

# Ultrasound RTZ Transmit Pulser Low Jitter MOSFET Driver

## Features

- ▶ 2-Pin Control 3-level RTZ Pulser
- ▶ Up to 200MHz clock input
- ▶ 6.0ns rise and fall time
- ▶ 2.0A peak output source/sink current
- ▶ 1.8 to 5.0V input CMOS compatible
- ▶ 5.0 to 10V total supply voltage
- ▶ Smart logic threshold
- ▶ Re timing register for low jitter
- ▶ Four matched channels
- ▶ Drives two P- and two N-channel MOSFETs
- ▶ Outputs can swing below ground
- ▶ Low inductance, quad flat no-lead package
- ▶ High performance, thermally enhanced packaging

## Applications

- ▶ Medical ultrasound imaging
- ▶ Piezoelectric transducer drivers
- ▶ Non-Destructive Testing (NDT)
- ▶ PIN diode driver
- ▶ CCD Clock driver/buffer
- ▶ High speed level translator

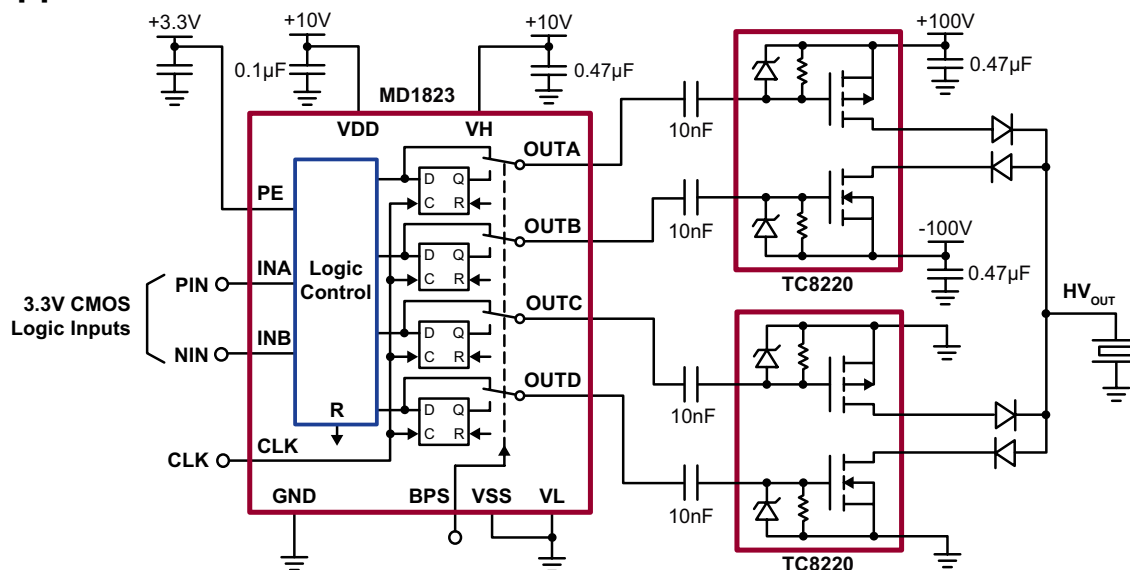
## General Description

The Supertex MD1823 is a high speed, four channel MOSFET driver designed to drive high voltage P- and N-channel MOSFETs for medical ultrasound applications and other applications requiring a high output current for a capacitive load. The high-speed input stage of the MD1823 can operate from a 1.8 to 5.0V logic interface with an optimum operating input signal range of 1.8 to 3.3V. An adaptive threshold circuit is used to set the level translator switch threshold to the average of the input logic 0 and logic 1 levels. The input logic levels can be ground referenced even though the driver is putting out bipolar signals. The level translator uses a proprietary circuit, which provides DC coupling together with high-speed operation.

The output stage of the MD1823 has separate power connections enabling the output signal L and H levels to be chosen independently from the supply voltages used for the majority of the circuit. The output stage is capable of peak currents of up to  $\pm 2.0A$ , depending on the supply voltages used and load capacitance present. The PE pin serves a dual purpose. First, its logic H level is used to compute the threshold voltage level for the channel input level translators. Second, when PE is low, the outputs are disabled, with the A & C output high and the B & D output low. This assists in properly pre charging the AC coupling capacitors that may be used in series in the gate drive circuit of an external PMOS and NMOS transistor pair. The PE pin should not be toggled on/off during imaging in an attempt to save power. Sporadic pulses may occur in doing so. It should be toggled only when the complete system goes in and out of standby mode.

