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## FDD2572 / FDU2572

### N-Channel PowerTrench<sup>®</sup> MOSFET 150V, 29A, 54m $\Omega$

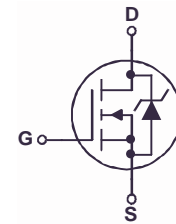
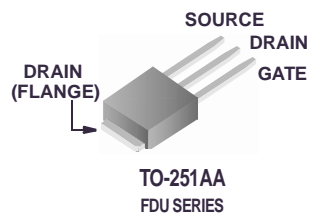
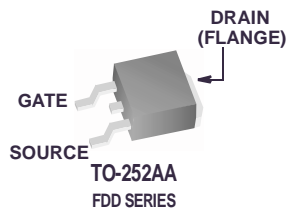
#### Features

- $r_{DS(ON)} = 45\text{m}\Omega$  (Typ.),  $V_{GS} = 10\text{V}$ ,  $I_D = 9\text{A}$
- $Q_g(\text{tot}) = 26\text{nC}$  (Typ.),  $V_{GS} = 10\text{V}$
- Low Miller Charge
- Low  $R_R$  Body Diode
- UIS Capability (Single Pulse and Repetitive Pulse)

#### Applications

- DC/DC converters and Off-Line UPS
- Distributed Power Architectures and VRMs
- Primary Switch for 24V and 48V Systems
- High Voltage Synchronous Rectifier

Formerly developmental type 82860



#### MOSFET Maximum Ratings $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Ratings	Units
$V_{DS}$	Drain to Source Voltage	150	V
$V_{GS}$	Gate to Source Voltage	$\pm 20$	V
$I_D$	Drain Current		
	Continuous ( $T_C = 25^\circ\text{C}$ , $V_{GS} = 10\text{V}$ )	29	A
	Continuous ( $T_C = 100^\circ\text{C}$ , $V_{GS} = 10\text{V}$ )	20	A
	Continuous ( $T_{\text{amb}} = 25^\circ\text{C}$ , $V_{GS} = 10\text{V}$ , $R_{\theta JA} = 52^\circ\text{C/W}$ )	4	
	Pulsed	Figure 4	A
$E_{AS}$	Single Pulse Avalanche Energy (Note 1)	36	mJ
$P_D$	Power dissipation	135	W
	Derate above $25^\circ\text{C}$	0.9	W/ $^\circ\text{C}$
$T_J, T_{STG}$	Operating and Storage Temperature	-55 to 175	$^\circ\text{C}$

#### Thermal Characteristics

$R_{\theta JC}$	Thermal Resistance Junction to Case TO-251, TO-252	1.11	$^\circ\text{C/W}$
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-251, TO-252	100	$^\circ\text{C/W}$
$R_{\theta JA}$	Thermal Resistance Junction to Ambient TO-252, 1in <sup>2</sup> copper pad area	52	$^\circ\text{C/W}$

## Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FDD2572	FDD2572	TO-252AA	330mm	16mm	2500 units
FDU2572	FDU2572	TO-251AA	Tube	N/A	75 units

## Electrical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

Symbol	Parameter	Test Conditions	Min	Typ	Max	Units
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### Off Characteristics

$B_{VDSS}$	Drain to Source Breakdown Voltage	$I_D = 250\mu\text{A}$ , $V_{GS} = 0\text{V}$	150	-	-	V
$I_{DSS}$	Zero Gate Voltage Drain Current	$V_{DS} = 120\text{V}$ $V_{GS} = 0\text{V}$ $T_C = 150^\circ$	-	-	1	$\mu\text{A}$
$I_{GSS}$	Gate to Source Leakage Current	$V_{GS} = \pm 20\text{V}$	-	-	$\pm 100$	nA

### On Characteristics

$V_{GS(TH)}$	Gate to Source Threshold Voltage	$V_{GS} = V_{DS}$ , $I_D = 250\mu\text{A}$	2	-	4	V
$r_{DS(ON)}$	Drain to Source On Resistance	$I_D = 9\text{A}$ , $V_{GS} = 10\text{V}$	-	0.045	0.054	$\Omega$
		$I_D = 4\text{A}$ , $V_{GS} = 6\text{V}$ ,	-	0.050	0.075	
		$I_D = 9\text{A}$ , $V_{GS} = 10\text{V}$ , $T_C = 175^\circ\text{C}$	-	0.126	0.146	

### Dynamic Characteristics

$C_{ISS}$	Input Capacitance	$V_{DS} = 25\text{V}$ , $V_{GS} = 0\text{V}$ , $f = 1\text{MHz}$	-	1770	-	pF
$C_{OSS}$	Output Capacitance		-	183	-	pF
$C_{RSS}$	Reverse Transfer Capacitance		-	40	-	pF
$Q_{g(TOT)}$	Total Gate Charge at 10V	$V_{GS} = 0\text{V}$ to 10V	-	26	34	nC
$Q_{g(TH)}$	Threshold Gate Charge	$V_{GS} = 0\text{V}$ to 2V	-	3.3	4.3	nC
$Q_{gs}$	Gate to Source Gate Charge	$V_{DD} = 75\text{V}$ $I_D = 9\text{A}$ $I_g = 1.0\text{mA}$	-	8	-	nC
$Q_{gs2}$	Gate Charge Threshold to Plateau		-	5	-	nC
$Q_{gd}$	Gate to Drain "Miller" Charge		-	6	-	nC

