Features



Low-Jitter, Wide Frequency Range, Programmable Clock Generator with 10 Outputs

General Description

The MAX3639 is a highly flexible, precision phase-locked loop (PLL) clock generator optimized for the next generation of network equipment that demands low-jitter clock generation and distribution for robust high-speed data transmission. The device features subpicosecond jitter generation, excellent power-supply noise rejection, and pin-programmable LVDS/LVPECL output interfaces. The MAX3639 provides nine differential outputs and one LVCMOS output, divided into three banks. The frequency and output interface of each output bank can be individually programmed, making this device an ideal replacement for multiple crystal oscillators and clock distribution ICs on a system board, saving cost and space.

This 3.3V IC is available in a 7mm x 7mm, 48-pin TQFN package and operates from -40°C to +85°C.

Applications

Ethernet Switch/Router
Wireless Base Station
SONET/SDH Line Cards

PCIe[®], Network Processors

Fibre Channel SAN

Typical Application Circuits and Pin Configuration appear at end of data sheet.

♦ Inputs

Crystal Interface: 18MHz to 33.5MHz LVCMOS Input: 15MHz to 160MHz Differential Input: 15MHz to 350MHz

Outputs

LVCMOS Output: Up to 160MHz LVPECL/LVDS Outputs: Up to 800MHz

- Three Individual Output Banks Pin-Programmable Dividers Pin-Programmable Output Interface
- ♦ Wide VCO Tuning Range (3.60GHz to 4.025GHz)
- Low Phase Jitter

 0.34psrms (12kHz to 20MHz)
 0.14psrms (1.875MHz to 20MHz)
- ◆ Excellent Power-Supply Noise Rejection
- -40°C to +85°C Operating Temperature Range
- ♦ +3.3V Supply

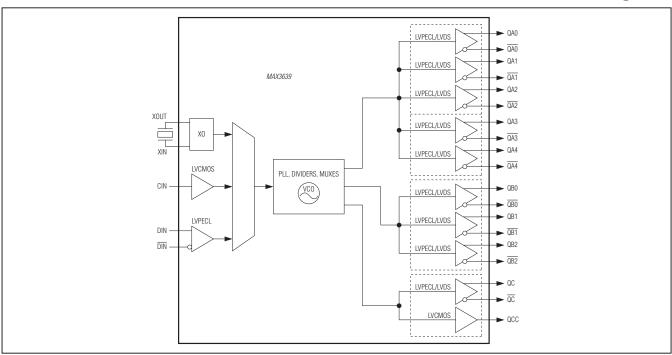
Ordering Information

| PART | TEMP RANGE | PIN-PACKAGE |
|-------------|----------------|-------------|
| MAX3639ETM+ | -40°C to +85°C | 48 TQFN-EP* |

+Denotes a lead(Pb)-free/RoHS-compliant package.

*EP = Exposed pad.

Functional Diagram



PCIe is a registered trademark of PCI-SIG Corp.

ABSOLUTE MAXIMUM RATINGS

| Supply Voltage Range (VCC, VCCA, VCCQA, VCCQB, VCCQC, VCCQCC)0.3V to +4.0V |
|---|
| Voltage Range at CIN, IN_SEL, DM, DF[1:0], |
| DP[1:0], PLL_BP, DA[1:0], DB[1:0], DC[1:0], |
| QA_CTRL1, QA_CTRL2, QB_CTRL, |
| QC_CTRL, QCC0.3V to (VCC + 0.3V) |
| Voltage Range at DIN, DIN (VCC - 2.35V) to (VCC - 0.35V) |
| Voltage Range at QA[4:0], QA[4:0], QB[2:0], |
| $\overline{QB[2:0]}$, QC, \overline{QC} when LVDS Output0.3V to (VCC + 0.3V) |

| Current into QA[4:0], QA[4:0], QB[2:0], Q | B[2:0], |
|---|----------------------|
| QC, QC when LVPECL Output | 56mA |
| Current into QCC | ±50mA |
| Voltage Range at XIN | 0.3V to +1.2V |
| Voltage Range at XOUT | 0.3V to (VCC - 0.6V) |
| Continuous Power Dissipation ($TA = +70^{\circ}$ | C) |
| 48-Pin TQFN (derate 40mW/°C above + | 70°C)3200mW |
| Operating Junction Temperature Range | 55°C to +150°C |
| Storage Temperature Range | 65°C to +160°C |
| | |

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +3.0 \text{V to } +3.6 \text{V}, T_A = -40 ^{\circ}\text{C} \text{ to } +85 ^{\circ}\text{C}. \text{ Typical values are at } V_{CC} = +3.3 \text{V}, T_A = +25 ^{\circ}\text{C}, \text{ unless otherwise noted. Signal applied to CIN or DIN/DIN only when selected as the reference clock.)} (Note 1)$

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---------------|--|--------------|--------------|------------------------|-------------------|
| Supply Current with PLL | laa | Configured with LVPECL outputs | | 170 | 215 | mA |
| Enabled (Note 2) | ICC | Configured with LVDS outputs | | 290 | 365 | mA |
| Supply Current with PLL | | Configured with LVPECL outputs | | 110 | | mA |
| Bypassed (Note 2) | | Configured with LVDS outputs | | 230 | | IIIA |
| LVCMOS/LVTTL CONTROL INP QA_CTRL2, QB_CTRL, QC_CTR | | _, DM, DF[1:0], DA[1:0], DB[1:0], DC[1:0], P | LL_BP, D | P[1:0], Q | A_CTRL1 | Ι, |
| Input High Voltage | VIH | | 2.0 | | | V |
| Input Low Voltage | VIL | | | | 0.8 | V |
| Input High Current | ІІН | V _{IN} = V _{CC} | | | 80 | μΑ |
| Input Low Current | lıL | VIN = 0V | -80 | | | μΑ |
| LVCMOS/LVTTL CLOCK INPUT | (CIN) | | | | | |
| Reference Clock Input Frequency | fREF | | 15 | | 160 | MHz |
| Input Amplitude Range | | Internally AC-coupled (Note 3) | 1.2 | | 3.6 | VP-P |
| Input High Current | IIH | V _{IN} = V _{CC} | | | 80 | μΑ |
| Input Low Current | IIL | V _{IN} = 0V | -80 | | | μΑ |
| Reference Clock Input Duty Cycle | | | 40 | | 60 | % |
| Input Capacitance | | | | 1.5 | | pF |
| DIFFERENTIAL CLOCK INPUT | (DIN, DIN) (N | ote 4) | | | | |
| Differential Input Frequency | fREF | | 15 | | 350 | MHz |
| Input Bias Voltage | Vсмі | | VCC - | VCC - 1.3 | | V |
| Input Differential Voltage Swing | | | 150 | | 1800 | mV _{P-P} |
| Single-Ended Voltage Range | | | Vcc - 2.0 | | V _C C - 0.7 | V |
| Input Differential Impedance | | | 80 | 100 | 120 | Ω |
| Differential Input Capacitance | | | | 1.5 | | pF |