The MD1823 has a clock input pin to help realign the input control pins to a master clock signal. This will help minimize delay time errors and clock jitter. A logic input pin, BPS, can be tied high for users not wanting to use the clock realignment feature. If BPS = 1, the clock and registers are ignored. The MD1823 only uses two input pins for the three level RTZ ultrasound pulser MOSFET gate driver.

## Typical Application Circuit



## Ordering Information

Part Number	Package	Packing
MD1823K6-G	16-Lead (3x3) QFN	3000/Reel

-G denotes a lead (Pb)-free / RoHS compliant package



ESD Sensitive Device

## Absolute Maximum Ratings

Parameter	Value
$V_{DD}-V_{SS}$ , Logic supply voltage	-0.5V to +12.5V
$V_H$ , Output high supply voltage	$V_L - 0.5V$ to $V_{DD} + 0.5V$
$V_L$ , Output low supply voltage	$V_{SS} - 0.5V$ to $V_H + 0.5V$
$V_{SS}$ , Low side supply voltage	-6.0V to +0.5V
Logic input levels	$V_{SS} - 0.5V$ to GND +5.5V
Maximum junction temperature	+125°C
Storage temperature	-65°C to 150°C
Operating temperature	-20°C to +85°C
Package power dissipation*	2.2W

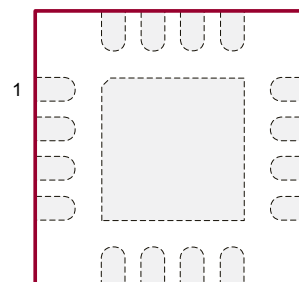
Absolute Maximum Ratings are those values beyond which damage to the device may occur. Functional operation under these conditions is not implied. Continuous operation of the device at the absolute rating level may affect device reliability. All voltages are referenced to device ground.

\* 1.0oz 4-layer 3x4" PCB

## Typical Thermal Resistance

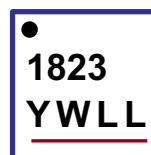
Package	$\theta_{ja}$
16-Lead (3x3) QFN	33°C/W

## Pin Configuration



16-Lead QFN  
(top view)

## Product Marking



Y = Last Digit of Year Sealed  
W = Code for Week Sealed  
L = Lot Number  
— = "Green" Packaging

Package may or may not include the following marks: Si or

16-Lead QFN

## DC Electrical Characteristics ( $V_H = V_{DD} = 10V$ , $V_L = V_{SS} = GND = 0V$ , $V_{PE} = 3.3V$ , $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$V_{DD} - V_{SS}$	Logic supply voltage	4.75	-	11.5	V	$4.0V \leq V_{DD} \leq 11.5V$
$V_{SS}$	Low side supply voltage	-5.5	-	0	V	---
$V_H$	Output high supply voltage	$V_{SS} + 2.0$	-	$V_{DD}$	V	---
$V_L$	Output low supply voltage	$V_{SS}$	-	$V_{DD} - 4.0$	V	---
$I_{DDQ}$	$V_{DD}$ quiescent current	-	60	-	$\mu A$	No input transitions, PE = 0
$I_{HQ}$	$V_H$ quiescent current	-	2.0	-	$\mu A$	
$I_{DDQ}$	$V_{DD}$ quiescent current	-	1.0	-	mA	No input transitions, PE = 1
$I_{HQ}$	$V_H$ quiescent current	-	2.0	-	$\mu A$	
$I_{DD}$	$V_{DD}$ average current	-	37.6	-	mA	One channel on at 5.0MHz, $f_{CLK} = 100MHz$ . No load
$I_H$	$V_H$ average current	-	11	-	mA	
$V_{IH}$	Input logic voltage high	$V_{PE} - 0.3$	-	$V_{PE}$	V	For logic inputs INA and INB
$V_{IL}$	Input logic voltage low	0	-	0.3	V	
$I_{IH}$	Input logic current high	-	-	1.0	$\mu A$	
$I_{IL}$	Input logic current low	-	-	1.0	$\mu A$	
$V_{IH}$	PE input logic voltage high	1.70	3.30	5.25	V	For logic input PE
$V_{IL}$	PE input logic voltage low	0	-	0.3	V	
$R_{IN PE}$	PE input impedance to GND	100	-	-	K $\Omega$	