### Resistive Switching Characteristics ( $V_{GS} = 10\text{V}$ )

$t_{ON}$	Turn-On Time	$V_{DD} = 75\text{V}$ , $I_D = 9\text{A}$ $V_{GS} = 10\text{V}$ , $R_{GS} = 11.0\Omega$	-	-	36	ns
$t_{d(ON)}$	Turn-On Delay Time		-	11	-	ns
$t_r$	Rise Time		-	14	-	ns
$t_{d(OFF)}$	Turn-Off Delay Time		-	31	-	ns
$t_f$	Fall Time		-	14	-	ns
$t_{OFF}$	Turn-Off Time		-	-	66	ns

### Drain-Source Diode Characteristics

$V_{SD}$	Source to Drain Diode Voltage	$I_{SD} = 9\text{A}$	-	-	1.25	V
		$I_{SD} = 4\text{A}$	-	-	1.0	V
$t_{rr}$	Reverse Recovery Time	$I_{SD} = 9\text{A}$ , $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	74	ns
$Q_{RR}$	Reverse Recovered Charge	$I_{SD} = 9\text{A}$ , $dI_{SD}/dt = 100\text{A}/\mu\text{s}$	-	-	169	nC

#### Notes:

1: Starting  $T_J = 25^\circ\text{C}$ ,  $L = 0.2\text{mH}$ ,  $I_{AS} = 19\text{A}$ .

