends the enable time.

### Output Ripple

The output ripple on CPO includes a spike component from the charge pump switches and a droop component which is dependent on the load current and the value of C3. The charge pump has been carefully designed to minimize the spike component, however, low ESR capacitors are essential to reduce the remaining spike energy effect on the CPO voltage.  $C_{CPO}$  should be increased for high load currents to minimize the droop component. Ripple components on CPO are greatly reduced at  $V_{OUT}$  by the LDO, however,  $C_{OUT}$  should also be a low ESR capacitor to improve filtering of the CPO noise.



### **APPLICATIONS INFORMATION**

#### Shutdown

When CN/SHDN = 0V, the part will be in shutdown, the supply current will be  $<8\mu$ A and V<sub>OUT</sub> will be shorted to ground through a 160 $\Omega$  switch. In addition, CPO will be high impedance and disconnected from V<sub>IN</sub> and CN/SHDN.

Shutdown is achieved by internally sampling the CN/SHDN pin for a low voltage. Time between shutdown samples is about  $30\mu$ s. During the sample time the charge pump switches are disabled and CN/SHDN must be pulled to ground within 400ns. A resistor value between  $100\Omega$  and 1k is recommended. Parasitic lead capacitance should be minimized on the CN/SHDN pin.

#### **Power-On Reset**

Upon initial power-up, a power-on reset circuit ensures that the internal functions are correctly initialized. Once  $V_{IN}$  reaches about 1V, the power-on reset circuit will enable the part as long as the CN/SHDN pin is not pulled low.

#### **Thermal Considerations**

The power handling capability of the device will be limited by the maximum rated junction temperature (125°C). The device dissipation  $P_D = I_{OUT}(2V_{IN} - V_{OUT}) + V_{IN}(2mA)$ . The

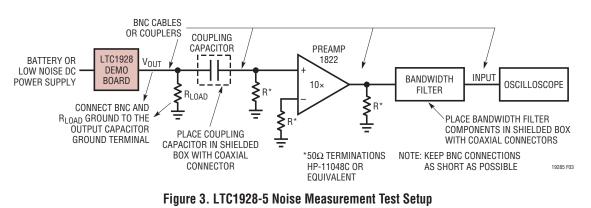
device dissipates the majority of its heat through its pins, especially GND (Pin 2). Thermal resistance to ambient can be optimized by connecting GND to a large copper region on the PCB, which serves as a heat sink. Applications that operate the LTC1928-5 near maximum power levels should maximize the copper area at all pins except CP and CN/SHDN and ensure that there is some airflow over the part to carry away excess heat.

#### **General Layout Considerations**

Due to the high switching frequency and high transient currents produced by the device, careful board layout is a must. A clean board layout using a ground plane and short connections to all capacitors will improve noise performance and ensure proper regulation.

#### **Measuring Output Noise**

Measuring the LTC1928 low noise levels requires care. Figure 3 shows a test setup for taking the measurement. Good connection and signal handling technique should yield about  $800\mu V_{P-P}$  over a 2.5MHz bandwidth. The noise measurement involves AC-coupling the LTC1928 output into the test setup's input and terminating this connection with  $50\Omega$ . Coaxial connections must be maintained to preserve measurement integrity.



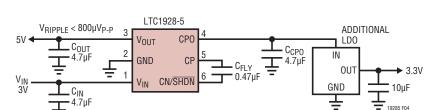
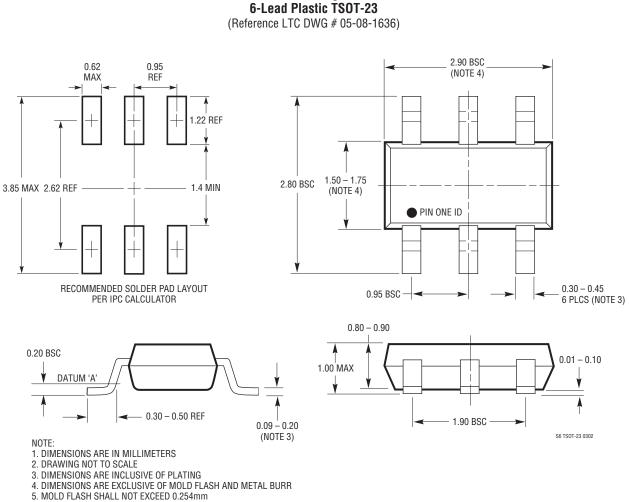


Figure 4. LTC1928-5, External Load on CPO, No Shutdown State



### PACKAGE DESCRIPTION

Please refer to http://www.linear.com/designtools/packaging/ for the most recent package drawings.



S6 Package

6. JEDEC PACKAGE REFERENCE IS MO-193



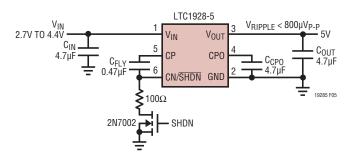
### **REVISION HISTORY**

REV	DATE	DESCRIPTION	PAGE NUMBER
А	09/15	Revised package drawing.	8





# TYPICAL APPLICATION





## **RELATED PARTS**

PART NUMBER	DESCRIPTION	COMMENTS
LTC1550/LTC1551	Low Noise, 900kHz Charge Pump	1mV <sub>P-P</sub> Typical Ripple, Up to 10mA
LT1611	Inverting 1.4MHz Switching Regulator 5V to -5V at 150mA, Low Output Noise	
LT1613	1.4MHz Boost Switching Regulator in ThinSOT	3.3V to 5V at 200mA, Low Noise PWM Operation
LTC1682	Doubler Charge Pump with Low Noise Linear Regulator 60µV <sub>RMS</sub> Noise, I <sub>OUT</sub> Up to 80mA, MSOP	
LTC1754-5	Micropower 5V Charge Pump in ThinSOT	$I_Q = 13\mu A$ , $I_{OUT}$ to 50mA, Shutdown
LT1761 Series	100mA ThinSOT, Low Noise LDO Regulators	20μΑ I <sub>Q</sub> , 20μV <sub>RMS</sub> Noise, 300mV Dropout
LTC3200	Constant Frequency Doubler Charge Pump	Low Noise, 5V Output or Adjustable



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