**DC Electrical Characteristics (cont.)** ( $V_H = V_{DD} = 10V$ ,  $V_L = V_{SS} = GND = 0V$ ,  $V_{PE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$C_{IN}$	Logic input capacitance	-	5.0	10	pF	---
$R_{SINK}$	Output sink resistance	-	1.5	-	$\Omega$	$I_{SINK} = 50mA$
$R_{SOURCE}$	Output source resistance	-	2.0	-	$\Omega$	$I_{SOURCE} = 50mA$
$I_{SINK}$	Peak output sink current	-	2.0	-	A	---
$I_{SOURCE}$	Peak output source current	-	2.0	-	A	---

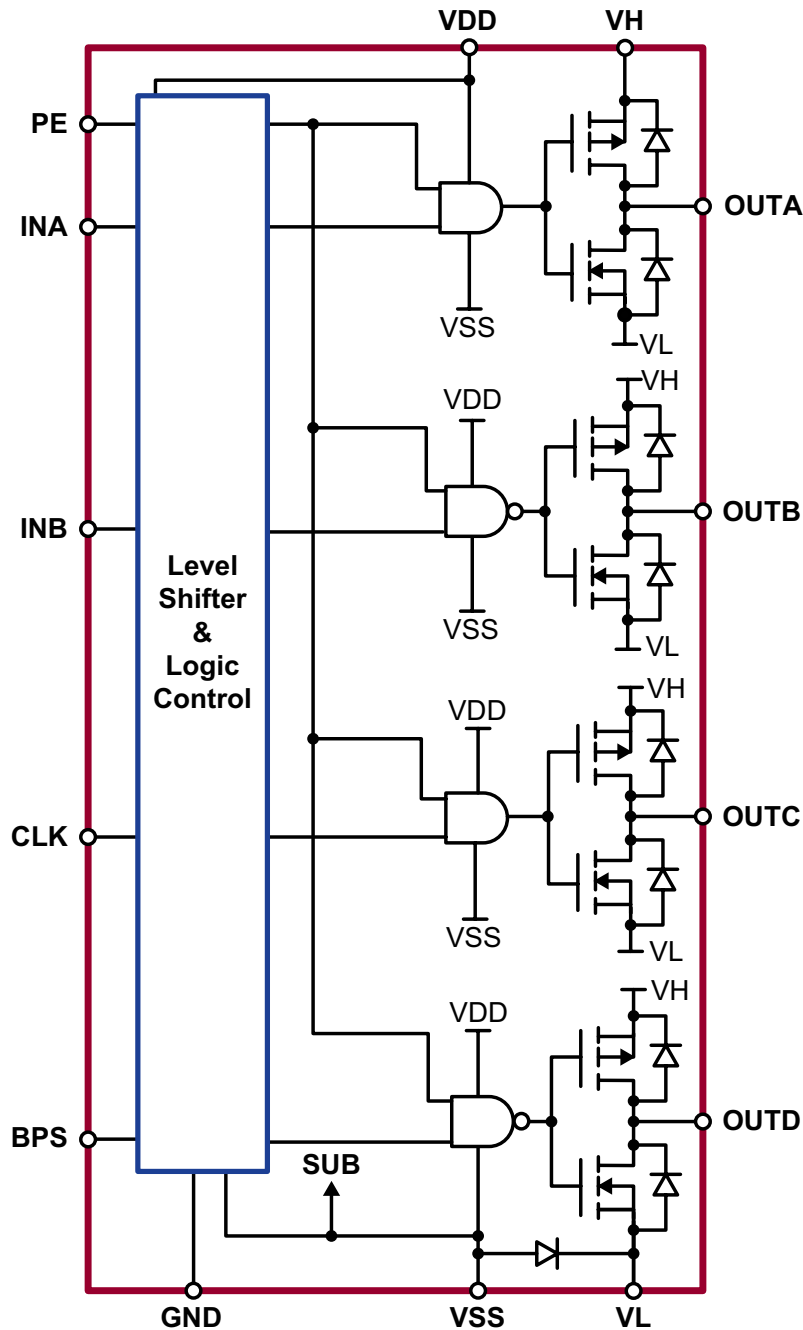
**AC Electrical Characteristics** ( $V_H = V_{DD} = 10V$ ,  $V_L = V_{SS} = GND = 0V$ ,  $V_{PE} = 3.3V$ ,  $T_A = 25^\circ C$ )