ELECTRICAL CHARACTERISTICS (continued)

 $(VCC = +3.0V \text{ to } +3.6V, TA = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}. \text{ Typical values are at } VCC = +3.3V, TA = +25^{\circ}\text{C}, \text{ unless otherwise noted. Signal applied to CIN or DIN/\overline{DIN} only when selected as the reference clock.) (Note 1)$

| PARAMETER | SYMBOL | CONDITIONS | MIN | TYP | MAX | UNITS |
|---|---------------------|--|------------------------|-------------------------|-------------------------|------------------|
| LVDS OUTPUTS (QA[4:0], QA[4 | :0], QB[2:0], | QB[2:0], QC, QC) (Note 5) | | | | |
| Output Frequency | | | | | 800 | MHz |
| Output High Voltage | VoH | | | | 1.475 | V |
| Output Low Voltage | VoL | | 0.925 | | | V |
| Differential Output Voltage | IV _{OD} I | | 250 | | 400 | mV |
| Change in Magnitude of Differential Output for Complementary States | ΔIVODI | | | | 25 | mV |
| Output Offset Voltage | Vos | | 1.125 | | 1.3 | V |
| Change in Magnitude of Output Offset Voltage for Complementary States | ΔIV _{OS} I | | | | 25 | mV |
| Differential Output Impedance | | | 80 | 100 | 140 | Ω |
| Outro de Outro de | | Short together | | 3 | | A |
| Output Current | | Short to ground | | 6 | | mA |
| Output Current When Disabled | | $V_{Q} = V_{\overline{Q}} = 0V \text{ to } V_{CC}$ | | 10 | | μΑ |
| Output Rise/Fall Time | | 20% to 80% | | 160 | 240 | ps |
| Output Duty-Cycle Distortion | | PLL enabled PLL bypassed (Note 6) | 48 | 50 50 | 52 | % |
| LVPECL OUTPUTS (QA[4:0], Q | V[4:0] OB[3 | 1 7 | <u> </u> | | | |
| Output Frequency | \ | (Hote 1) | | | 800 | MHz |
| Output High Voltage | VoH | | V _{CC} - | V _C C - 0.98 | V _C C - 0.83 | V |
| Output Low Voltage | V _{OL} | | V _{CC} - 1.85 | V _C C - 1.70 | V _C C - 1.55 | V |
| Output-Voltage Swing (Single-Ended) | | | 0.5 | 0.7 | 0.9 | V _{P-P} |
| Output Current When Disabled | | $VQ_{-} = V\overline{Q_{-}} = 0V \text{ to VCC}$ | | 10 | | μΑ |
| Output Rise/Fall Time | | 20% to 80% | | 140 | 240 | ps |
| Outrant Data Outla Distantian | | PLL enabled | 48 | 50 | 52 | 0/ |
| Output Duty-Cycle Distortion | | PLL bypassed (Note 6) | | 50 | | % |
| LVCMOS/LVTTL OUTPUT (QCC | ;) | | • | | | |
| Output Frequency | | | | | 160 | MHz |
| Output High Voltage | | $I_{OH} = -12mA$ | 2.6 | | Vcc | V |
| Output Low Voltage | | IOL = 12mA | | | 0.4 | V |
| Output Rise/Fall Time | | 20% to 80% (Note 8) | 150 | 400 | 850 | ps |
| Output Duty Ovela Distantian | | PLL enabled | 42 | 50 | 58 | 0/ |
| Output Duty-Cycle Distortion | | PLL bypassed (Note 6) | | 50 | | % |
| Output Impedance | | | | 15 | | Ω |

| PARAMETER | SYMBOL | CC | ONDITIONS | MIN | TYP | MAX | UNITS | |
|--|---------|--|-----------------------|-------|------|------|-------|--|
| PLL SPECIFICATIONS | OTIMBOL | | SILDITIONS | IVIII | | WAX | OMITO | |
| | _ | Low VCO (DP1 = | = 0 or NC) | 3600 | 3750 | 3830 | | |
| VCO Frequency Range | fvco | High VCO (DP1 : | | 3830 | 3932 | 4025 | MHz | |
| Phase-Frequency Detector Compare Frequency | fPFD | | | 15 | | 42 | MHz | |
| PLL Jitter Transfer Bandwidth | | | | | 130 | | kHz | |
| | | 25MHz crystal | 12kHz to 20MHz | | 0.34 | 1.0 | | |
| La contra la Colonia de la Col | D.1 | input (Note 9) | 1.875MHz to 20MHz | | 0.14 | | | |
| Integrated Phase Jitter | RJ | 25MHz LVCMOS (Notes 9, 10) | or differential input | | 0.34 | | psRM | |
| Supply-Noise Induced Phase Spur at LVPECL/LVDS Output | | (Note 11) | | | -56 | | dBc | |
| Supply-Noise Induced Phase Spur at LVCMOS Output | | (Note 11) | | | -45 | | dBc | |
| Determinisitic Jitter Induced by Power-Supply Noise | | LVPECL or LVDS (Note 11) | | | 6 | | psp-F | |
| Nonharmonic and Subharmonic Spurs | | (Note 12) | | | -70 | | dBc | |
| | | fOFFSET = 1kHz | | | -111 | | | |
| | | foffset = 10kHz | | | -113 | | dBc/ | |
| SSB Phase Noise at 491.52MHz | | foffset = 100kH | | | -119 | | Hz | |
| | | foffset = 1MHz | | | -136 | | _ | |
| | | foffset ≥ 10MH; | Z | | -147 | | | |
| | | foffset = 1kHz | | | -115 | | - | |
| | | foffset = 10kHz | | -116 | | dBc/ | | |
| SSB Phase Noise at 312.5MHz | | foffset = 100kHz | | -122 | | Hz | | |
| | | fOFFSET = 1MHz | | | -139 | | - | |
| | | foffset ≥ 10MHz | Z | | -149 | | | |
| | | fOFFSET = 1kHz | | | -117 | | - | |
| SSB Phase Noise at 245.76MHz | | fOFFSET = 10kHz -119 fOFFSET = 100kHz -125 | | | dBc, | | | |
| | | foffset = 1MHz | | -142 | | Hz | | |
| | | fOFFSET ≥ 10MH; | | | -151 | | - | |
| | | foffset = 1kHz | | | -122 | | | |
| | | fOFFSET = 10kHz | · | -123 | | | + | |
| SSB Phase Noise at 156.25MHz | | foffset = 100kHz | | | -129 | | dBc/ | |
| | | foffset = 1MHz | | | -145 | | - Hz | |
| | | foffset ≥ 10MH; | | | -152 | | 1 | |

ELECTRICAL CHARACTERISTICS (continued)

