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June 2003

## FDB20AN06A0 / FDP20AN06A0

# N-Channel PowerTrench<sup>®</sup> MOSFET 60V, 45A, 20m $\Omega$

### **Features**

- $r_{DS(ON)} = 17m\Omega$  (Typ.),  $V_{GS} = 10V$ ,  $I_D = 45A$
- $Q_q(tot) = 15nC (Typ.), V_{GS} = 10V$
- Low Miller Charge
- Low Q<sub>RR</sub> Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)
- Qualified to AEC Q101

Formerly developmental type 82547

## **Applications**

- Motor / Body Load Control
- ABS Systems
- Powertrain Management
- Injection Systems
- DC-DC converters and Off-line UPS
- · Distributed Power Architectures and VRMs
- · Primary Switch for 12V and 24V systems







## MOSFET Maximum Ratings T<sub>C</sub> = 25°C unless otherwise noted

Symbol	Parameter	Ratings	Units
V <sub>DSS</sub>	Drain to Source Voltage	60	V
V <sub>GS</sub>	Gate to Source Voltage	±20	V
	Drain Current		
	Continuous ( $T_C = 25^{\circ}C$ , $V_{GS} = 10V$ )	45	Α
$I_D$	Continuous ( $T_C = 100^{\circ}$ C, $V_{GS} = 10V$ )	32	А
	Continuous ( $T_{amb} = 25^{\circ}C$ , $V_{GS} = 10V$ , $R_{\theta JA} = 43^{\circ}C/W$ )	9	А
	Pulsed	Figure 4	А
E <sub>AS</sub>	Single Pulse Avalanche Energy (Note 1)	50	mJ
P <sub>D</sub>	Power dissipation	90	W
	Derate above 25°C	0.60	W/°C
T <sub>J</sub> , T <sub>STG</sub>	Operating and Storage Temperature	-55 to 175	°C

## **Thermal Characteristics**

$R_{\theta JC}$	Thermal Resistance Junction to Case TO-220, TO-263	1.67	°C/W
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-220, TO-263 (Note 2)	62	°C/W
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-263, 1in <sup>2</sup> copper pad area	43	°C/W

This product has been designed to meet the extreme test conditions and environment demanded by the automotive industry. For a copy of the requirements, see AEC Q101 at: http://www.aecouncil.com/

Reliability data can be found at: http://www.fairchildsemi.com/products/discrete/reliability/index.html.

All Fairchild Semiconductor products are manufactured, assembled and tested under ISO9000 and QS9000 quality systems certification.

Device Marking Device		Package	Reel Size	Tape Width	Quantity
FDB20AN06A0	FDB20AN06A0	TO-263AB	330mm	24mm	800 units
FDP20AN06A0	FDP20AN06A0	TO-220AB	Tube	N/A	50 units