# Typical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

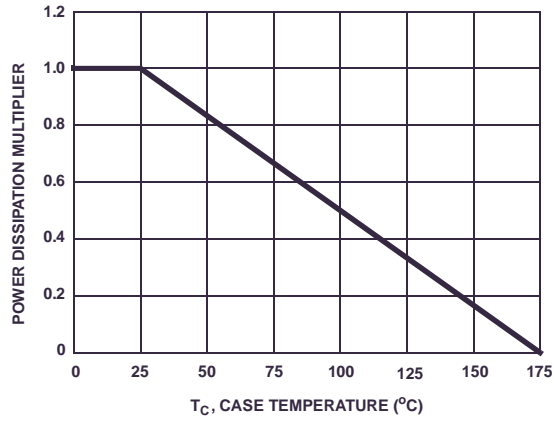


Figure 1. Normalized Power Dissipation vs Ambient Temperature

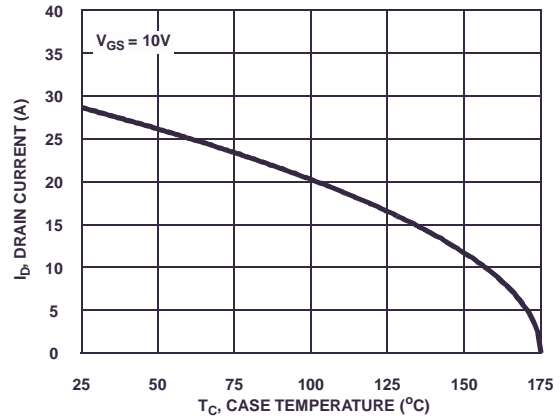


Figure 2. Maximum Continuous Drain Current vs Case Temperature

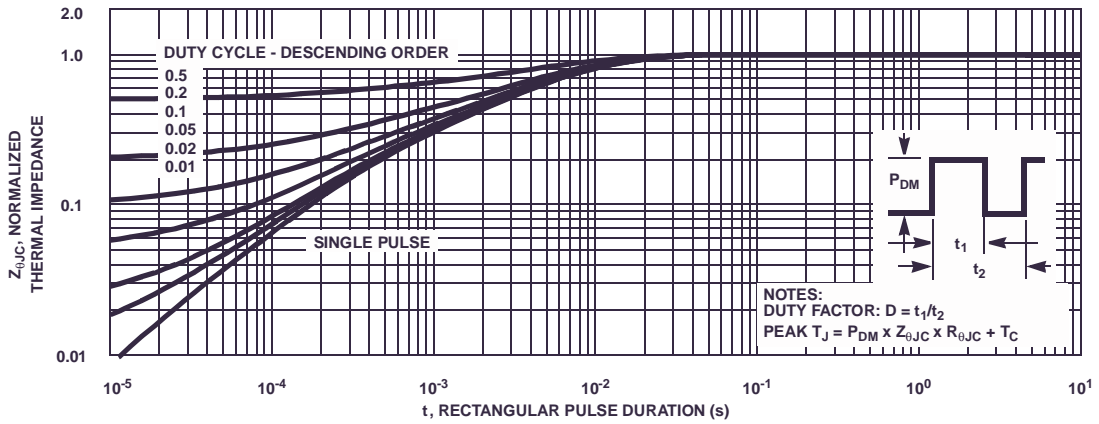


Figure 3. Normalized Maximum Transient Thermal Impedance

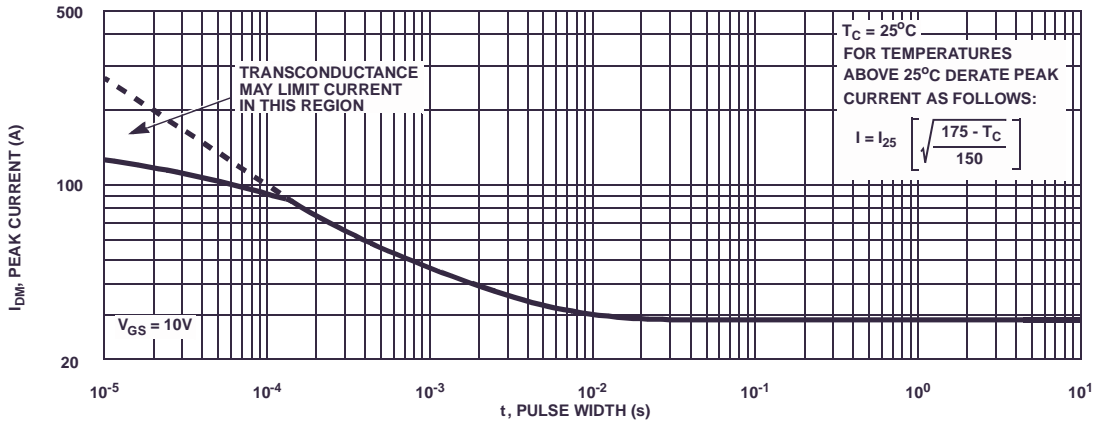


Figure 4. Peak Current Capability

## Typical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

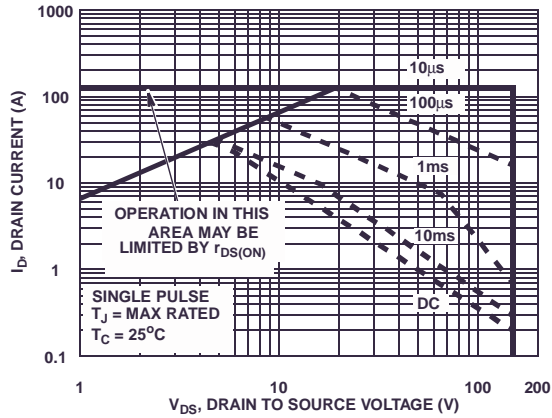
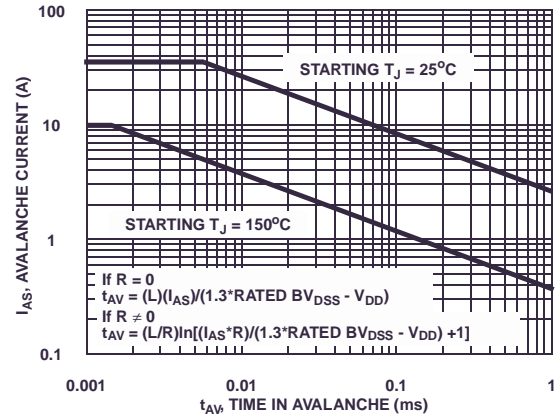


Figure 5. Forward Bias Safe Operating Area



NOTE: Refer to Fairchild Application Notes AN7514 and AN7515  
Figure 6. Unclamped Inductive Switching Capability