 $(VCC = +3.0V \text{ to } +3.6V, T_A = -40^{\circ}\text{C} \text{ to } +85^{\circ}\text{C}.$ Typical values are at $VCC = +3.3V, T_A = +25^{\circ}\text{C}$, unless otherwise noted. Signal applied to CIN or DIN/\overline{DIN} only when selected as the reference clock.) (Note 1)

| PARAMETER | SYMBOL | CONDITIONS MIN TYP M | | MAX | UNITS | | |
|---------------------------|--------|----------------------|-----------------|------|-------|------------|--|
| | | foffset = 1kHz | | -123 | | | |
| | | fOFFSET = 10kHz | | -124 | | -10-/ | |
| SSB Phase Noise at 125MHz | | foffset = 100kHz | | -130 | | dBc/ Hz | |
| | | foffset = 1MHz | | -147 | | 112 | |
| | | | foffset ≥ 10MHz | | -153 | | |
| SSB Phase Noise at 100MHz | | fOFFSET = 1kHz | | -126 | | | |
| | | foffset = 10kHz | | -127 | | JD. / | |
| | | foffset = 100kHz | | -133 | | dBc/ Hz | |
| | | foffset = 1MHz | | -148 | |] ''2 | |
| | | foffset ≥ 10MHz | | -152 | | | |

- **Note 1:** A series resistor of up to 10.5Ω is allowed between V_{CC} and V_{CCA} for filtering supply noise when system power-supply tolerance is V_{CC} = $3.3V \pm 5\%$. See Figure 3.
- Note 2: Measured with all outputs enabled and unloaded.
- Note 3: CIN can be AC- or DC-coupled. See Figure 8. Input high voltage must be ≤ VCC + 0.3V.
- Note 4: DIN can be AC- or DC-coupled. See Figure 10.
- **Note 5:** Measured with 100Ω differential load.
- Note 6: Measured with crystal input, or with 50% duty cycle LVCMOS or differential input.
- **Note 7:** Measured with output termination of 50Ω to VCC 2V or Thevenin equivalent.
- **Note 8:** Measured with a series resistor of 33Ω to a load capacitance of 3.0pF. See Figure 1.
- Note 9: Measured at 156.25MHz.
- **Note 10:** Measured using LVCMOS/LVTTL input with slew rate ≥ 1.0V/ns, or differential input with slew rate ≥ 0.5V/ns.
- **Note 11:** Measured at 156.25MHz output with 200kHz, 50mV_{P-P} sinusoidal signal on the supply using the crystal input and the power-supply filter shown in Figure 3. See the *Typical Operating Characteristics* for other supply noise frequencies. Deterministic jitter is calculated from the measured power-supply-induced spurs. For more information, refer to Application Note 4461: *HFAN-04.5.5*: *Characterizing Power-Supply Noise Rejection in PLL Clock Synthesizers*.
- Note 12: Measured with all outputs enabled and all three banks at different frequencies.

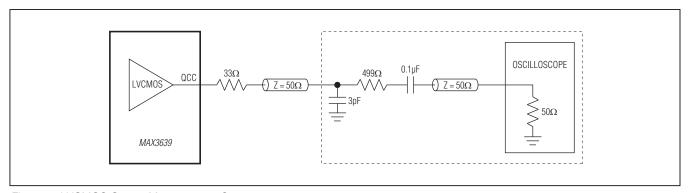
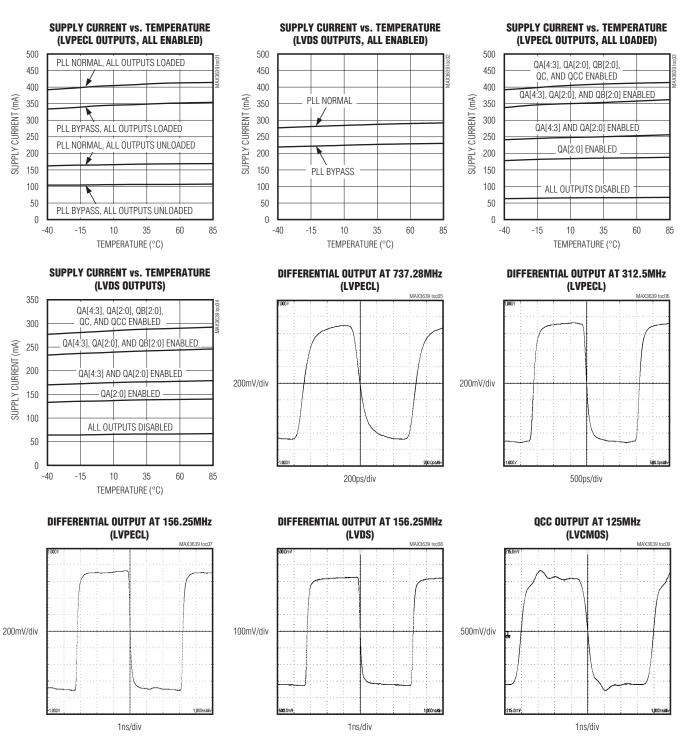


Figure 1. LVCMOS Output Measurement Setup

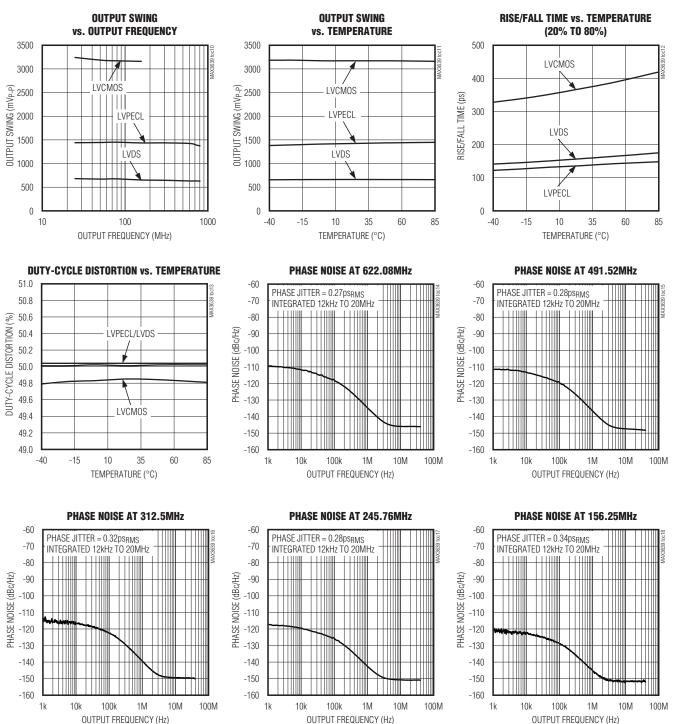
Typical Operating Characteristics

 $(V_{CC} = 3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$



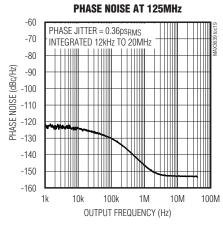
Typical Operating Characteristics (continued)

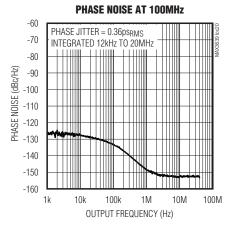
 $(V_{CC} = 3.3V, T_A = +25^{\circ}C, unless otherwise noted.)$

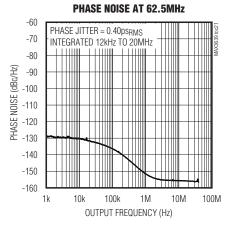


Typical Operating Characteristics (continued)

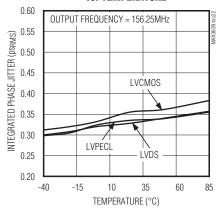
(VCC = 3.3V, TA = +25°C, unless otherwise noted.)

