



TLV62569 2-A High Efficiency Synchronous Buck Converter in SOT Package

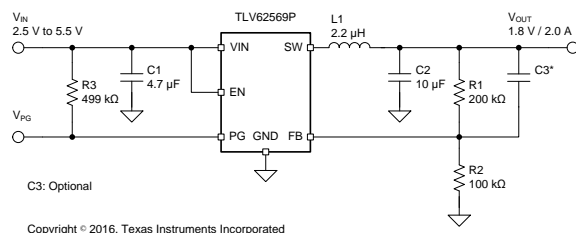
1 Features

- Up to 95% Efficiency
- Low $R_{DS(ON)}$ Switches 100 m Ω / 60 m Ω
- 2.5-V to 5.5-V Input Voltage Range
- Adjustable Output Voltage from 0.6 V to V_{IN}
- Power Save Mode for Light Load Efficiency
- 100% Duty Cycle for Lowest Dropout
- 35- μ A Operating Quiescent Current
- 1.5-MHz Typical Switching Frequency
- Power Good Output
- Over Current Protection
- Internal Soft Startup
- Thermal Shutdown Protection
- Available in SOT Package
- Pin-to-Pin Compatible with [TLV62568](#)
- Create a Custom Design Using the TLV62568 With the [WEBENCH® Power Designer](#)

2 Applications

- General Purpose POL Supply
- Set Top Box
- Network Video Camera
- Wireless Router
- Hard Disk Driver

4 Simplified Schematic



3 Description

The TLV62569 device is a synchronous step-down buck DC-DC converter optimized for high efficiency and compact solution size. The device integrates switches capable of delivering an output current up to 2 A.

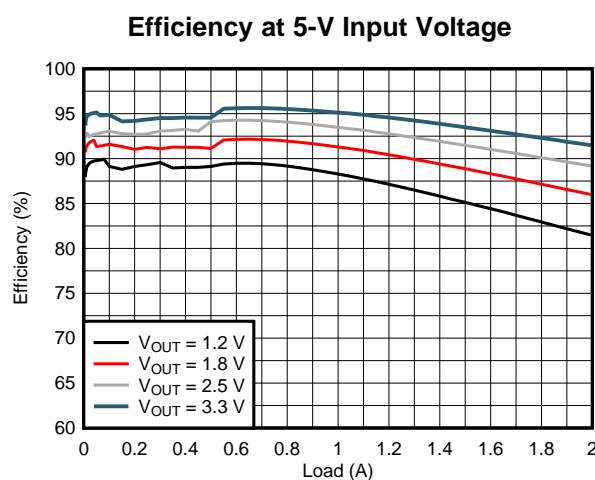
At medium to heavy loads, the device operates in pulse width modulation (PWM) mode with 1.5-MHz switching frequency. At light load, the device automatically enters Power Save Mode (PSM) to maintain high efficiency over the entire load current range. In shutdown, the current consumption is reduced to less than 2 μ A.

The TLV62569 provides an adjustable output voltage via an external resistor divider. An internal soft start circuit limits the inrush current during startup. Other features like over current protection, thermal shutdown protection and power good are built-in. The device is available in a SOT23 and SOT563 package.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TLV62569	SOT23 (5)	2.90 mm × 2.80 mm
TLV62569P	SOT23 (6)	
TLV62569	SOT563 (6)	1.60 mm x 1.60 mm
TLV62569P	SOT563 (6)	

(1) For all available packages, see the orderable addendum at the end of the datasheet.



D008



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5 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Changes from Original (December 2016) to Revision A	Page
• Added WEBENCH® Model	1

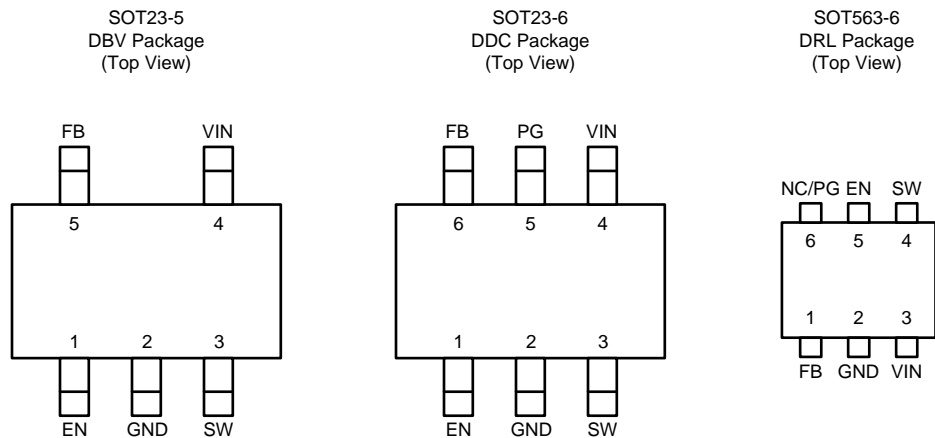
6 Device Options

PART NUMBER ⁽¹⁾	FUNCTION	PACKAGE MARKING	PACKAGE
TLV62569DBV	-	16AF	SOT23-5
TLV62569PDDC ⁽²⁾	Power Good		SOT23-6
TLV62569DRL ⁽²⁾	-		SOT563-6
TLV62569PDRL ⁽²⁾	Power Good		SOT563-6

(1) For detailed ordering information, please check [Mechanical, Packaging, Orderable Information](#) at the end of this datasheet.

(2) Product preview

7 Pin Configuration and Functions



Pin Functions

NAME	PIN NUMBER			I/O/PW R	DESCRIPTION
	SOT23-5	SOT23-6	SOT563-6		
EN	1	1	5	I	Device enable logic input. Logic high enables the device, logic low disables the device and turns it into shutdown. Do not leave floating.
GND	2	2	2	PWR	Ground pin.
SW	3	3	4	PWR	Switch pin connected to the internal FET switches and inductor terminal. Connect the inductor of the output filter to this pin.
VIN	4	4	3	PWR	Power supply voltage input.
PG	-	5	6	O	Power good open drain output pin for TLV62569P. The pull-up resistor should not be connected to any voltage higher than 5.5V. If it's not used, leave the pin floating.
FB	5	6	1	I	Feedback pin for the internal control loop. Connect this pin to an external feedback divider.
NC	-	-	6	O	No connection pin for TLV62569DRL. The pin can be connected to the output or the ground. Or leave it floating.