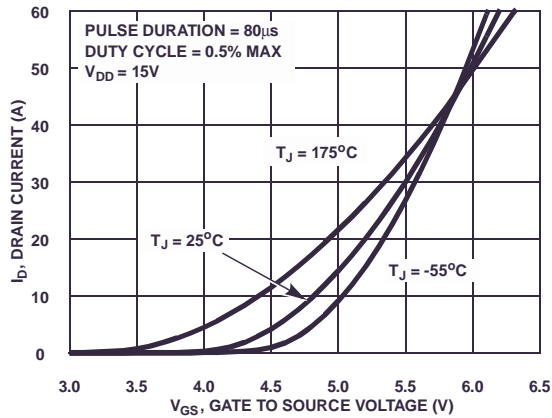


Figure 7. Transfer Characteristics

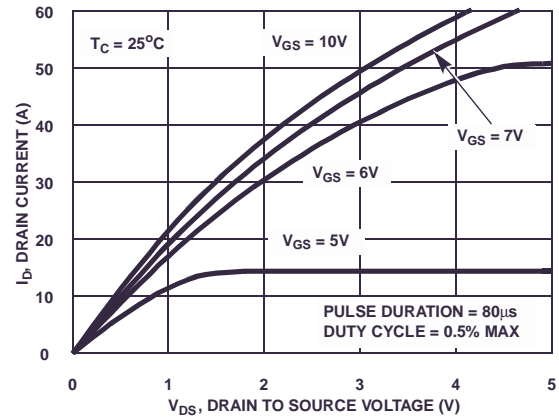


Figure 8. Saturation Characteristics

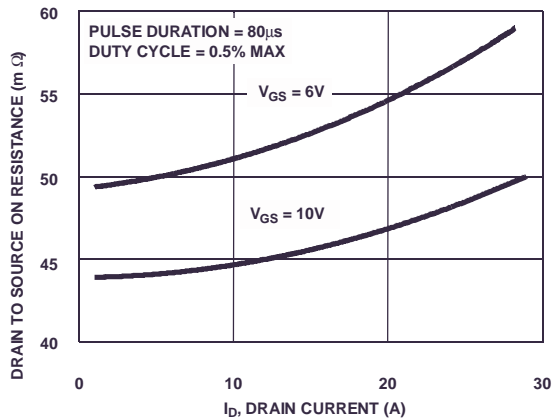


Figure 9. Drain to Source On Resistance vs Drain Current

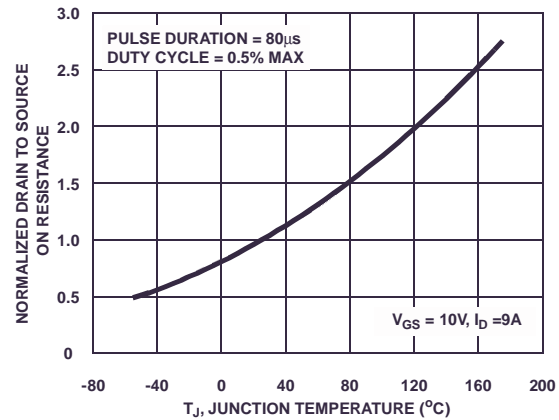


Figure 10. Normalized Drain to Source On Resistance vs Junction Temperature

# Typical Characteristics $T_C = 25^\circ\text{C}$ unless otherwise noted

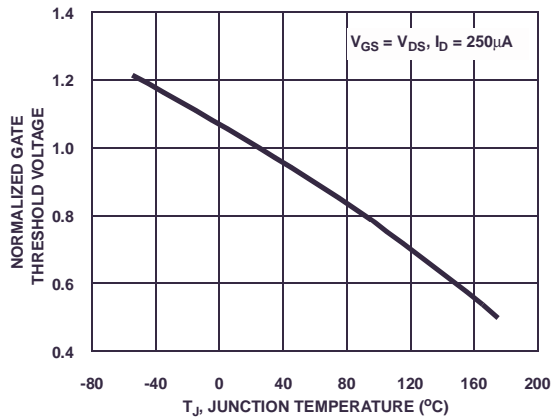


Figure 11. Normalized Gate Threshold Voltage vs Junction Temperature

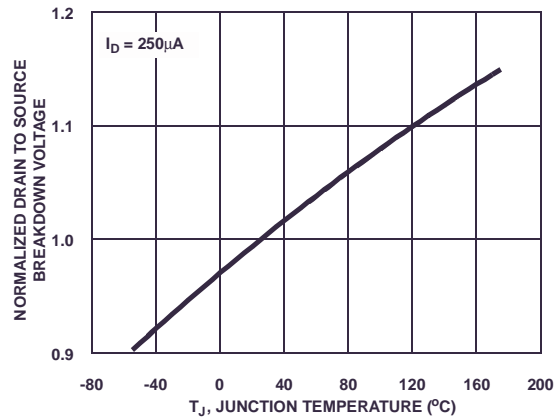


Figure 12. Normalized Drain to Source Breakdown Voltage vs Junction Temperature

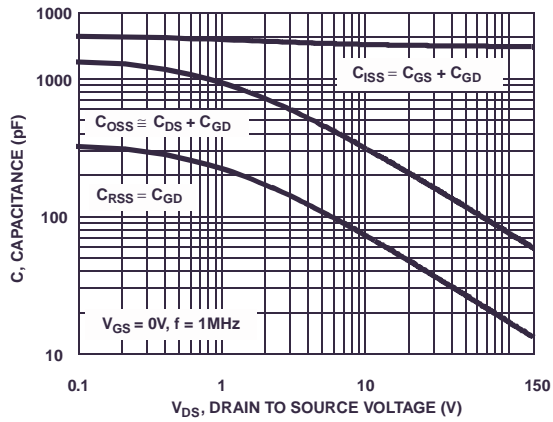


Figure 13. Capacitance vs Drain to Source Voltage

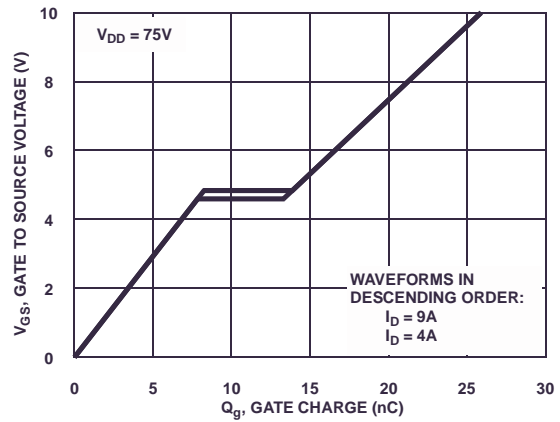


Figure 14. Gate Charge Waveforms for Constant Gate Currents

## Test Circuits and Waveforms

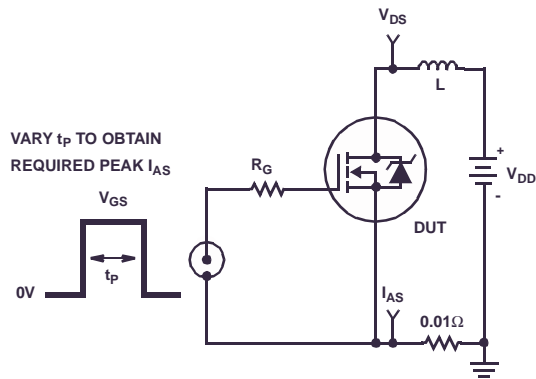


Figure 15. Unclamped Energy Test Circuit