# **Electrical Characteristics** $T_C = 25^{\circ}C$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Тур	Max	Units
Off Char	racteristics					
B <sub>VDSS</sub>	Drain to Source Breakdown Voltage	$I_D = 250 \mu A, V_{GS} = 0 V$	60	-	-	V
	Zoro Coto Voltago Droin Current	V <sub>DS</sub> = 50V	-	-	1	^
I <sub>DSS</sub>	Zero Gate Voltage Drain Current	$V_{GS} = 0V$ $T_C = 150^{\circ}C$	-	-	250	μΑ
I <sub>GSS</sub>	Gate to Source Leakage Current	V <sub>GS</sub> = ±20V	-	-	±100	nA
On Char	acteristics					
V <sub>GS(TH)</sub>	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}, I_{D} = 250 \mu A$	2	-	4	V
		I <sub>D</sub> = 45A, V <sub>GS</sub> = 10V	-	0.017	0.020	
r <sub>DS(ON)</sub>	Drain to Source On Resistance	$I_D = 45A, V_{GS} = 10V,$ $T_J = 175$ °C	-	0.039	0.047	Ω
Dynamic	Characteristics		•			
C <sub>ISS</sub>	Input Capacitance		-	950	-	pF
C <sub>OSS</sub>	Output Capacitance	$V_{DS} = 25V, V_{GS} = 0V,$ $= 1MHz$	-	185	-	pF
C <sub>RSS</sub>	Reverse Transfer Capacitance	I = IMIHZ	-	60	-	pF
Q <sub>g(TOT)</sub>	Total Gate Charge at 10V	V <sub>GS</sub> = 0V to 10V		15	19	nC
Q <sub>g(TH)</sub>	Threshold Gate Charge	$V_{GS} = 0V \text{ to } 2V$ $V_{DD} = 30V$	-	2	2.6	nC
Q <sub>gs</sub>	Gate to Source Gate Charge	I <sub>D</sub> = 45A	-	6	-	nC
Q <sub>gs2</sub>	Gate Charge Threshold to Plateau	$I_g = 1.0 \text{mA}$	-	4	-	nC
$Q_{gd}$	Gate to Drain "Miller" Charge		-	4.5	-	nC
	ng Characteristics (V <sub>GS</sub> = 10V)					
t <sub>ON</sub>	Turn-On Time	1	-	-	164	ns
t <sub>d(ON)</sub>	Turn-On Delay Time		-	11	-	ns
t <sub>r</sub>	Rise Time	$V_{DD} = 30V, I_D = 45A$	-	98	-	ns
t <sub>d(OFF)</sub>	Turn-Off Delay Time	$V_{GS} = 10V$ , $R_{GS} = 20\Omega$	-	23	-	ns
t <sub>f</sub>	Fall Time		-	33	-	ns
t <sub>OFF</sub>	Turn-Off Time		-	-	84	ns
Drain-Sc	ource Diode Characteristics					
V <sub>SD</sub>	Source to Drain Diode Voltage	I <sub>SD</sub> = 45A	-	-	1.25	V

 $I_{SD} = 45A$ ,  $dI_{SD}/dt = 100A/\mu s$ 

 $I_{SD} = 45A$ ,  $dI_{SD}/dt = 100A/\mu s$ 

# $Q_{RR}$

Notes: 1: Starting  $T_J = 25^{\circ}C$ ,  $L = 80\mu H$ ,  $I_{AS} = 36A$ . 2: Pulse width = 100s.

Reverse Recovery Time

Reverse Recovered Charge

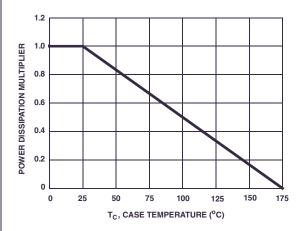
32

25

ns

nC





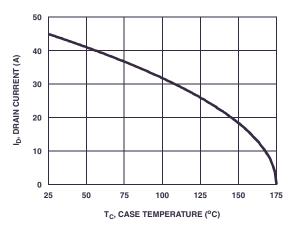


Figure 1. Normalized Power Dissipation vs
Ambient Temperature

Figure 2. Maximum Continuous Drain Current vs Case Temperature

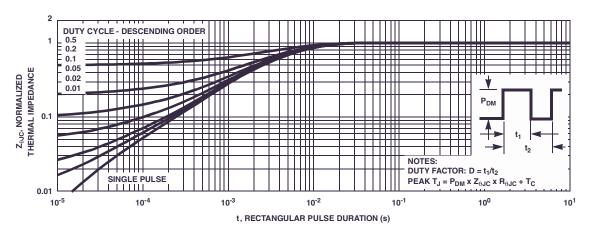


Figure 3. Normalized Maximum Transient Thermal Impedance

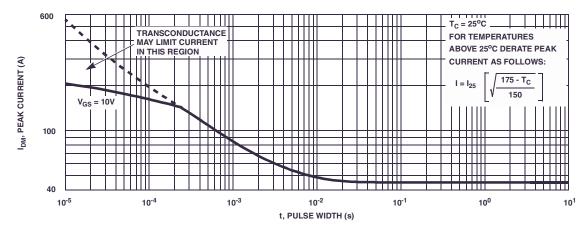


Figure 4. Peak Current Capability



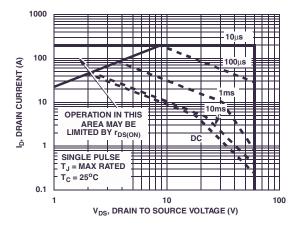
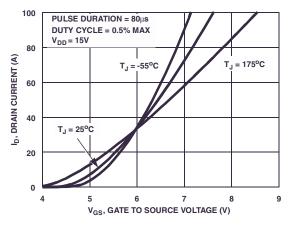