8 Specifications

8.1 Absolute Maximum Ratings

Over operating temperature range (unless otherwise noted)⁽¹⁾

		MIN	MAX	UNIT
Voltage ⁽²⁾	VIN, EN, PG	−0.3	6	V
	SW (DC)	−0.3	VIN+0.3	V
	SW (AC, less than 10ns) ⁽³⁾	−3.0	9	V
	FB	−0.3	5.5	V
Operating junction temperature, TJ		−40	150	°C
Storage temperature, Tstg		−65	150	°C

- (1) Stresses beyond those listed under *absolute maximum ratings* may cause permanent damage to the device. These are stress ratings only and the device is not switching. Functional operation of the device at these or any other conditions beyond those indicated under *recommended operating conditions* is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values are with respect to network ground terminal.
- (3) Normal switching operation

8.2 ESD Ratings

			VALUE	UNIT
V(ESD)	Electrostatic discharge	Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 ⁽¹⁾	±2000	V
		Charged-device model (CDM), per JEDEC specification JESD22-C101 ⁽²⁾	±500	V

- (1) JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.
- (2) JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.

8.3 Recommended Operating Conditions⁽¹⁾

		MIN	TYP	MAX	UNIT
VIN	Input voltage	2.5		5.5	V
VOU	Output voltage	0.6		VIN	V
IOUT	Output current	0		2	A
TJ	Operating junction temperature	−40		125	°C
ISINK_PG	Sink current at PG pin			1	mA

- (1) Refer to the [Application and Implementation](#) section for further information.

8.4 Thermal Information

	THERMAL METRIC ⁽¹⁾	TLV62569 DBV (5 Pins)	UNIT
RθJA	Junction-to-ambient thermal resistance	188.2	°C/W
RθJC(top)	Junction-to-case (top) thermal resistance	137.5	°C/W
RθJB	Junction-to-board thermal resistance	41.2	°C/W
ψJT	Junction-to-top characterization parameter	31.4	°C/W
ψJB	Junction-to-board characterization parameter	40.6	°C/W
RθJC(bot)	Junction-to-case (bottom) thermal resistance	n/a	°C/W

- (1) For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, [SPRA953](#).

8.5 Electrical Characteristics

 $V_{IN} = 5.0\text{ V}$, $T_J = 25^\circ\text{C}$, unless otherwise noted

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
SUPPLY						
I _Q	Quiescent current into VIN pin	Not switching	35			uA
I _{SD}	Shutdown current into VIN pin	EN = 0 V	0.1	2		μA
V _{UVLO}	Under voltage lock out	V _{IN} falling	2.3	2.45		V
	Under voltage lock out hysteresis		100			mV
T _{JSD}	Thermal shutdown	Junction temperature rising	150			°C
		Junction temperature falling	130			
LOGIC INTERFACE						
V _{IH}	High-level input voltage at EN pin	2.5 V ≤ V _{IN} ≤ 5.5 V	1.2	0.95		V
V _{IL}	Low-level input voltage at EN pin	2.5 V ≤ V _{IN} ≤ 5.5 V		0.85	0.4	V
t _{SS}	Soft startup time			800		μs
V _{PG}	Power good threshold	V _{FB} rising, referenced to V _{FB} nominal		95%		
		V _{FB} falling, referenced to V _{FB} nominal		90%		
V _{PG,OL}	Power good low-level output voltage	I _{SINK} = 1 mA			0.4	V
I _{PG,LKG}	Input leakage current into PG pin	V _{PG} = 5.0 V		0.01		μA
t _{PG,DLY}	Power good delay time	V _{FB} falling		40		μs
OUTPUT						
V _{FB}	Feedback regulation voltage		0.588	0.6	0.612	V
R _{DS(on)}	High-side FET on resistance		100			mΩ
	Low-side FET on resistance		60			
I _{LIM}	High-side FET current limit		3			A
f _{SW}	Switching frequency	V _{OUT} = 2.5 V	1.5			MHz