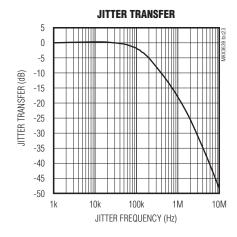




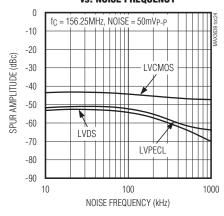


INTEGRATED PHASE JITTER (12kHz TO 20MHz) vs. Temperature

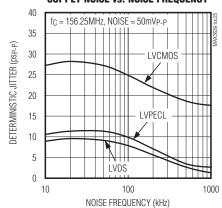




SPURS INDUCED BY POWER-SUPPLY NOISE vs. Noise Frequency



DETERMINISTIC JITTER INDUCED BY POWER SUPPLY NOISE vs. NOISE FREQUENCY



Pin Description

| PIN | NAME | FUNCTION | | | |
|--------|----------|--|--|--|--|
| 1 | DM | LVCMOS/LVTTL Input. Three-level control for input divider M. See Table 3. | | | |
| 2 | XIN | Crystal Oscillator Input | | | |
| 3 | XOUT | Crystal Oscillator Output | | | |
| 4 | Vcc | Core Power Supply. Connect to +3.3V. | | | |
| 5 | IN_SEL | LVCMOS/LVTTL Input. Three-level control for input mux. See Table 1. | | | |
| 6 | PLL_BP | LVCMOS/LVTTL Input. Three-level control for PLL bypass mode. See Table 2. | | | |
| 7, 8 | DF1, DF0 | LVCMOS/LVTTL Inputs. Three-level controls for feedback divider F. See Table 4. | | | |
| 9 | QC_CTRL | LVCMOS/LVTTL Input. Three-level control input for C-bank output interface. See Table 10. | | | |
| 10 | VCCA | Power Supply for Internal Voltage-Controlled Oscillators (VCOs). See Figure 3. | | | |
| 11, 12 | DP1, DP0 | LVCMOS/LVTTL Inputs. Three-level controls for VCO select and prescale divider P. See Table 7. | | | |
| 13, 14 | DB1, DB0 | LVCMOS/LVTTL Inputs. Three-level controls for output divider B. See Table 5. | | | |
| 15, 16 | DA1, DA0 | LVCMOS/LVTTL Inputs. Three-level controls for output divider A. See Table 5. | | | |
| 17, 18 | DC1, DC0 | LVCMOS/LVTTL Inputs. Three-level controls for output divider C. See Table 6. | | | |
| 19 | QA_CTRL2 | LVCMOS/LVTTL Input. Three-level control for QA[4:3] output interface. See Table 8. | | | |
| 20 | Vccqcc | Power Supply for QCC Output. Connect to +3.3V. | | | |
| 21 | QCC | C-Bank LVCMOS Clock Output | | | |
| 22, 23 | QC, QC | C-Bank Differential Output. Configured as LVPECL or LVDS with the QC_CTRL pin. | | | |
| 24 | Vccqc | Power Supply for C-Bank Differential Output. Connect to +3.3V. | | | |
| 25, 36 | VCCQA | Power Supply for A-Bank Differential Outputs. Connect to +3.3V. | | | |
| 26, 27 | QA4, QA4 | A-Bank Differential Output. Configured as LVPECL or LVDS with the QA_CTRL2 pin. | | | |
| 28, 29 | QA3, QA3 | A-Bank Differential Output. Configured as LVPECL or LVDS with the QA_CTRL2 pin. | | | |
| 30, 31 | QA2, QA2 | A-Bank Differential Output. Configured as LVPECL or LVDS with the QA_CTRL1 pin. | | | |
| 32, 33 | QA1, QA1 | A-Bank Differential Output. Configured as LVPECL or LVDS with the QA_CTRL1 pin. | | | |
| 34, 35 | QAO, QAO | A-Bank Differential Output. Configured as LVPECL or LVDS with the QA_CTRL1 pin. | | | |
| 37 | VCCQB | Power Supply for B-Bank Differential Outputs. Connect to +3.3V. | | | |
| 38, 39 | QB0, QB0 | B-Bank Differential Output. Configured as LVPECL or LVDS with the QB_CTRL pin. | | | |
| 40, 41 | QB1, QB1 | B-Bank Differential Output. Configured as LVPECL or LVDS with the QB_CTRL pin. | | | |
| 42, 43 | QB2, QB2 | B-Bank Differential Output. Configured as LVPECL or LVDS with the QB_CTRL pin. | | | |
| 44 | QA_CTRL1 | LVCMOS/LVTTL Input. Three-level control for QA[2:0] output interface. See Table 8. | | | |
| 45 | QB_CTRL | LVCMOS/LVTTL Input. Three-level control for B-bank output interface. See Table 9. | | | |
| 46, 47 | DIN, DIN | Differential Clock Input. Operates up to 350MHz. This input can accept DC-coupled LVPECL signals, and is internally biased to accept AC-coupled LVDS, CML, and LVPECL signals. | | | |
| 48 | CIN | LVCMOS Clock Input. Operates up to 160MHz. | | | |
| _ | EP | Exposed Pad. Connect to supply ground for proper electrical and thermal performance. | | | |

Detailed Description

The MAX3639 is a low-jitter clock generator designed to operate over a wide range of frequencies. It consists of a selectable reference clock (on-chip crystal oscillator, LVCMOS input, or differential input), PLL with on-chip VCO, pin-programmable dividers and muxes, and three

banks of clock outputs. See Figure 2. The output banks include nine pin-programmable LVDS/LVPECL output buffers and one LVCMOS output buffer. The frequency, enabling, and output interface of each output bank can be individually programmed. In addition the A-bank is split into two banks with programmable enabling and

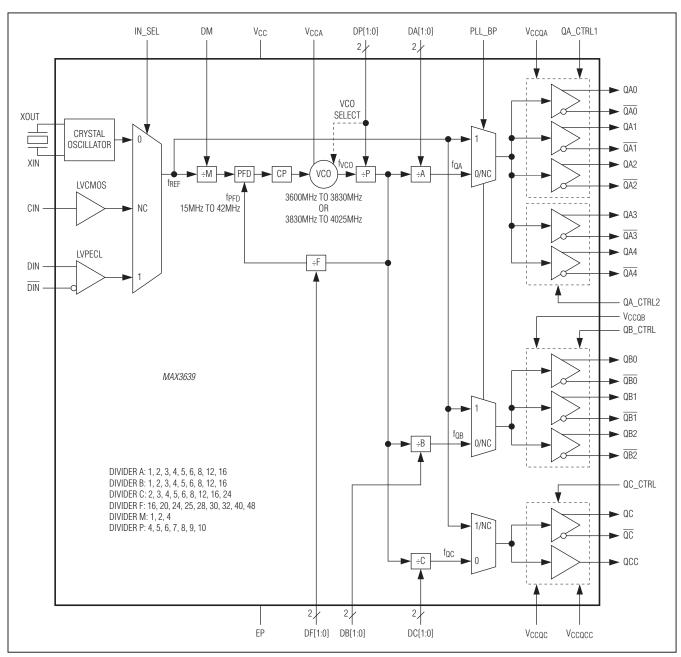


Figure 2. Detailed Functional Diagram

output interface. A PLL bypass mode is also available for system testing or clock distribution.