Figure 5. Forward Bias Safe Operating Area

NOTE: Refer to Fairchild Application Notes AN7514 and AN7515

Figure 6. Unclamped Inductive Switching

Capability



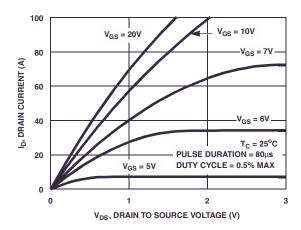
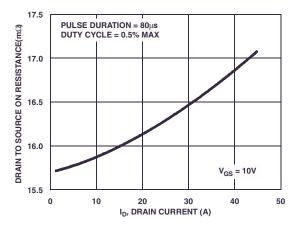


Figure 7. Transfer Characteristics

Figure 8. Saturation Characteristics



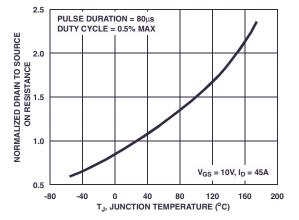


Figure 9. Drain to Source On Resistance vs Drain Current

Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature

# Typical Characteristics $T_C = 25^{\circ}C$ unless otherwise noted

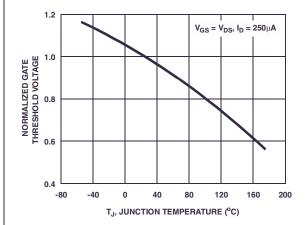


Figure 11. Normalized Gate Threshold Voltage vs Junction Temperature

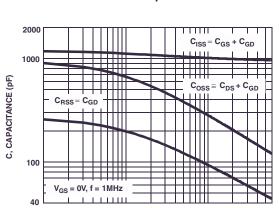


Figure 13. Capacitance vs Drain to Source Voltage

V<sub>DS</sub>, DRAIN TO SOURCE VOLTAGE (V)

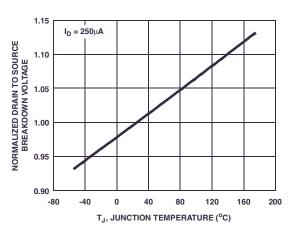


Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

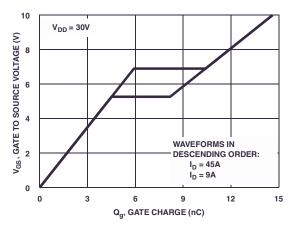


Figure 14. Gate Charge Waveforms for Constant Gate Current

## **Test Circuits and Waveforms**

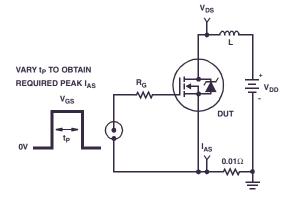


Figure 15. Unclamped Energy Test Circuit

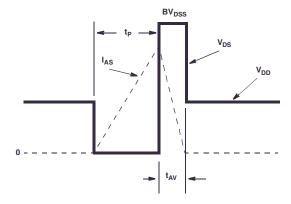


Figure 16. Unclamped Energy Waveforms

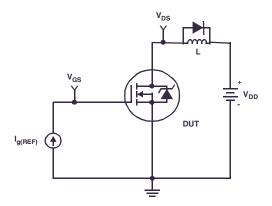


Figure 17. Gate Charge Test Circuit

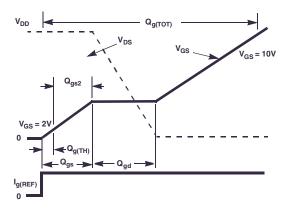


Figure 18. Gate Charge Waveforms

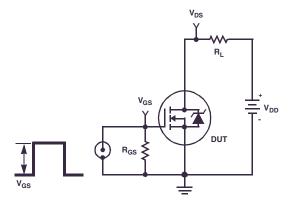


Figure 19. Switching Time Test Circuit

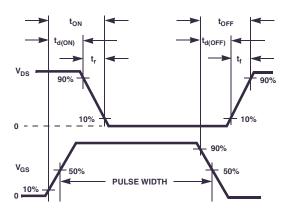


Figure 20. Switching Time Waveforms

## Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature,  $T_{JM}$ , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation,  $P_{DM}$ , in an application. Therefore the application's ambient temperature,  $T_A$  (°C), and thermal resistance  $R_{\theta JA}$  (°C/W) must be reviewed to ensure that  $T_{JM}$  is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \tag{EQ. 1}$$