8.6 Typical Characteristics

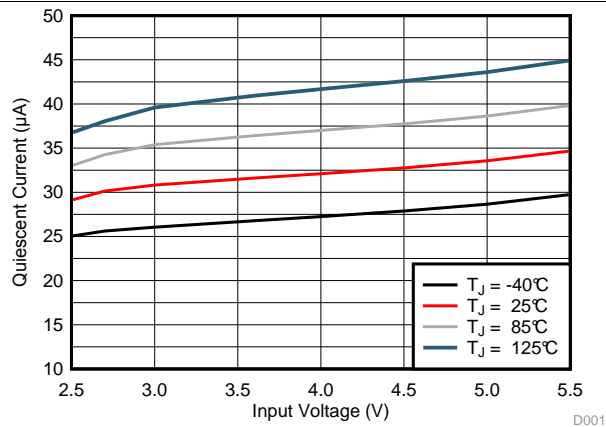


Figure 1. Quiescent Current vs Input Voltage

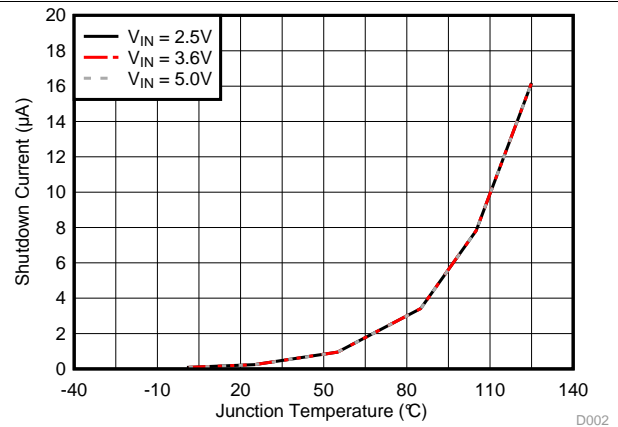


Figure 2. Shutdown Current vs Junction Temperature

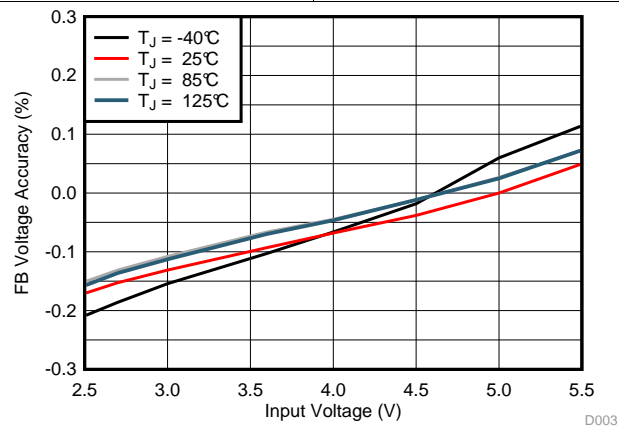


Figure 3. FB Voltage Accuracy

9 Detailed Description

9.1 Overview

The TLV62569 is a high-efficiency synchronous step-down converter. The device operates with an adaptive off time with peak current control scheme. The device operates at typically 1.5-MHz frequency pulse width modulation (PWM) at moderate to heavy load currents. Based on the V_{IN}/V_{OUT} ratio, a simple circuit sets the required off time for the low-side MOSFET. It makes the switching frequency relatively constant regardless of the variation of input voltage, output voltage, and load current.

9.2 Functional Block Diagrams

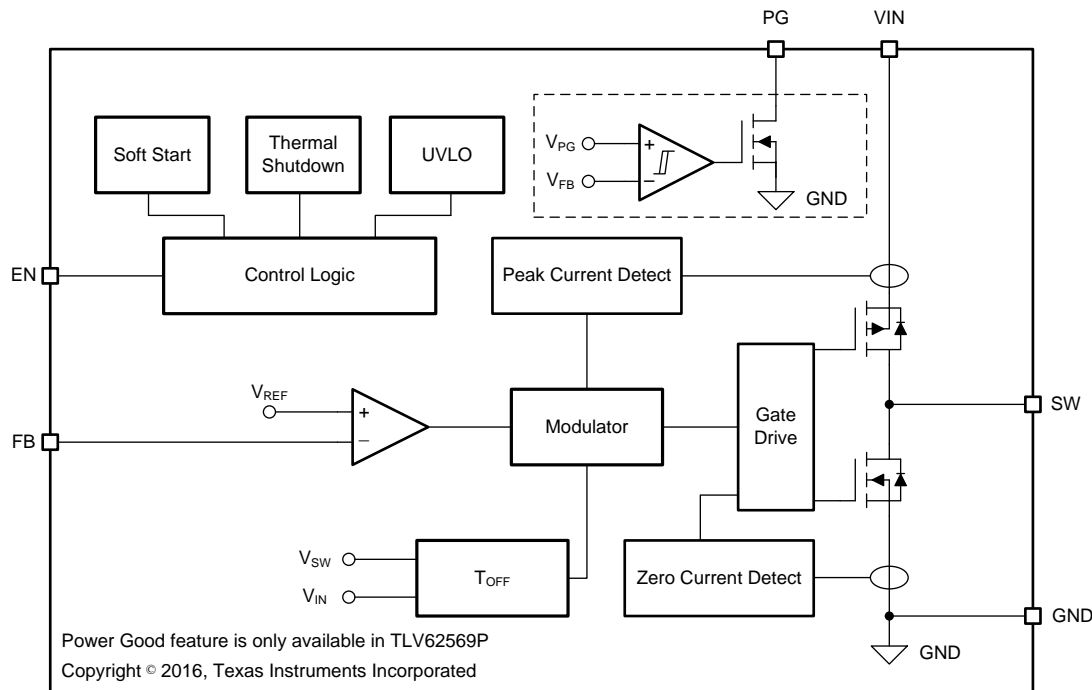


Figure 4. TLV62569 Functional Block Diagram

9.3 Feature Description

9.3.1 Power Save Mode

The device automatically enters Power Save Mode to improve efficiency at light load when the inductor current becomes discontinuous. In Power Save Mode, the converter reduces switching frequency and minimizes current consumption. In Power Save Mode, the output voltage rises slightly above the nominal output voltage. This effect is minimized by increasing the output capacitor.

9.3.2 100% Duty Cycle Low Dropout Operation

The device offers a low input-to-output voltage differential by entering 100% duty cycle mode. In this mode, the high-side MOSFET switch is constantly turned on and the low-side MOSFET is switched off. The minimum input voltage to maintain output regulation, depending on the load current and output voltage, is calculated as:

$$V_{IN(MIN)} = V_{OUT} + I_{OUT} \times (R_{DS(ON)} + R_L)$$

where

- $R_{DS(ON)}$ = High side FET on-resistance
- R_L = Inductor ohmic resistance (DCR)

(1)

Feature Description (continued)

9.3.3 Soft Startup

After enabling the device, internal soft startup circuitry ramps up the output voltage which reaches nominal output voltage during a startup time. This avoids excessive inrush current and creates a smooth output voltage rise slope. It also prevents excessive voltage drops of primary cells and rechargeable batteries with high internal impedance.

The TLV62569 is able to start into a pre-biased output capacitor. The converter starts with the applied bias voltage and ramps the output voltage to its nominal value.

9.3.4 Switch Current Limit

The switch current limit prevents the device from high inductor current and drawing excessive current from a battery or input voltage rail. Excessive current might occur with a heavy load or shorted output circuit condition. The TLV62569 adopts the peak current control by sensing the current of the high-side switch. Once the high-side switch current limit is reached, the high-side switch is turned off and low-side switch is turned on to ramp down the inductor current with an adaptive off-time.

9.3.5 Under Voltage Lockout

To avoid mis-operation of the device at low input voltages, under voltage lockout is implemented that shuts down the device at voltages lower than V_{UVLO} with V_{HYS_UVLO} hysteresis.

9.3.6 Thermal Shutdown

The device enters thermal shutdown once the junction temperature exceeds the thermal shutdown rising threshold, T_{JSD} . Once the junction temperature falls below the falling threshold, the device returns to normal operation automatically.

9.4 Device Functional Modes

9.4.1 Enabling/Disabling the Device

The device is enabled by setting the EN input to a logic High. Accordingly, a logic Low disables the device. If the device is enabled, the internal power stage starts switching and regulates the output voltage to the set point voltage. The EN input must be terminated and should not be left floating.