Sym	Parameter	Min	Typ	Max	Units	Conditions
$t_{irf}$	Input or PE rise & fall time	-	-	10	ns	Logic input edge speed requirement
$t_{PLH}$	Propagation delay when output is from low to high	-	6.5	-	ns	$C_{LOAD} = 1000pF$ , see timing diagram
$t_{PHL}$	Propagation delay when output is from high to low	-	6.5	-	ns	
$t_r$	Output rise time	-	7.0	-	ns	Input signal rise/fall time 2.0ns
$t_f$	Output fall time	-	7.0	-	ns	
$ t_r - t_f $	Rise and fall time matching	-	1.0	-	ns	For each channel
$ t_{PLH} - t_{PHL} $	Propagation low to high and high to low matching	-	1.0	-	ns	
$\Delta t_{dm}$	Propagation delay matching	-	$\pm 2.0$	-	ns	Device to device delay match
$t_{PE-ON}$	PE on-time	-	-	5.0	$\mu s$	$V_{PE} = 1.7 \sim 5.25V$ $V_{DD} = 7.5 \sim 11.5V$ $-20 \sim 85^\circ C$
$t_{PE-OFF}$	PE off-time	-	-	4.0	$\mu s$	
$f_{CLK}$	Clock input frequency	-	-	200	MHz	
$t_{RC}, t_{FC}$	Clock rise and fall timing	-	0.5	5.0	ns	
$t_{SU}$	D-FF setup time	2.0	-	-	ns	
$t_{H}$	D-FF holdup time	1.0	-	-	ns	

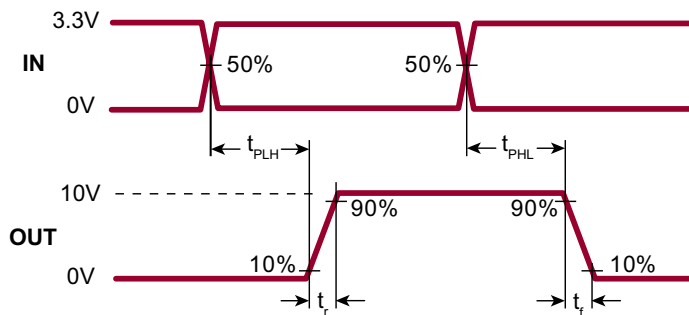
**Logic Truth Table**

Logic Inputs					Output				Notes
PE	BPS	CLK	INA	INB	OUTA	OUTB	OUTC	OUTD	
1	0	$\uparrow$	0	0	1	0	0	1	Output update on CLK rising edge
1	0	$\uparrow$	0	1	1	1	1	0	
1	0	$\uparrow$	1	0	0	0	1	0	
1	0	$\uparrow$	1	1	1	0	1	0	
1	1	X	0	0	1	0	0	1	D-FF register bypassed
1	1	X	0	1	1	1	1	0	
1	1	X	1	0	0	0	1	0	
1	1	X	1	1	1	0	1	0	
1	$\bar{1}$	X	X	X	Output asynchronous update				
$\uparrow$	0	X	X	X	1	0	1	0	D-FF async reset
0	X	X	X	X	1	0	1	0	Power down