In using surface mount devices such as the TO-263 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of  $\mathsf{P}_{\mathsf{DM}}$  is complex and influenced by many factors:

- Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
- 2. The number of copper layers and the thickness of the board.
- 3. The use of external heat sinks.
- 4. The use of thermal vias.
- 5. Air flow and board orientation.
- 6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

Fairchild provides thermal information to assist the designer's preliminary application evaluation. Figure 21 defines the  $R_{\theta JA}$  for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the Fairchild device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeters square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\theta JA} = 26.51 + \frac{19.84}{(0.262 + Area)}$$
 (EQ. 2)

Area in Inches Squared

$$R_{\theta JA} = 26.51 + \frac{128}{(1.69 + Area)}$$
 (EQ. 3)

Area in Centimeters Squared

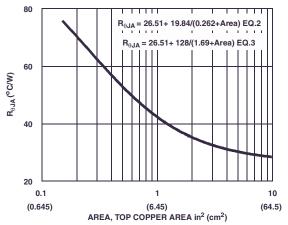


Figure 21. Thermal Resistance vs Mounting
Pad Area

#### **PSPICE Electrical Model** .SUBCKT FDP20AN06A0 2 1 3; rev April 2003 Ca 12 8 4.4e-10 Cb 15 14 4.4e-10 **LDRAIN DPLCAP** DRAIN Cin 6 8 9.2e-10 10 Dbody 7 5 DbodyMOD RLDRAIN ₹RSLC1 Dbreak 5 11 DbreakMOD DBREAK T Dplcap 10 5 DplcapMOD RSLC<sub>2</sub> <u>5</u> **ESLC** Ebreak 11 7 17 18 67.2 50 Eds 14 8 5 8 1 Egs 13 8 6 8 1 DBODY RDRAIN **EBREAK ESG** Esg 6 10 6 8 1 **EVTHRES** Evthres 6 21 19 8 1 $\left(\frac{19}{8}\right)$ Evtemp 20 6 18 22 1 MWFAK **LGATE EVTEMP** GATE **RGATE** -m **◆**MMED It 8 17 1 9 20 MSTRO BLGATE Lgate 1 9 6.3e-9 LSOURCE CIN SOURCE Ldrain 2 5 1.0e-9 Lsource 3 7 2.5e-9 RSOURCE RLSOURCE **RLgate 1 9 63** S1A **RBREAK** RLdrain 2 5 10 <u>13</u> 8 <u>14</u> 13 RLsource 3 7 25 **≨**RVTEMP o S2B S<sub>1</sub>B Mmed 16 6 8 8 MmedMOD CE 19 CA Mstro 16 6 8 8 MstroMOD IT 14 Mweak 16 21 8 8 MweakMOD VBAT EGS **EDS** Rbreak 17 18 RbreakMOD 1 8 Rdrain 50 16 Rdrain MOD 1e-3 **RVTHRES** Rgate 9 20 4.7 RŠLC1 5 51 RSLCMOD 1e-6 RSLC2 5 50 1e3 Rsource 8 7 RsourceMOD 10e-3 Rvthres 22 8 RvthresMOD 1 Rvtemp 18 19 RvtempMOD 1 S1a 6 12 13 8 S1AMOD S1b 13 12 13 8 S1BMOD S2a 6 15 14 13 S2AMOD S2b 13 15 14 13 S2BMOD Vbat 22 19 DC 1 ESLC 51 50 VALUE={(V(5,51)/ABS(V(5,51)))\*(PWR(V(5,51)/(1e-6\*150),2.7))} .MODEL DbodyMOD D (IS=3.8E-12 N=1.06 RS=4e-3 TRS1=2.4e-3 TRS2=1.1e-6 + CJO=6.8e-10 M=0.53 TT=2.3e-8 XTI=3.9) .MODEL DbreakMOD D (RS=1.8 TRS1=1e-3 TRS2=-8.9e-6) .MODEL DplcapMOD D (CJO=2.7e-10 IS=1e-30 N=10 M=0.44) .MODEL MmedMOD NMOS (VTO=3.8 KP=2 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=4.7 T ABS=25) .MODEL MstroMOD NMOS (VTO=4.34 KP=35 IS=1e-30 N=10 TOX=1 L=1u W=1u T\_ABS=25) .MODEL MweakMOD NMOS (VTO=3.27 KP=0.03 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=47 RS=0.1 T\_ABS=25) .MODEL RbreakMOD RES (TC1=9e-4 TC2=1e-7) .MODEL RdrainMOD RES (TC1=6e-3 TC2=8e-5) .MODEL RSLCMOD RES (TC1=1e-3 TC2=3.5e-5) .MODEL RsourceMOD RES (TC1=9e-3 TC2=1e-6) .MODEL RvthresMOD RES (TC1=-5.1e-3 TC2=-1.3e-5) .MODEL RytempMOD RES (TC1=-3e-3 TC2=1e-7) .MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-8 VOFF=-5) .MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-5 VOFF=-8) .MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-2 VOFF=-1.5) .MODEL S2BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-1.5 VOFF=-2) Note: For further discussion of the PSPICE model, consult A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global