9.4.2 Power Good

The TLV62569P has a power good output. The PG pin goes high impedance once the output is above 95% of the nominal voltage, and is driven low once the output voltage falls below typically 90% of the nominal voltage. The PG pin is an open-drain output and is specified to sink up to 1 mA. The power good output requires a pull-up resistor connecting to any voltage rail less than 5.5 V. The PG signal can be used for sequencing of multiple rails by connecting it to the EN pin of other converters. Leave the PG pin unconnected when not used.

Table 1. PG Pin Logic

DEVICE CONDITIONS		LOGIC STATUS	
		HIGH Z	LOW
Enable	EN = High, $V_{FB} \geq V_{PG}$	√	
	EN = High, $V_{FB} \leq V_{PG}$		√
Shutdown	EN = Low		√
Thermal Shutdown	$T_J > T_{JSD}$		√
UVLO	$1.4\text{ V} < V_{IN} < V_{UVLO}$		√
Power Supply Removal	$V_{IN} \leq 1.4\text{ V}$	√	

10 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

10.1 Application Information

The following section discusses the design of the external components to complete the power supply design for several input and output voltage options by using typical applications as a reference.

10.2 Typical Application

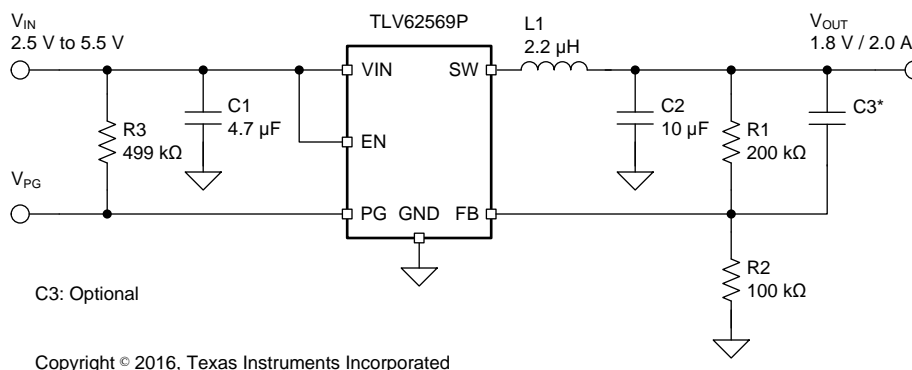


Figure 5. TLV62569 1.8-V Output Application

10.2.1 Design Requirements

For this design example, use the parameters listed in [Table 2](#) as the input parameters.

Table 2. Design Parameters

DESIGN PARAMETER	EXAMPLE VALUE
Input voltage	2.5 V to 5.5 V
Output voltage	1.8 V
Maximum output current	2.0 A

[Table 3](#) lists the components used for the example.

Table 3. List of Components

REFERENCE	DESCRIPTION	MANUFACTURER ⁽¹⁾
C1	4.7 μF, Ceramic Capacitor, 10 V, X7R, size 0805, GRM21BR71A475KA73L	Murata
C2	10 μF, Ceramic Capacitor, 10 V, X7R, size 0805, GRM21BR71A106KE51L	Murata
L1	2.2 μH, Power Inductor, size 4mmx4mm, XAL4020-222ME	Coilcraft
R1,R2,R3	Chip resistor, 1%, size 0603	Std.
C3	Optional, 6.8 pF if it is needed	Std.

(1) See [Third-party Products Disclaimer](#)

10.2.2 Detailed Design Procedure

10.2.2.1 Custom Design With WEBENCH® Tools

[Click here](#) to create a custom design using the TLV62569 device with the WEBENCH® Power Designer.

1. Start by entering the input voltage (V_{IN}), output voltage (V_{OUT}), and output current (I_{OUT}) requirements.
2. Optimize the design for key parameters such as efficiency, footprint, and cost using the optimizer dial.
3. Compare the generated design with other possible solutions from Texas Instruments.

The WEBENCH Power Designer provides a customized schematic along with a list of materials with real-time pricing and component availability.

In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance
- Run thermal simulations to understand board thermal performance
- Export customized schematic and layout into popular CAD formats
- Print PDF reports for the design, and share the design with colleagues

Get more information about WEBENCH tools at www.ti.com/WEBENCH.

10.2.2.2 Setting the Output Voltage

An external resistor divider is used to set output voltage according to [Equation 2](#).

When sizing R2, in order to achieve low current consumption and acceptable noise sensitivity, use a maximum of 200 k Ω for R2. Larger currents through R2 improve noise sensitivity and output voltage accuracy but increase current consumption.

$$V_{OUT} = V_{FB} \times \left(1 + \frac{R1}{R2}\right) = 0.6V \times \left(1 + \frac{R1}{R2}\right) \quad (2)$$

A feed forward capacitor, C3 improves the loop bandwidth to make a fast transient response (shown in [Figure 19](#)). 6.8-pF capacitance is recommended for R2 of 100-k Ω resistance. A more detailed discussion on the optimization for stability vs. transient response can be found in [SLVA289](#).

10.2.2.3 Output Filter Design

The inductor and output capacitor together provide a low-pass filter. To simplify this process, [Table 4](#) outlines possible inductor and capacitor value combinations. Checked cells represent combinations that are proven for stability by simulation and lab test. Further combinations should be checked for each individual application.

Table 4. Matrix of Output Capacitor and Inductor Combinations

V_{OUT} [V]	L [μ H] ⁽¹⁾	C_{OUT} [μ F] ⁽²⁾				
		4.7	10	22	2 x 22	100
$0.6 \leq V_{OUT} < 1.2$	1				+	
	2.2				++ ⁽³⁾	
$1.2 \leq V_{OUT} < 1.8$	1			+	+	
	2.2			++ ⁽³⁾	+	
$1.8 \leq V_{OUT}$	1		+	+	+	
	2.2		++ ⁽³⁾	+	+	

(1) Inductor tolerance and current de-rating is anticipated. The effective inductance can vary by +20% and -30%.

(2) Capacitance tolerance and bias voltage de-rating is anticipated. The effective capacitance can vary by +20% and -50%.

(3) This LC combination is the standard value and recommended for most applications.