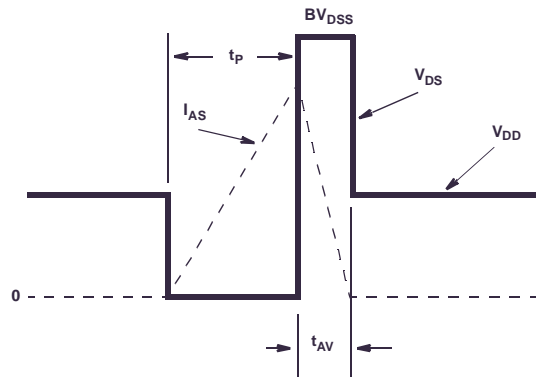


Figure 16. Unclamped Energy Waveforms

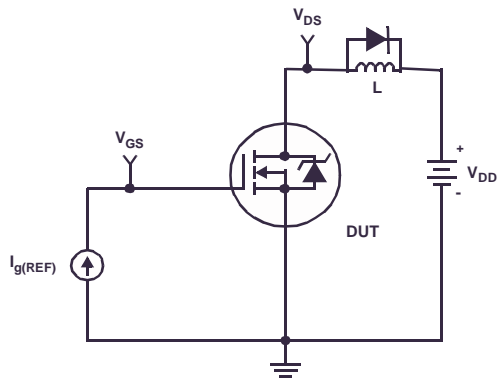


Figure 17. Gate Charge Test Circuit

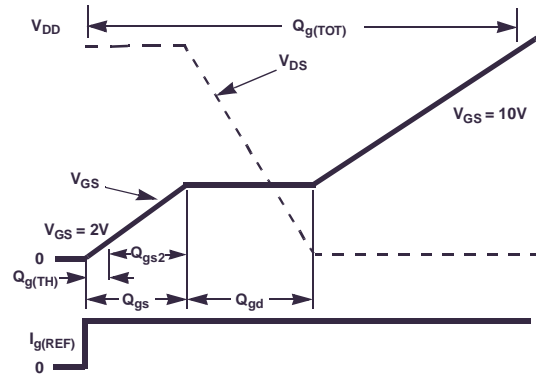


Figure 18. Gate Charge Waveforms

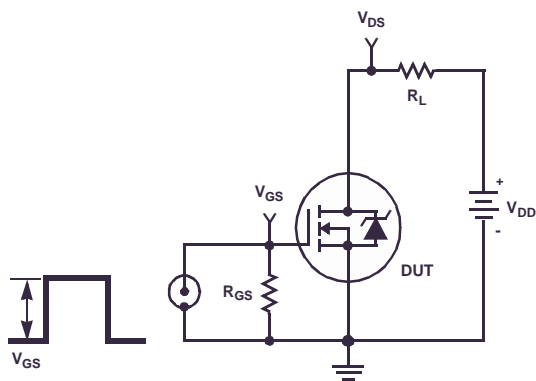


Figure 19. Switching Time Test Circuit

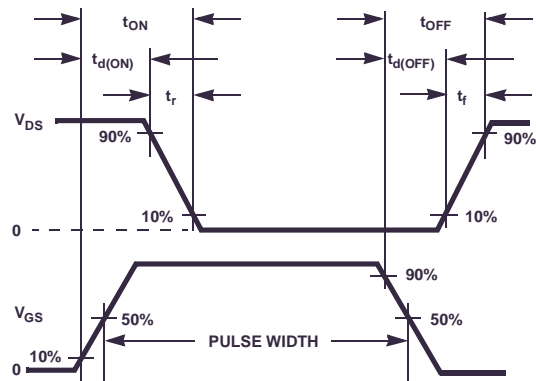


Figure 20. Switching Time Waveforms

## Thermal Resistance vs. Mounting Pad Area

The maximum rated junction temperature,  $T_{JM}$ , and the thermal resistance of the heat dissipating path determines the maximum allowable device power dissipation,  $P_{DM}$ , in an application. Therefore the application's ambient temperature,  $T_A$  ( $^{\circ}\text{C}$ ), and thermal resistance  $R_{\theta JA}$  ( $^{\circ}\text{C/W}$ ) must be reviewed to ensure that  $T_{JM}$  is never exceeded. Equation 1 mathematically represents the relationship and serves as the basis for establishing the rating of the part.