Temperature Options; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank

Wheatley.

#### SABER Electrical Model rev April 2003 template FDP20AN06A0 n2,n1,n3 =m temp electrical n2,n1,n3 number m\_temp=25 var i iscl dp.,model dbodymod = (isl=3.8e-12.nl=1.06.rs=4e-3.trs1=2.4e-3.trs2=1.1e-6.cio=6.8e-10.m=0.53.tt=2.3e-8.xti=3.9) dp..model dbreakmod = (rs=1.8,trs1=1e-3,trs2=-8.9e-6) dp..model dplcapmod = (cjo=2.7e-10,isl=10e-30,nl=10,m=0.44) m..model mmedmod = (type=\_n,vto=3.8,kp=2,is=1e-30, tox=1) m..model mstrongmod = (type= n, vto=4.34, kp=35, is=1e-30, tox=1)m..model mweakmod = $(type=_n, vto=3.27, kp=0.03, is=1e-30, tox=1, rs=0.1)$ **LDRAIN** sw\_vcsp..model s1amod = (ron=1e-5,roff=0.1,von=-8,voff=-5) DPLCAP DRAIN sw\_vcsp..model s1bmod = (ron=1e-5,roff=0.1,von=-5,voff=-8) 10 sw\_vcsp..model s2amod = (ron=1e-5,roff=0.1,von=-2,voff=-1.5) BLDBAIN €RSLC1 sw\_vcsp..model s2bmod = (ron=1e-5,roff=0.1,von=-1.5,voff=-2) c.ca n12 n8 = 4.4e-10RSLC2 ₹ c.cb n15 n14 = 4.4e-10ISCL c.cin n6 n8 = 9.2e-10 DBREAK I dp.dbody n7 n5 = model=dbodymod **≨**RDRAIN <u>6</u> ESG( 11 dp.dbreak n5 n11 = model=dbreakmod DBODY **EVTHRES** dp.dplcap n10 n5 = model=dplcapmod 16 (<u>19</u>) 8 MWEAK **EVTEMP LGATE** spe.ebreak n11 n7 n17 n18 = 67.2 GATE **RGATE** MMED EBREAK $\frac{1}{100}$ spe.eds n14 n8 n5 n8 = 1 20 MSTRO spe.eqs n13 n8 n6 n8 = 1 **RLGATE** spe.esg n6 n10 n6 n8 = 1 LSOURCE CIN SOURCE spe.evthres n6 n21 n19 n8 = 1 spe.evtemp n20 n6 n18 n22 = 1 **RSOURCE** RLSOURCE i.it n8 n17 = 1**RBREAK** 14 13 <u>13</u> 8 17 I.lgate n1 n9 = 6.3e-9I.Idrain n2 n5 = 1.0e-9RVTEMP S1B o S2B I.lsource n3 n7 = 2.5e-9СВ 19 CA IT 14 res.rlgate n1 n9 = 63 VBAT EGS EDS res.rldrain n2 n5 = 10res.rlsource n3 n7 = 25 **RVTHRES** m.mmed n16 n6 n8 n8 = model=mmedmod, l=1u, w=1u, temp=m\_temp m.mstrong n16 n6 n8 n8 = model=mstrongmod, l=1u, w=1u ,temp=m\_temp m.mweak n16 n21 n8 n8 = model=mweakmod, l=1u, w=1u ,temp=m\_temp res.rbreak n17 n18 = 1, tc1=9e-4,tc2=1e-7 res.rdrain n50 n16 = 1e-3, tc1=6e-3,tc2=8e-5 res.rgate n9 n20 = 4.7res.rslc1 n5 n51 = 1e-6, tc1=1e-3,tc2=3.5e-5 res.rslc2 n5 n50 = 1e3res.rsource n8 n7 = 10e-3, tc1=9e-3,tc2=1e-6 res.rvthres n22 n8 = 1, tc1=-5.1e-3, tc2=-1.3e-5res.rvtemp n18 n19 = 1. tc1=-3e-3.tc2=1e-7sw\_vcsp.s1a n6 n12 n13 n8 = model=s1amod sw\_vcsp.s1b n13 n12 n13 n8 = model=s1bmod sw\_vcsp.s2a n6 n15 n14 n13 = model=s2amod sw\_vcsp.s2b n13 n15 n14 n13 = model=s2bmod v.vbat n22 n19 = dc=1 equations { i (n51->n50) +=iscl iscl: v(n51,n50) = ((v(n5,n51)/(1e-9+abs(v(n5,n51))))\*((abs(v(n5,n51)\*1e6/150))\*\* 2.7))