10.2.2.4 Inductor Selection

The main parameters for inductor selection is inductor value and then saturation current of the inductor. To calculate the maximum inductor current under static load conditions, Equation 3 is given:

$$I_{L,MAX} = I_{OUT,MAX} + \frac{\Delta I_L}{2}$$

$$\Delta I_L = V_{OUT} \times \frac{1 - \frac{V_{OUT}}{V_{IN}}}{L \times f_{SW}}$$

where:

- $I_{OUT,MAX}$ is the maximum output current
- ΔI_L is the inductor current ripple
- f_{SW} is the switching frequency
- L is the inductor value

(3)

It is recommended to choose a saturation current for the inductor that is approximately 20% to 30% higher than $I_{L,MAX}$. In addition, DC resistance and size should also be taken into account when selecting an appropriate inductor.

10.2.2.5 Input and Output Capacitor Selection

The architecture of the TLV62569 allows use of tiny ceramic-type output capacitors with low equivalent series resistance (ESR). These capacitors provide low output voltage ripple and are thus recommended. To keep its resistance up to high frequencies and to achieve narrow capacitance variation with temperature, it is recommended to use X7R or X5R dielectric.

The input capacitor is the low impedance energy source for the converter that helps provide stable operation. A low ESR multilayer ceramic capacitor is recommended for best filtering. For most applications, 4.7- μ F input capacitance is sufficient; a larger value reduces input voltage ripple.

The TLV62569 is designed to operate with an output capacitor of 10 μ F to 47 μ F, as outlined in Table 4.

10.2.3 Application Performance Curves

$V_{IN} = 5$ V, $V_{OUT} = 1.8$ V, $L = 2.2$ μ H, $T_A = 25$ °C, unless otherwise noted.