Crystal Oscillator

The on-chip crystal oscillator provides the low-frequency reference clock for the PLL. This oscillator requires an external crystal connected between XIN and XOUT. See the *Crystal Selection and Layout* section for more information. The XIN and XOUT pins can be left open if not used.

LVCMOS Clock Input

An LVCMOS-compatible clock source can be connected to CIN to serve as the PLL reference clock. The input is internally biased to allow AC- or DC-coupling (see the *Applications Information* section). It is designed to operate from 15MHz to 160MHz. No signal should be applied to CIN if not used.

Differential Clock Input

A differential clock source can be connected to DIN to serve as the PLL reference clock. This input operates from 15MHz to 350MHz and contains an internal 100Ω differential termination. This input can accept DC-coupled LVPECL signals, and is internally biased to accept AC-coupled LVDS, CML, and LVPECL signals

(see the *Applications Information* section). No signal should be applied to DIN if not used.

Phase-Locked Loop (PLL)

The PLL takes the signal from the crystal oscillator, LVCMOS clock input, or differential clock input and synthesizes a low-jitter, high-frequency clock. The PLL contains a phase-frequency detector (PFD), a charge pump (CP), and two low phase noise VCOs that combined give a wide 3.60GHz to 4.025GHz frequency range. The high-frequency VCO output is divided by prescale divider P and then is connected to the PFD input through a feedback divider F. The PFD compares the reference frequency to the divided-down VCO output and generates a control signal that keeps the VCO locked to the reference clock. The high-frequency VCO/P output clock is sent to the output dividers. To minimize noise-induced jitter, the VCO supply (VCCA) is isolated from the core logic and output buffer supplies.

Dividers and Muxes

The dividers and muxes are set with three-level control inputs. Divider settings and routing information are given in Tables 1 to 7. See Table 11 for example divider configurations used in various applications.

Table 1. PLL Input

| IN_SEL | INPUT | | | |
|--------|--|--|--|--|
| 0 | Crystal Input. XO circuit is disabled when not selected. | | | |
| 1 | Differential Input. No signal should be applied to DIN if not selected | | | |
| NC | LVCMOS Input. No signal should be applied to CIN if not selected. | | | |

Table 2. PLL Bypass

| PLL_BP | PLL OPERATION |
|--------|---|
| 0 | PLL Enabled for Normal Operation. All outputs from the A, B, and C banks are derived from the VCO. |
| 1 | PLL Bypassed. Selected input passes directly to the outputs. Both VCOs are disabled to minimize power consumption and intermodulation spurs. Used for system testing or clock distribution. |
| NC | The outputs from A-bank and B-bank are derived from the VCO, but the C-bank outputs are directly driven from the input signal for purposes of daisy chaining. |

Table 3. Input Divider M

| DM | M DIVIDER RATIO | | | | |
|----|-----------------|--|--|--|--|
| 0 | ÷1 | | | | |
| 1 | ÷2 | | | | |
| NC | ÷4 | | | | |

Note: When the on-chip XO is selected (IN_SEL = 0), the setting DM = 0 is required.

Table 4. PLL Feedback Divider F

| DF1 | DF0 | F DIVIDER RATIO |
|-----|-----|-----------------|
| 0 | 0 | ÷25 |
| 0 | 1 | ÷20 |
| 1 | 0 | ÷16 |
| 1 | 1 | ÷32 |
| 1 | NC | ÷24 |
| NC | 1 | ÷30 |
| 0 | NC | ÷40 |
| NC | 0 | ÷48 |
| NC | NC | ÷28 |

Table 5. Output Divider A, B

| DA1/DB1 | DA0/DB0 | A, B DIVIDER RATIO |
|---------|---------|--------------------|
| 0 | 0 | ÷2 |
| 0 | 1 | ÷3 |
| 1 | 0 | ÷4 |
| 1 | 1 | ÷5 |
| 1 | NC | ÷6 |
| NC | 1 | ÷8 |
| 0 | NC | ÷12 |
| NC | 0 | ÷16 |
| NC | NC | ÷1 |

Table 6. Output Divider C

| DC1 | DC0 | C DIVIDER RATIO |
|-----|-----|-----------------|
| 0 | 0 | ÷2 |
| 0 | 1 | ÷3 |
| 1 | 0 | ÷4 |
| 1 | 1 | ÷5 |
| 1 | NC | ÷6 |
| NC | 1 | ÷8 |
| 0 | NC | ÷12 |
| NC | 0 | ÷16 |
| NC | NC | ÷24 |

Table 7. VCO Select and Prescale Divider P

| DP1 | DP0 | VCO FREQUENCY RANGE (MHz) | P DIVIDER RATIO | (VCO/P) FREQUENCY RANGE (MHz) |
|-----|-----|------------------------------------|-----------------------|-------------------------------------|
| | 0 | Low | ÷5 | 720 to 766 |
| 0 | 1 | (3600 to | ÷6 | 600 to 638.33 |
| | NC | 3830) | ÷9 | 400 to 425.50 |
| | 0 | | ÷7 | 547.14 to 575 |
| NC | 1 | | ÷10 | 383 to 402.50 |
| | NC | High | ÷8 | 478.75 to 503.12 |
| | 0 | (3830 to 4025) | ÷4 | 957.50 to 1006.25 |
| 1 | 1 | | ÷5 | 766 to 805 |
| | NC | | ÷6 | 638.33 to 670.83 |

Table 8. A-Bank Output Interface

| QA_CTRL1 | QA[2:0] OUTPUT | | | |
|----------|------------------------------------|--|--|--|
| 0 | QA[2:0] = LVDS | | | |
| 1 | QA[2:0] = LVPECL | | | |
| NC | QA[2:0] disabled to high impedance | | | |
| QA_CTRL2 | QA[4:3] OUTPUT | | | |
| 0 | QA[4:3] = LVDS | | | |
| 1 | QA[4:3] = LVPECL | | | |
| NC | QA[4:3] disabled to high impedance | | | |

Table 9. B-Bank Output Interface

| QB_CTRL | QB[2:0] OUTPUT | | | | |
|---------|------------------------------------|--|--|--|--|
| 0 | QB[2:0] = LVDS | | | | |
| 1 | QB[2:0] = LVPECL | | | | |
| NC | QB[2:0] disabled to high impedance | | | | |

Table 10. C-Bank Output Interface

| QC_CTRL | QC AND QCC OUTPUT |
|---------|---------------------------------------|
| 0 | QC = LVDS, QCC = LVCMOS |
| 1 | QC = LVPECL, QCC = LVCMOS |
| NC | QC and QCC disabled to high impedance |

LVDS/LVPECL Clock Outputs

The differential clock outputs (QA[4:0], QB[2:0], QC) operate up to 800MHz and have a pin-programmable LVDS/LVPECL output interface. See Tables 8 to 10. When configured as LVDS, the buffers are designed to drive transmission lines with a 100Ω differential

termination. When configured as LVPECL, the buffers are designed to drive transmission lines terminated with 50Ω to VCC - 2V. Unused output banks can be disabled to high impedance and unused outputs can be left open.