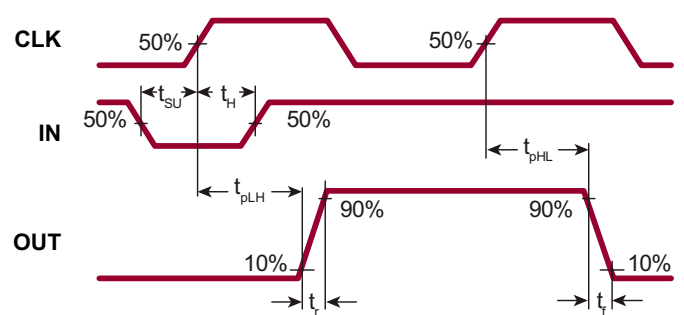
Block Diagram



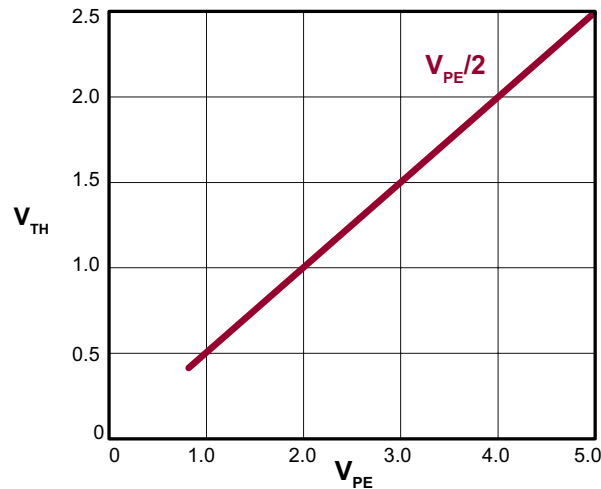
Timing Diagram (BPS = 1)



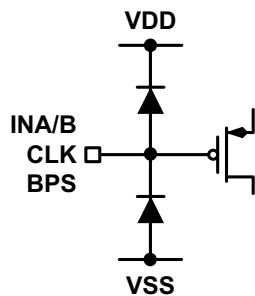
Timing Diagram (BPS = 0)



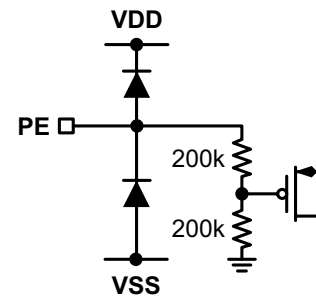
## $V_{TH}$ vs $V_{PE}$ Curve



## Equivalent Circuits



Equivalent Logic Input Circuit



Equivalent PE Input Circuit

## Application Information

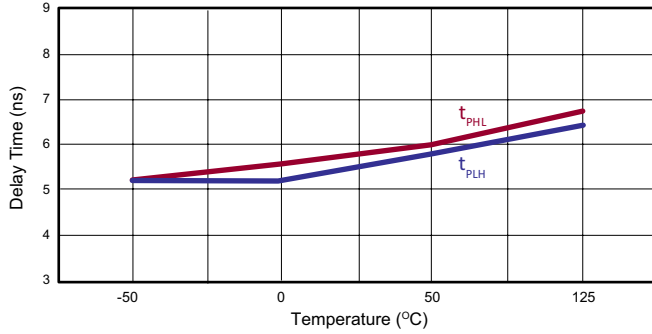
For proper operation of the MD1823, low inductance bypass capacitors should be used on the various supply pins. The GND pin should be connected to the logic ground. The INA, INB, CLK, BPS, and PE pins should be connected to a logic source with a swing of GND to PE, where PE is 1.8 to 5.0V. Good trace practices should be followed corresponding to the desired operating speed. The internal circuitry of the MD1823 is capable of operating up to 200MHz, with the primary speed limitation being the loading effects of the load capacitance. Because of this speed and the high transient currents that result with capacitive loads, the bypass capacitors should be as close to the chip pins as possible. Unless the load specifically requires bipolar drive, the VSS and VL pins should have low inductance feed-through connections directly to a ground plane. If these voltages are not zero, then they need bypass capacitors in a manner similar to the positive power supplies. The power connection VDD should have a ceramic bypass capacitor to the ground plane with short leads and decoupling components to prevent resonance in the power leads.