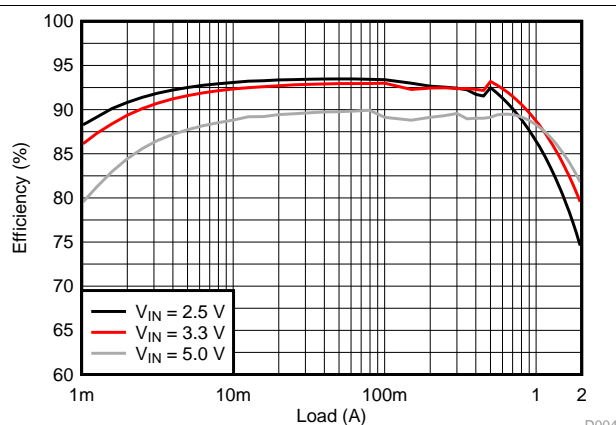


Figure 6. 1.2-V Output Efficiency

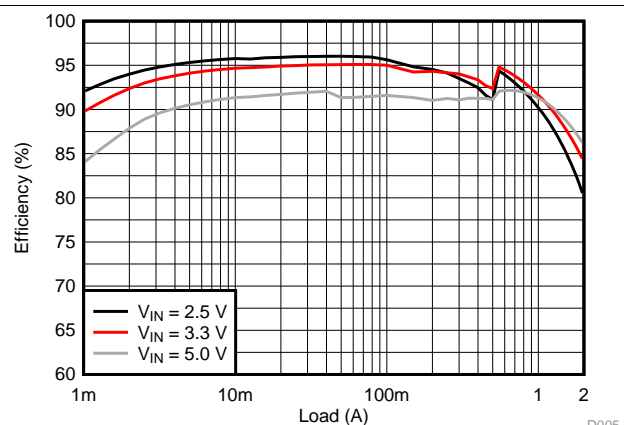
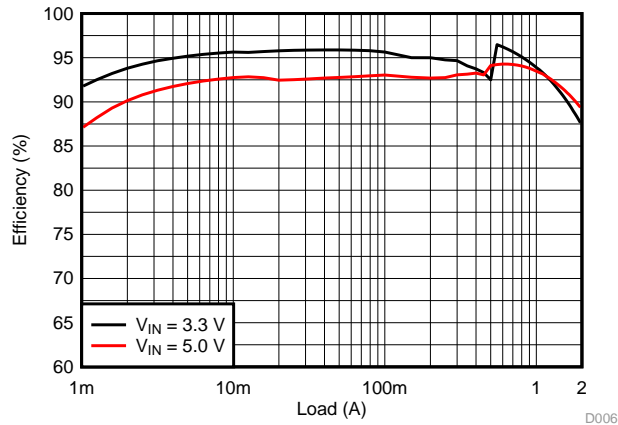
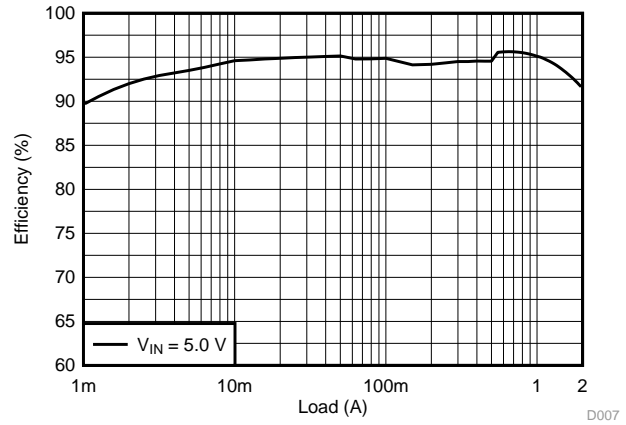
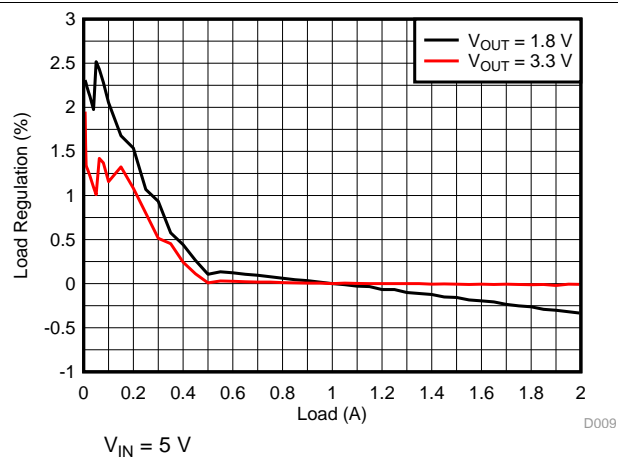
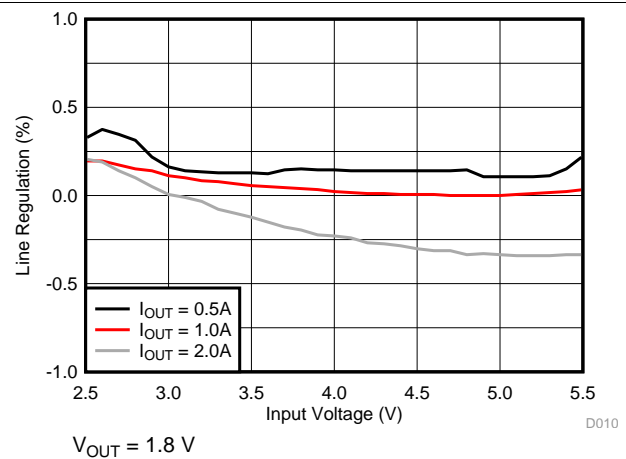
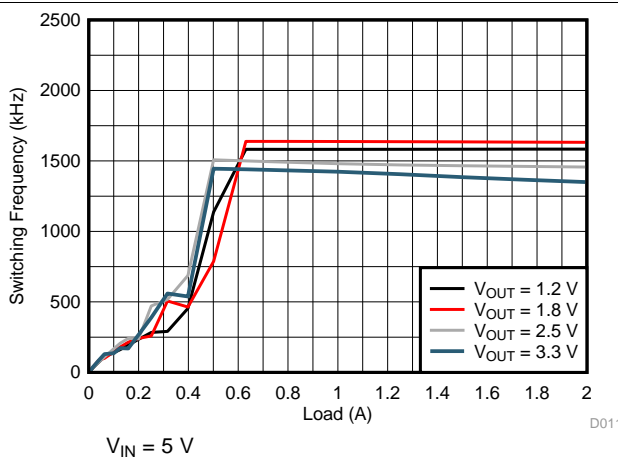
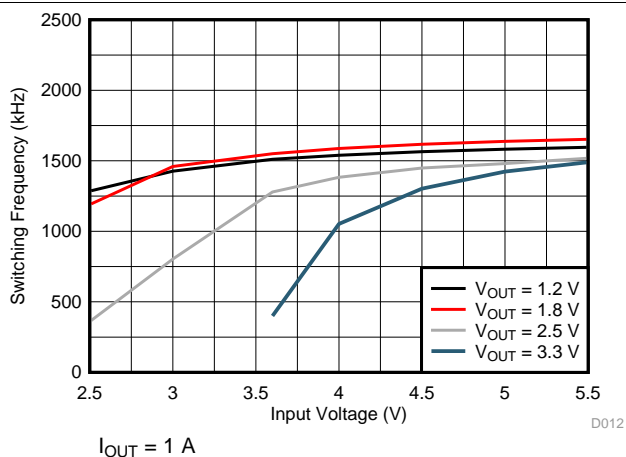
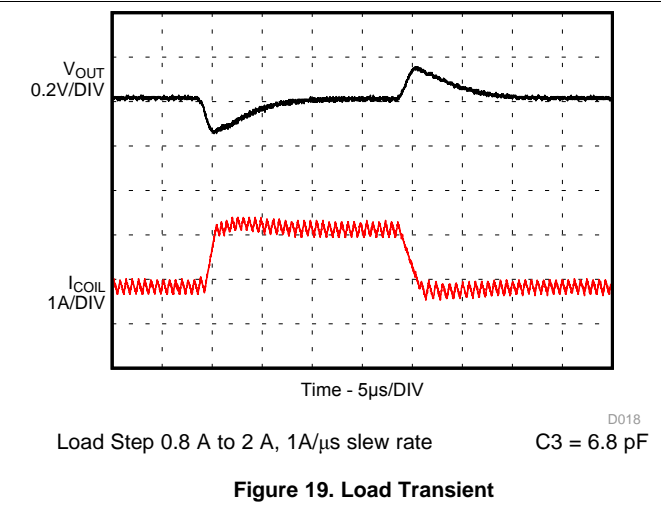
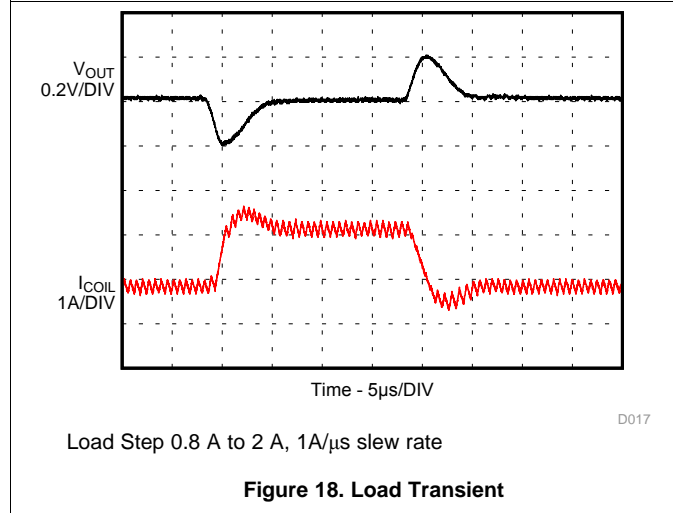
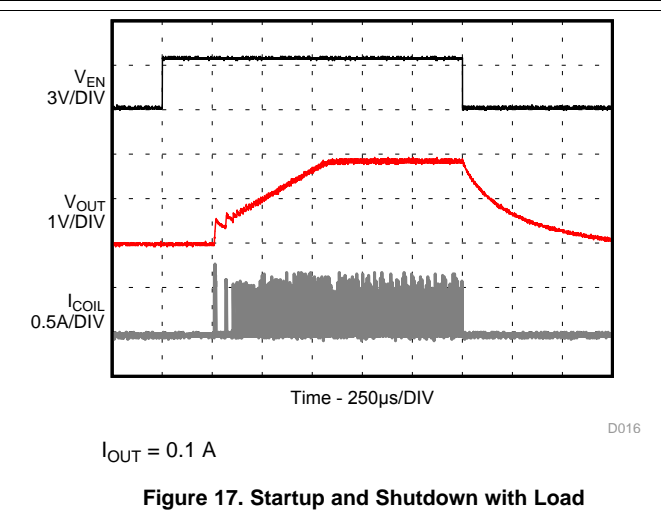
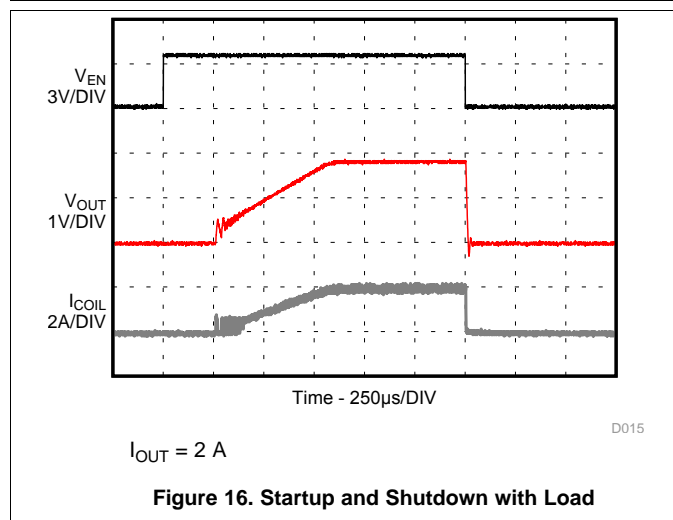
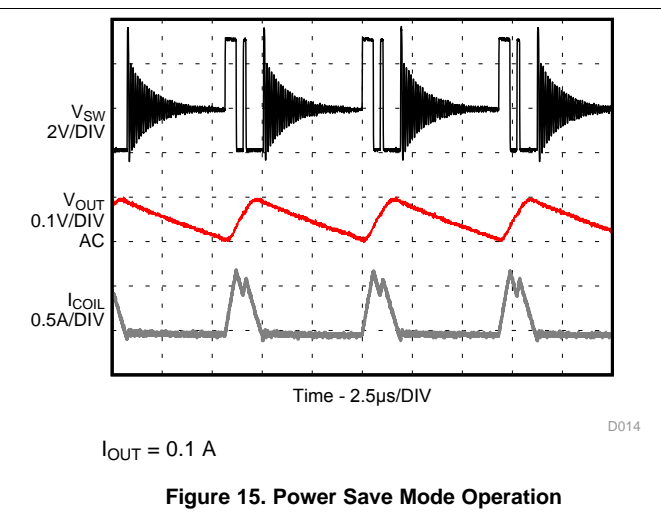
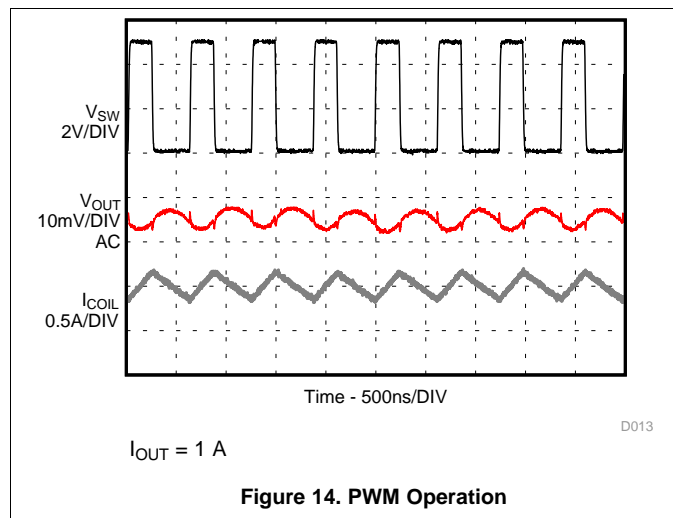