LVCMOS Clock Output

The LVCMOS clock output operates up to 160MHz and is designed to drive a single-ended high-impedance load. If unused, this output can be left open or the C-bank can be disabled to high impedance.

Internal Reset

During power-on, a power-on reset (POR) signal is generated to synchronize all dividers. A reset signal is also generated if any control pin is changed. Outputs within a bank are phase aligned, but outputs bank-to-bank may not be phase aligned.

Applications Information

Output Frequency Configuration

The MAX3639 output frequencies (fQA, fQB, fQC) are functions of the reference frequency (fREF) and the pin-programmable dividers (A, B, C, F, M). The relationships can be expressed as:

$$f_{QA} = \frac{f_{REF}}{M} \times \frac{F}{A}$$
 (1)

$$f_{QB} = \frac{f_{REF}}{M} \times \frac{F}{B}$$
 (2)

$$f_{QC} = \frac{f_{REF}}{M} \times \frac{F}{C}$$
 (3)

The frequency ranges for the selected reference clocks are 18MHz to 33.5MHz for the crystal oscillator input, 15MHz to 160MHz for the LVCMOS input, and 15MHz to 350MHz for the differential input. The available dividers are given in Tables 3 to 6.

For a given reference frequency fREF, the input divider M, the PLL feedback divider F, and VCO prescale divider P must be configured so the VCO frequency (fvco) falls within the specified ranges. Invalid PLL configuration leads to VCO frequencies beyond the specified ranges and can result in loss of lock. An expression for the VCO frequency along with the specified ranges is given by:

$$f_{VCO} = \frac{f_{REF}}{M} \times F \times P \tag{4}$$

 $3600MHz \le f_{VCO} \le 3830MHz \text{ (when DP1 = 0) (5)}$

 $3830MHz \le f_{VCO} \le 4025MHz \text{ (when DP1 = 1 or NC) } (6)$

The prescale divider P is set by pins DP1 and DP0 as given in Table 7.

In addition, the reference clock frequency and input divider M must also be selected so the PFD compare frequency (fPFD) falls within the specified range of 15MHz to 42MHz. If applicable, the higher fPFD should be selected for optimal litter performance.

$$f_{PFD} = \frac{f_{REF}}{M} = \frac{f_{VCO}}{P \times F}$$
 (7)

 $15MHz \le fppd \le 42MHz$ (8)

Note that the reference clock frequency is not limited by the fPFD range when the PLL is in bypass mode.

Example Frequency Configuration

The following is an example of how to find divider ratios for a valid PLL configuration, given a requirement of input and output frequencies.

1) Select input and output frequencies for an Ethernet application.

fREF = 25MHz

fQA = 312.5MHz

fQB = 156.25MHz

fQC = 125MHz

- 2) Find the input divider M for a valid PFD compare frequency. Using Table 3 and equations (7) and (8), it is determined that $M = \div 1$ is the only valid option.
- 3) Find the feedback divider F and prescale divider P for a valid fvco. Using Tables 4 and 7 along with equations (4), (5), and (6), it is determined that F = ÷25 and P = ÷6 results in fvco = 3750MHz, which is within the valid range of the low VCO.
- 4) Find the output dividers A, B, C for the required output frequencies. Using Tables 5 and 6 and equations (1), (2), and (3), it is determined that A = ÷2 gives fQA = 312.5MHz, B = ÷4 gives fQB = 156.25MHz, and C = ÷5 gives fQC = 125MHz.

Table 11 provides input and output frequencies along with valid divider ratios for a variety of applications.

Table 11. Reference Frequencies and Divider Ratios for Various Applications

| fREF (MHz) | INPUT DIVIDER (M) | PLL FEEDBACK DIVIDER (F) | VCO FREQUENCY (MHz) | VCO PRESCALE DIVIDER (P) | OUTPUT DIVIDER (A, B, C) | OUTPUT FREQUENCY (MHz) | APPLICATIONS | | | | | |
|------------------------|-------------------------|-----------------------------------|---------------------------|-----------------------------------|--------------------------------|------------------------------|------------------------------|---------|-------|-------|--------|--|
| 15.00 | 4 | 48 | | 5 | 1 | 737.28 | | | | | | |
| 15.36 | 1 | 48 | | 5 | 2 | 368.64 | | | | | | |
| 20.70 | 4 | 0.4 | | 5 | 3 | 245.76 | | | | | | |
| 30.72 | 1 | 24 | 3686.4 | 5 | 4 | 184.32 | | | | | | |
| 61.44 | 2 | 24 | 3080.4 | 5 | 6 | 122.88 | | | | | | |
| | | | | 5 | 8 | 92.16 | | | | | | |
| 122.88 | 4 | 24 | | 5 | 12 | 61.44 | | | | | | |
| | | | | 5 | 24 | 30.72 | | | | | | |
| 15.36 | 1 | 40 | | 6 | 1 | 614.4 | Wireless Base Station: | | | | | |
| 19.2 | 1 | 32 | | 6 | 2 | 307.2 | WCDMA, cdma2000®, | | | | | |
| 30.72 | 1 | 20 | | 6 | 3 | 204.8 | LTE, TD_SCDMA, | | | | | |
| 38.4 | 1 | 16 | 00004 | 6 | 4 | 153.6 | WiMAX™, GSM | | | | | |
| 61.44 | 2 | 20 | 3686.4 | 6 | 5 | 122.88 | | | | | | |
| 76.8 | 2 | 16 | | 6 | 6 | 102.4 | | | | | | |
| 122.88 | 4 | 20 | | 6 | 8 | 76.8 | | | | | | |
| 153.6 | 4 | 16 | | 6 | 12 | 51.2 | | | | | | |
| 30.72 | 1 | 32 | | 4 (8) | 2 (1) | 491.52 | | | | | | |
| 61.44 | 2 | 32 | 0000 40 | 4 (8) | 4 (2) | 245.76 | | | | | | |
| 100.00 | , | 4 | 0.0 | 00 | | | | 3932.16 | 4 (8) | 8 (4) | 122.88 | |
| 122.88 | 4 | 32 | | 4 (8) | 16 (8) | 61.44 | | | | | | |
| 13 | 1 | 32 | | 9 | 1 | 416 | | | | | | |
| 26 | 1 | 16 | 0744 | 9 | 4 | 104 | 0014 | | | | | |
| 50 | | 10 | 3744 | 9 | 8 | 52 | GSM | | | | | |
| 52 | 2 | 16 | | 9 | 16 | 26 | | | | | | |
| 25/31.25/ 62.5/125/ | 1/1/2/4/4 | 25/20/20/ | | 6 | 1 | 625 | | | | | | |
| 156.25 | 1, 1, 4, 7, 7, 7 | 20/16 | 3750 | 6 | 2 | 312.5 | Ethernet | | | | | |
| 26.04166 | 66 1 | 24 | 3750 | 6 | 4 | 156.25 | | | | | | |
| | | _ : | | 6 | 5 | 125 | | | | | | |
| | | | | 5 | 3 | 250 | | | | | | |
| 25/31.25/ | | | 3750 | 5 | 4 | 187.5 | | | | | | |
| 62.5/125 | 1/1/2/4 | 30/24/24/24 | | 5 | 5 | 150 | Ethernet | | | | | |
| ., | 52.0/120 | | | 5 | 6 | 125 | | | | | | |
| | | | | 5 | 12 | 62.5 | | | | | | |