$$P_{DM} = \frac{(T_{JM} - T_A)}{R_{\theta JA}} \quad (\text{EQ. 1})$$

In using surface mount devices such as the TO-252 package, the environment in which it is applied will have a significant influence on the part's current and maximum power dissipation ratings. Precise determination of  $P_{DM}$  is complex and influenced by many factors:

1. Mounting pad area onto which the device is attached and whether there is copper on one side or both sides of the board.
2. The number of copper layers and the thickness of the board.
3. The use of external heat sinks.
4. The use of thermal vias.
5. Air flow and board orientation.
6. For non steady state applications, the pulse width, the duty cycle and the transient thermal response of the part, the board and the environment they are in.

Fairchild provides thermal information to assist the designer's preliminary application evaluation. Figure 21 defines the  $R_{\theta JA}$  for the device as a function of the top copper (component side) area. This is for a horizontally positioned FR-4 board with 1oz copper after 1000 seconds of steady state power with no air flow. This graph provides the necessary information for calculation of the steady state junction temperature or power dissipation. Pulse applications can be evaluated using the Fairchild device Spice thermal model or manually utilizing the normalized maximum transient thermal impedance curve.

Thermal resistances corresponding to other copper areas can be obtained from Figure 21 or by calculation using Equation 2 or 3. Equation 2 is used for copper area defined in inches square and equation 3 is for area in centimeter square. The area, in square inches or square centimeters is the top copper area including the gate and source pads.

$$R_{\theta JA} = 33.32 + \frac{23.84}{(0.268 + \text{Area})} \quad (\text{EQ. 2})$$

Area in Inches Squared

$$R_{\theta JA} = 33.32 + \frac{154}{(1.73 + \text{Area})} \quad (\text{EQ. 3})$$

Area in Centimeters Squared

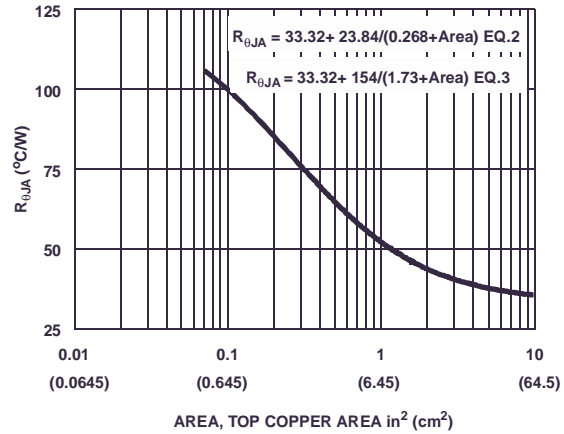


Figure 21. Thermal Resistance vs Mounting Pad Area



## PSPICE Electrical Model

.SUBCKT FDD2572 2 1 3 ; rev April 2002

CA 12 8 5.5e-10

Cb 15 14 7.4e-10

Cin 6 8 1.7e-9

Dbody 7 5 DbodyMOD  
Dbreak 5 11 DbreakMOD  
Dplcap 10 5 DplcapMOD

Ebreak 11 7 17 18 160  
Eds 14 8 5 8 1  
Egs 13 8 6 8 1  
Esg 6 10 6 8 1  
Evthres 6 21 19 8 1  
Etemp 20 6 18 22 1

It 8 17 1

Lgate 1 9 121e-9  
Ldrain 2 5 1.0e-9  
Lsource 3 7 4.45e-9

RLgate 1 9 12.1  
RLdrain 2 5 10  
RLsource 3 7 44.5

Mmed 16 6 8 8 MmedMOD  
Mstro 16 6 8 8 MstroMOD  
Mweak 16 21 8 8 MweakMOD

Rbreak 17 18 RbreakMOD 1  
Rdrain 50 16 RdrainMOD 35e-3  
Rgate 9 20 1.6  
RSLC1 5 51 RSLCMOD 1.0e-6  
RSLC2 5 50 1.0e3  
Rsource 8 7 RsourceMOD 3.0e-3  
Rvthres 22 8 RvthresMOD 1  
Rvtemp 18 19 RvtempMOD 1  
S1a 6 12 13 8 S1AMOD  
S1b 13 12 13 8 S1BMOD  
S2a 6 15 14 13 S2AMOD  
S2b 13 15 14 13 S2BMOD