The voltages of VH and VL decide the output signal levels. These two pins can draw fast transient currents of up to 2.0A, so they should be provided with an appropriate bypass capacitor

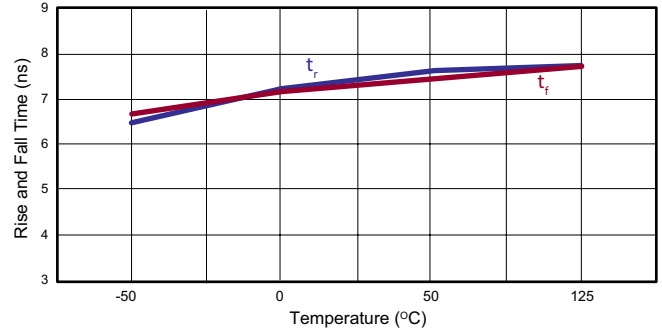
located next to the chip pins. A ceramic capacitor of up to 1.0 $\mu$ F may be appropriate, with a series ferrite bead to prevent resonance in the power supply lead coming to the capacitor. Pay particular attention to minimizing trace lengths, current loop area and using sufficient trace width to reduce inductance. Surface mount components are highly recommended. Since the output impedance of this driver is very low, in some cases it may be desirable to add a small series resistance in series with the output signal to obtain better waveform transitions at the load terminals. This will of course reduce the output voltage slew rate at the terminals of a capacitive load.

Pay particular attention that parasitic couplings are minimized from the output to the input signal terminals. The parasitic feedback may cause oscillations or spurious waveform shapes on the edges of signal transitions. Since the input operates with signals down to 1.8V even small coupled voltages may cause problems. Use of a solid ground plane and good power and signal layout practices will prevent this problem. Be careful that a circulating ground return current from a capacitive load cannot react with common inductance to cause noise voltages in the input logic circuitry.