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Table 11. Reference Frequencies and Divider Ratios for Various Applications (continued)

| f _{REF} (MHz) | INPUT DIVIDER (M) | PLL FEEDBACK DIVIDER (F) | VCO FREQUENCY (MHz) | VCO PRESCALE DIVIDER (P) | OUTPUT DIVIDER (A, B, C) | OUTPUT FREQUENCY (MHz) | APPLICATIONS | | |
|---------------------------|-------------------------|-----------------------------------|---------------------------|-----------------------------------|--------------------------------|------------------------------|---|--|--|
| | | | | 6 | 2 | 318.75 | | | |
| | | | | 6 | 3 | 212.5 | | | |
| 26.5625 | 1 | 24 | 3825 | 6 | 4 | 159.375 | FC-SAN | | |
| | | | | 6 | 6 | 106.25 | | | |
| | | | | 6 | 12 | 53.125 | | | |
| 19.44 | 1 | 32 | | 6 | 1 | 622.08 | | | |
| 38.88 | 1 | 16 | | 6 | 2 | 311.04 | | | |
| | | | 3732.48 | 6 | 4 | 155.52 | SONET/SDH, STM-N | | |
| 155.52 | 4 | 16 | | 6 | 8 | 77.76 | | | |
| | | | | 6 | 16 | 38.88 | | | |
| 33.3/66.7/ 133.3 | 1/2/4 | 24 | | 5 | 2 | 400 | | | |
| | | | | 5 | 3 | 266.67 | Server, FB-DIMM, | | |
| | | | | 5 | 4 | 200 | Network Processor, | | |
| 25/50/100 1/2/4 | 4 32 | 4000 | 5 | 6 | 133.333 | DDR/QDR Memory, | | | |
| | | | 5 | 8 | 100 | PCIe, SATA | | | |
| | | | 5 | 12 | 66.67 | | | | |
| | | | | 5 | 16 | 50 | | | |
| 33.3/66.7/ 133.3 | 1/2/4 | 30 | | 4 | 2 | 500 | | | |
| 25/50/100 | 1/2/4/8 | 40 | | 4 | 3 | 333.33 | Server, FB-DIMM, | | |
| | | 4 | 4000 | 4 | 4 | 250 | Processor Clock and | | |
| 31.25/ | 5/ | | | 4 | 5 | 200 | DDR/QDR Memory, PCIe, SATA | | |
| 62.5/125 | 1/2/4 | 32 | | 4 | 6 | 166.67 | TOIE, SATA | | |
| | | | | 4 | 8 | 125 | | | |
| | , | | 00010 | 5 | 6 | 131.04 | | | |
| 32.76 | 1 | 24 | 3931.2 | 5 | 12 | 65.52 | Microwave Radio Link | | |
| 20.82857 | | 32 | | 6 | 1 | 666.514 | 07111 100: 00: | | |
| 44.0574 | 1 | 10 | 3999.084 | 6 | 2 | 333.257 | OTU1, 10Gbps SONET | | |
| 41.6571 | | 16 | | 6 | 4 | 166.6285 | with FEC | | |
| 0E 7010E | 4 | 0.5 | 2007 1075 | 6 | 1 | 644.53125 | 10Gbps Ethernet with | | |
| 25.78125 | 1 | 25 | 3867.1875 | 6 | 4 | 161.1328125 | FEC | | |
| 07.000570 | 4 | 0.4 | 2044 52125 | 6 | 1 | 657.421875 | 100km F0 | | |
| 27.392578 | 1 | 24 | 24 3944.53125 | | 4 | 164.355 | 10Gbps FC | | |
| 20.916 | 1 | 32 | | 6 | 1 | 669.3265 | OTUO 1001 0015T | | |
| 41 0000 | | | 4015 95949 | | 2 | 334.66 | OTU2, 10Gbps SONET with Digital Wrapper | | |
| 41.8329 1 | 16 | | 6 | 4 | 167.33 | with Digital Wapper | | | |

Power-Supply Filtering

The MAX3639 is a mixed analog/digital IC. The PLL contains analog circuitry susceptible to random noise. To take full advantage of on-board filtering and noise attenuation, in addition to excellent on-chip power-supply rejection, this part provides a separate power-supply pin, VCCA, for the VCO circuitry. Figure 3 illustrates the recommended power-supply filter network for VCCA. The purpose of this design technique is to ensure clean input power supply to the VCO circuitry and to improve the overall immunity to power-supply noise. This network requires that the power supply is +3.3V ±5%. Decoupling capacitors should be used on all other supply pins for best performance. All supply connections should be driven from the same source.

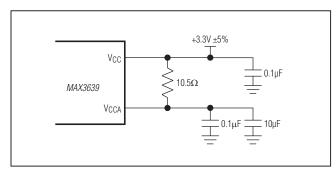


Figure 3. Power-Supply Filter

Ground Connection

The 48-pin TQFN package features an exposed pad (EP), which provides a low resistance thermal path for heat removal from the IC and also the electrical ground. For proper operation, the EP must be connected to the circuit board ground plane with multiple vias.

Crystal Selection and Layout

The MAX3639 features an integrated on-chip crystal oscillator to minimize system implementation cost. The crystal oscillator is designed to drive a fundamental mode, AT-cut crystal resonator. See Table 12 for recommended crystal specifications. See Figure 4 for the crystal equivalent circuit and Figure 5 for the recommended external capacitor connections. The crystal, trace, and two external capacitors should be placed on the board as close as possible to the XIN and XOUT pins to reduce crosstalk of active signals into the oscillator.