Figure 7. 1.8-V Output Efficiency


Figure 8. 2.5-V Output Efficiency

Figure 9. 3.3-V Output Efficiency

Figure 10. Load Regulation

Figure 11. Line Regulation

Figure 12. Switching Frequency vs Load

Figure 13. Switching Frequency vs Input Voltage



11 Power Supply Recommendations

The power supply to the TLV62569 must have a current rating according to the supply voltage, output voltage and output current.

12 Layout

12.1 Layout Guidelines

The PCB layout is an important step to maintain the high performance of the TLV62569 device.

- The input/output capacitors and the inductor should be placed as close as possible to the IC. This keeps the power traces short. Routing these power traces direct and wide results in low trace resistance and low parasitic inductance.
- The low side of the input and output capacitors must be connected properly to the power GND to avoid a GND potential shift.
- The sense traces connected to FB are signal traces. Special care should be taken to avoid noise being induced. Keep these traces away from SW nodes.
- GND layers might be used for shielding.

12.2 Layout Example

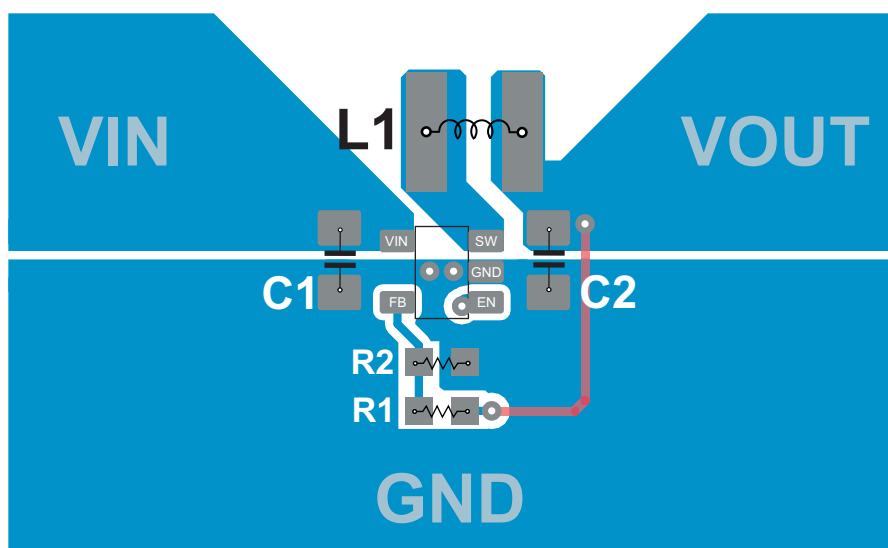


Figure 20. TLV62569 Layout

12.3 Thermal Considerations

Implementation of integrated circuits in low-profile and fine-pitch surface-mount packages typically requires special attention to power dissipation. Many system-dependent issues such as thermal coupling, airflow, convection surfaces, and the presence of other heat-generating components affect the power dissipation limits of a given component.