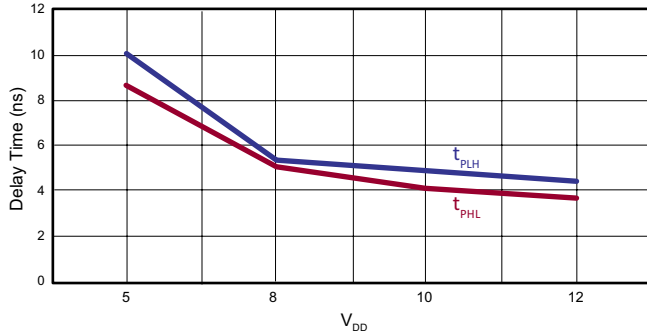
Delay Time vs Temperature



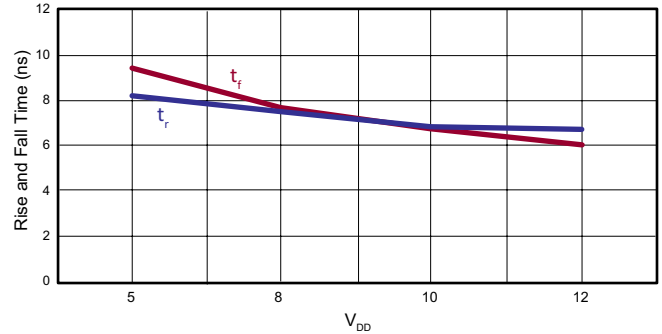
Rise and Fall Time vs Temperature



Delay Time vs V<sub>DD</sub>



Rise and Fall Time vs V<sub>DD</sub>

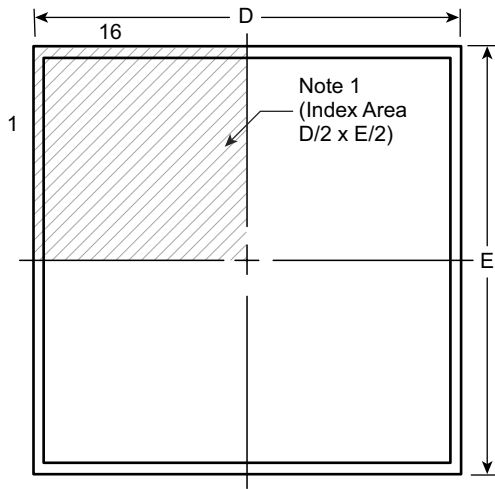


Pin Description

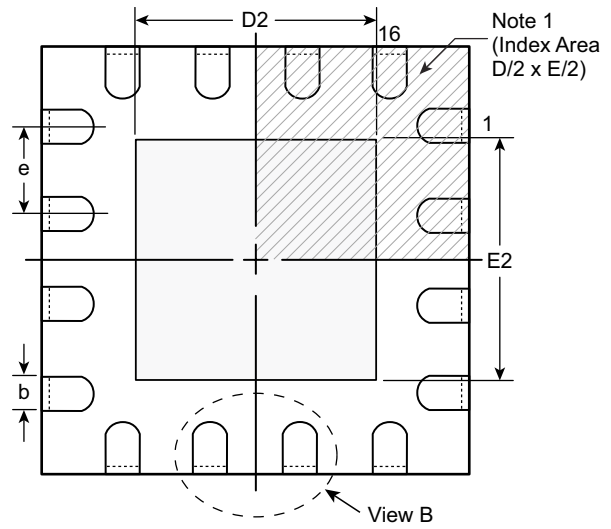
Pin #	Function	Description
1	INB	Logic input.
2	VDD	High side supply voltage.
3	VSS	Low side supply voltage. VSS is also connected to the IC substrate. It is required to connect to the most negative potential of voltage supplies.
4	CLK	D-FF Register clock input.
5	BPS	Logic input of D-FF registers bypass control, Hi = bypass, CLK will be ignored.
6	GND	Logic input ground reference.
7	VL	Supply voltage for N-channel output stage.
8	OUTC	Output drivers
9	OUTD	
10, 11	VH	Supply voltage for P-channel output stage.
12	OUTA	Output drivers
13	OUTB	
14	VL	Supply voltage for N-channel output stage.
15	PE	Enable logic input. When at logic high, the device is enabled and the input logic threshold is set. PE should go low when the complete system is in standby mode. When PE is low all outputs are at default state (see truth table).
16	INA	Logic input.
Substrate		The IC substrate is internally connected to the thermal pad. Thermal pad and VSS must be connected externally.

# 16-Lead QFN Package Outline (K6)

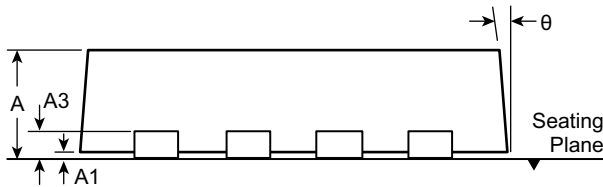
3.00x3.00mm body, 1.00mm height (max), 0.50mm pitch



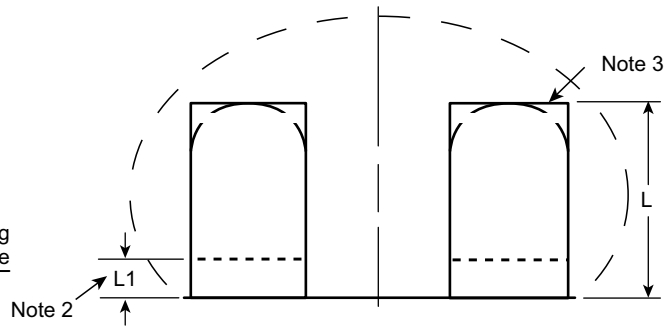
**Top View**



**Bottom View**



**Side View**



**View B**

**Notes:**

1. A Pin 1 identifier must be located in the index area indicated. The Pin 1 identifier can be: a molded mark/identifier; an embedded metal marker; or a printed indicator.
2. Depending on the method of manufacturing, a maximum of 0.15mm pullback (L1) may be present.
3. The inner tip of the lead may be either rounded or square.

Symbol		A	A1	A3	b	D	D2	E	E2	e	L	L1	$\theta$
Dimension (mm)	MIN	0.80	0.00	0.20 REF	0.18	2.85*	1.50	2.85*	1.50	0.50 BSC	0.20†	0.00	0°
	NOM	0.90	0.02		0.25	3.00	1.65	3.00	1.65		0.30†	-	-
	MAX	1.00	0.05		0.30	3.15*	1.80	3.15*	1.80		0.45	0.15	14°

JEDEC Registration MO-220, Variation VEED-4, Issue K, June 2006.

\* This dimension is not specified in the JEDEC drawing.

† This dimension differs from the JEDEC drawing.

**Drawings not to scale.**

**Supertex Doc.#:** DSPD-16QFNK63X3P050, Version A092909.

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information go to <http://www.supertex.com/packaging.html>.)

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