Vbat 22 19 DC 1

ESLC 51 50 VALUE={{(V(5,51)/ABS(V(5,51)))\*(PWR(V(5,51)/(1e-6\*52),3))}}

.MODEL DbodyMOD D (IS=6.0E-11 N=1.14 RS=3.9e-3 TRS1=3.5e-3 TRS2=3.0e-6  
+ CJO=1.1e-9 M=0.63 TT=6.2e-8 XTI=4.5)

.MODEL DbreakMOD D (RS=10 TRS1=5.0e-3 TRS2=-5.0e-6)

.MODEL DplcapMOD D (CJO=3.5e-10 IS=1.0e-30 N=10 M=0.65)

.MODEL MmedMOD NMOS (VTO=3.55 KP=3 IS=1e-40 N=10 TOX=1 L=1u W=1u RG=1.6)

.MODEL MstroMOD NMOS (VTO=4.0 KP=25 IS=1e-30 N=10 TOX=1 L=1u W=1u)

.MODEL MweakMOD NMOS (VTO=2.95 KP=0.05 IS=1e-30 N=10 TOX=1 L=1u W=1u RG=16 RS=0.1)

.MODEL RbreakMOD RES (TC1=1.15e-3 TC2=-9.5e-7)

.MODEL RdrainMOD RES (TC1=9.0e-3 TC2=2.5e-5)

.MODEL RSLCMOD RES (TC1=3.0e-3 TC2=2.5e-6)

.MODEL RsourceMOD RES (TC1=4.0e-3 TC2=1.0e-6)

.MODEL RvthresMOD RES (TC1=-4.1e-3 TC2=-1.0e-5)

.MODEL RvtempMOD RES (TC1=-4.0e-3 TC2=1.0e-6)

.MODEL S1AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-5.0 VOFF=-3.5)

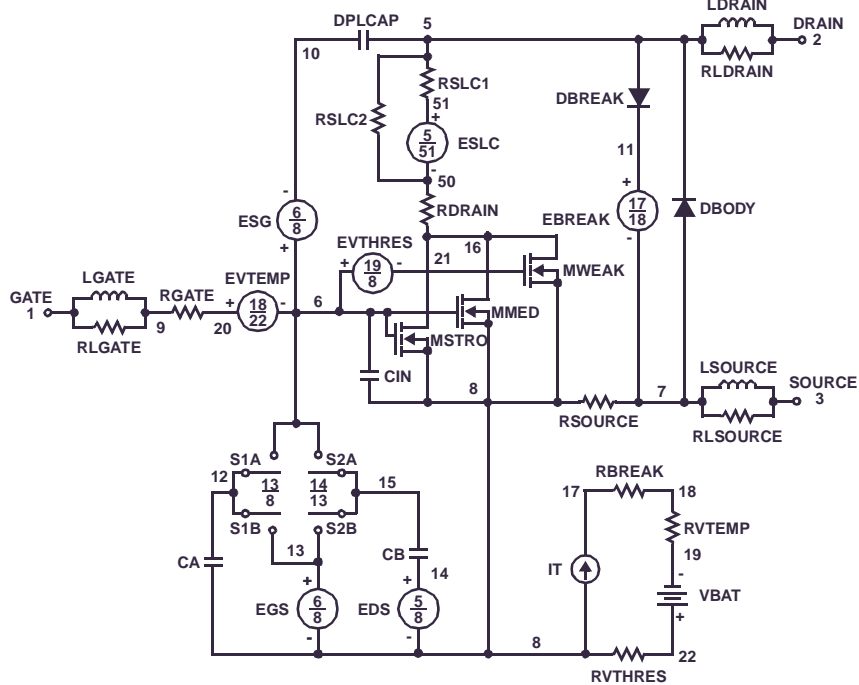
.MODEL S1BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-3.5 VOFF=-5.0)

.MODEL S2AMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=-0.5 VOFF=0.3)

.MODEL S2BMOD VSWITCH (RON=1e-5 ROFF=0.1 VON=0.3 VOFF=-0.5)

.ENDS

Note: For further discussion of the PSPICE model, consult **A New PSPICE Sub-Circuit for the Power MOSFET Featuring Global Temperature Options**; IEEE Power Electronics Specialist Conference Records, 1991, written by William J. Hepp and C. Frank Wheatley.