Two basic approaches for enhancing thermal performance are listed below:

- Improving the power dissipation capability of the PCB design
- Introducing airflow in the system

For more details on how to use the thermal parameters, see the application notes: Thermal Characteristics Application Notes [SZZA017](#) and [SPRA953](#).

13 Device and Documentation Support

13.1 Device Support

13.1.1 Third-Party Products Disclaimer

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13.1.2 Development Support

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In most cases, these actions are available:

- Run electrical simulations to see important waveforms and circuit performance
- Run thermal simulations to understand board thermal performance
- Export customized schematic and layout into popular CAD formats
- Print PDF reports for the design, and share the design with colleagues

Get more information about WEBENCH tools at www.ti.com/WEBENCH.

13.2 Documentation Support

13.2.1 Related Documentation

Semiconductor and IC Package Thermal Metrics Application Report ([SPRA953](#))

Thermal Characteristics of Linear and Logic Packages Using JEDEC PCB Designs Application Report ([SZZA017](#))

13.3 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

13.4 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

TI E2E™ Online Community *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

13.5 Trademarks

E2E is a trademark of Texas Instruments.

WEBENCH is a registered trademark of Texas Instruments.

13.5 Trademarks (continued)

All other trademarks are the property of their respective owners.

13.6 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

13.7 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

14 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TLV62569DBVR	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	16AF	Samples
TLV62569DBVT	ACTIVE	SOT-23	DBV	5	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 125	16AF	Samples
XLV62569PDDCR	ACTIVE	SOT-23-THIN	DDC	6	3000	TBD	Call TI	Call TI	-40 to 125		Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead/Ball Finish - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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TAPE AND REEL INFORMATION


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TLV62569DBVR	SOT-23	DBV	5	3000	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
TLV62569DBVT	SOT-23	DBV	5	250	180.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

TAPE AND REEL BOX DIMENSIONS

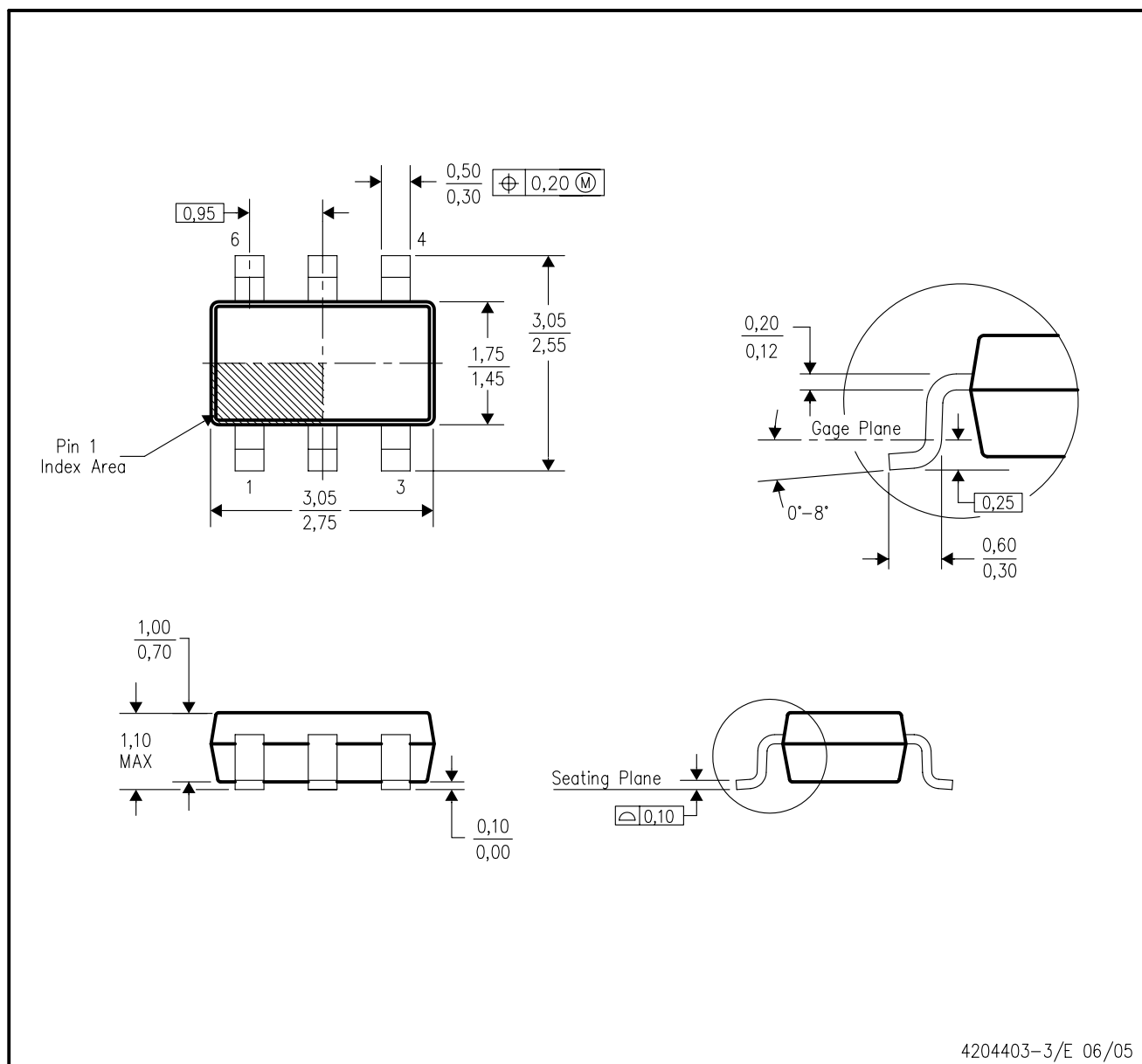


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TLV62569DBVR	SOT-23	DBV	5	3000	210.0	185.0	35.0
TLV62569DBVT	SOT-23	DBV	5	250	210.0	185.0	35.0

DDC (R-PDSO-G6)

PLASTIC SMALL-OUTLINE



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusion.
 - Falls within JEDEC MO-193 variation AA (6 pin).

DBV (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.
 - Falls within JEDEC MO-178 Variation AA.

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