# **PSPICE Thermal Model**

**REV 23 April 2003** 

FDP20AN06A0T

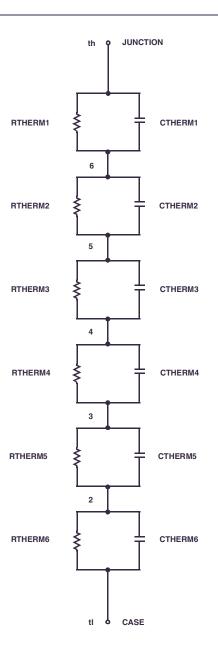
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RTHERM1 TH 6 3.0e-2 RTHERM2 6 5 1.0e-1 RTHERM3 5 4 1.4e-1 RTHERM4 4 3 2.3e-1 RTHERM5 3 2 4.1e-1 RTHERM6 2 TL 4.2e-1

## SABER Thermal Model

SABER thermal model FDP20AN06A0T template thermal\_model th tl thermal\_c th, tl { therm.ctherm1 th 6 = 1.8e-3 ctherm.ctherm2 6 5 = 8.0e-3 ctherm.ctherm3 5 4 = 9.0e-3 ctherm.ctherm4 4 3 = 1.1e-2 ctherm.ctherm5 3 2 = 1.2e-2 ctherm.ctherm6 2 tl = 2.0e-2 rtherm.rtherm1 th 6 = 3.0e-2 rtherm.rtherm2 6 5 = 1.0e-1 rtherm.rtherm3 5 4 = 1.4e-1 rtherm.rtherm4 4 3 = 2.3e-1

rtherm.rtherm5 3 2 =4.1e-1 rtherm.rtherm6 2 tl =4.2e-1



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ActiveArray™	FACT Quiet Series™	ISOPLANAR™	POP™	Stealth™
Bottomless™	FAST®	LittleFET™	Power247™	SuperSOT™-3
CoolFET™	FASTr™	MicroFET™	PowerTrench®	SuperSOT™-6
CROSSVOLT™	FRFET™	MicroPak™	QFET®	SuperSOT™-8
DOME™	GlobalOptoisolator™	MICROWIRE™	QS™	SyncFET™
EcoSPARK™	GTO™ .	MSX™	QT Optoelectronics™	TinyLogic <sup>®</sup>
E <sup>2</sup> CMOS™	HiSeC™	MSXPro™	Quiet Series™	TruTranslation™
EnSigna™	I <sup>2</sup> C <sup>TM</sup>	$OCX^{TM}$	RapidConfigure™	UHC™
Across the board.	Around the world.™	OCXPro™	RapidConnect™	UltraFET <sup>®</sup>
The Power Franch	nise™	OPTOLOGIC®	SILENT SWITCHER®	VCX™
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