## SPICE Thermal Model

REV 26 April 2002

FDD2572

CTHERM1 TH 6 3.8e-3  
CTHERM2 6 5 4.0e-3  
CTHERM3 5 4 4.2e-3  
CTHERM4 4 3 4.3e-3  
CTHERM5 3 2 8.5e-3  
CTHERM6 2 TL 3.0e-2

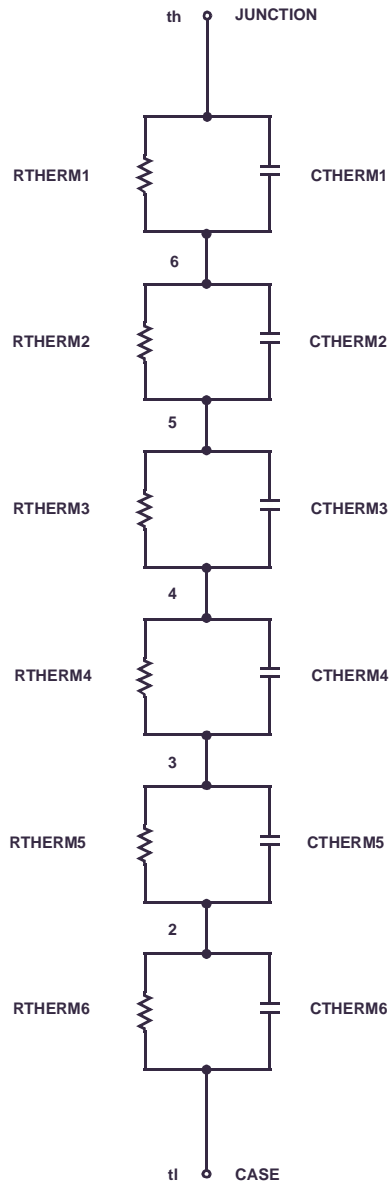
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RTHERM3 5 4 4.5e-2  
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RTHERM5 3 2 3.7e-1  
RTHERM6 2 TL 3.8e-1

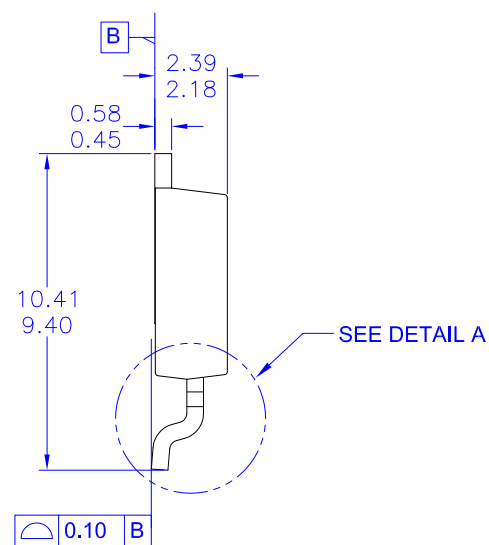
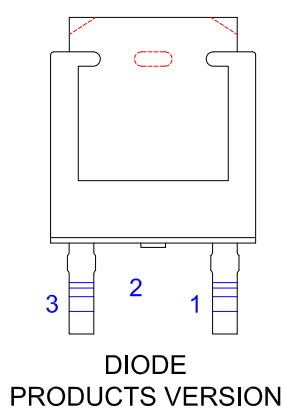
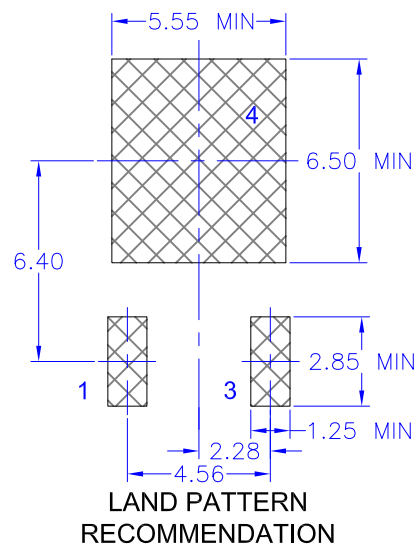
## SABER Thermal Model

SABER thermal model FDD2572  
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thermal\_c th, tl

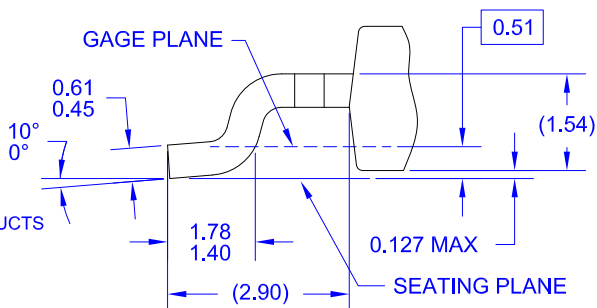
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ctherm.ctherm1 th 6 =3.8e-3
ctherm.ctherm2 6 5 =4.0e-3
ctherm.ctherm3 5 4 =4.2e-3
ctherm.ctherm4 4 3 =4.3e-3
ctherm.ctherm5 3 2 =8.5e-3
ctherm.ctherm6 2 tl =3.0e-2
```

```
rtherm.rtherm1 th 6 =5.5e-4
rtherm.rtherm2 6 5 =5.0e-3
rtherm.rtherm3 5 4 =4.5e-2
rtherm.rtherm4 4 3 =10.5e-2
rtherm.rtherm5 3 2 =3.7e-1
rtherm.rtherm6 2 tl =3.8e-1
}
```





H) DRAWING NUMBER AND REVISION: MKT-TO252A03REV11



**DETAIL A**  
(ROTATED -90°)  
SCALE: 12X



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