The total load capacitance for the crystal is a combination of external and on-chip capacitance. The layout shown in Figure 6 gives approximately 1.7pF of trace plus footprint capacitance per side of the crystal. Note the ground plane is removed under the crystal to minimize capacitance. There is approximately 2.5pF of on-chip capacitance between XIN and XOUT. With an external 27pF capacitor connected to XIN and a 33pF capacitor connected to XOUT, the total load capacitance for the crystal is approximately 18pF. The XIN and XOUT pins can be left open if not used.

Table 12. Crystal Selection Parameters

| PARAMETER | SYMBOL | MIN | TYP | MAX | UNITS |
|------------------------------------|--------|-----|-----|------|-------|
| Crystal Oscillation Frequency | fosc | 18 | 25 | 33.5 | MHz |
| Shunt Capacitance | Co | | 2.0 | 7.0 | pF |
| Load Capacitance | CL | | 18 | | pF |
| Equivalent Series Resistance (ESR) | Rs | | 10 | 50 | Ω |
| Maximum Crystal Drive Level | | | | 200 | μW |

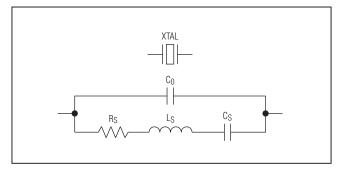


Figure 4. Crystal Equivalent Circuit

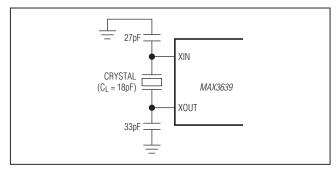


Figure 5. Crystal, Capacitor Connections

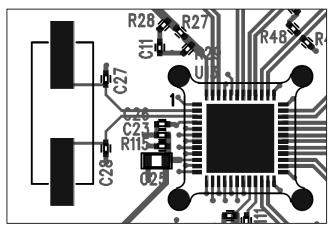


Figure 6. Crystal Layout

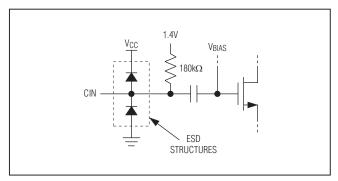


Figure 7. Equivalent CIN Circuit

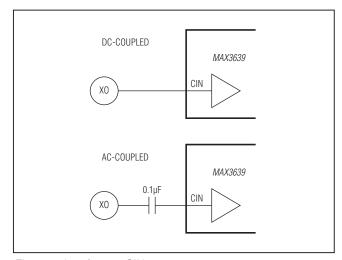


Figure 8. Interface to CIN

Interfacing with LVCMOS Input

The equivalent LVCMOS input circuit for CIN is given in Figure 7. This input is internally biased to allow AC- or DC-coupling, and has $180 k\Omega$ input impedance. See Figure 8 for the interface circuit. No signal should be applied to CIN if not used.

Interfacing with Differential Input

The equivalent input circuit for DIN is given in Figure 9. This input operates up to 350MHz and contains an internal 100Ω differential termination as well as a 35Ω common-mode termination. The common-mode termination ensures good signal integrity when connected to a source with large common-mode signals. The input can accept DC-coupled LVPECL signals, and is internally biased to accept AC-coupled LVDS, CML, and LVPECL signals (Figure 10). No signal should be applied to DIN if not used.

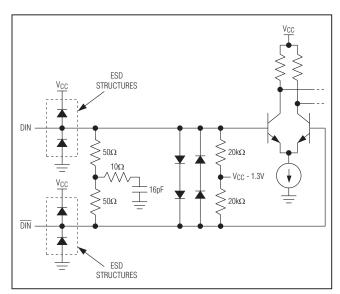


Figure 9. Equivalent DIN Circuit

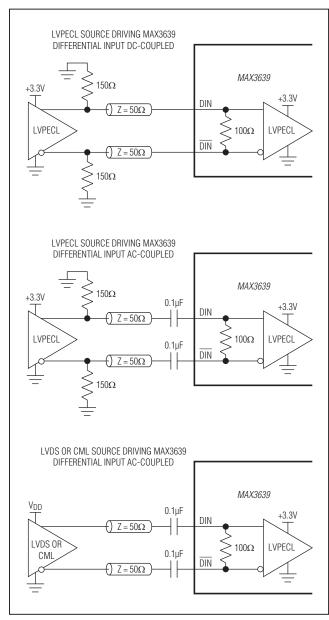


Figure 10. Interfacing to DIN

Interfacing with LVPECL Outputs

The equivalent LVPECL output circuit is given in Figure 11. These outputs are designed to drive a pair of 50Ω transmission lines terminated with 50Ω to V_{TT} = V_{CC} - 2V. If a separate termination voltage (V_{TT}) is not available, other terminations methods can be used, as shown in Figure 12. For more information on LVPECL terminations and how to interface with other logic families, refer to Application Note 291: HFAN-01.0: Introduction to LVDS, PECL, and CML.

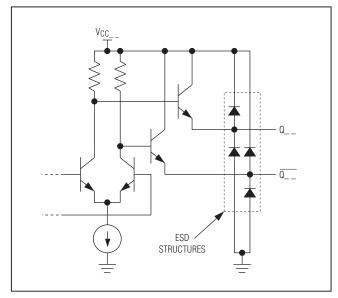


Figure 11. Equivalent LVPECL Output Circuit

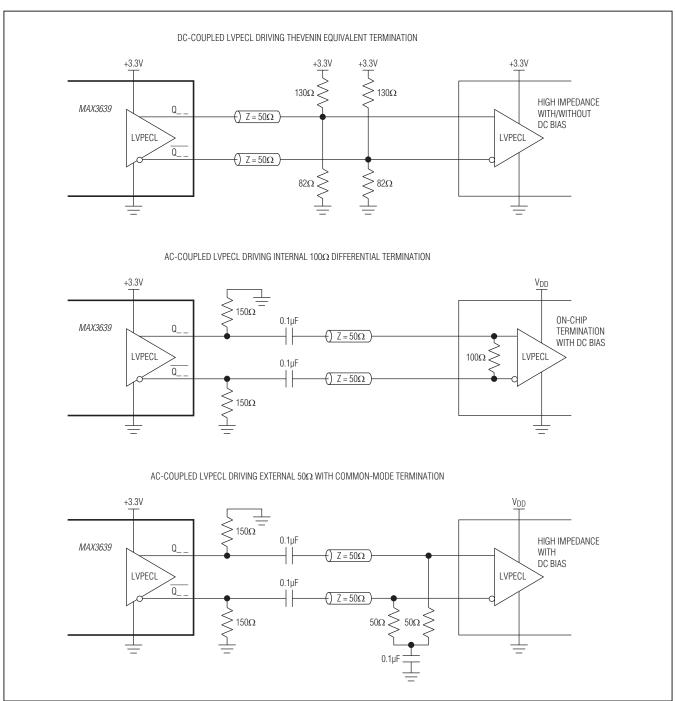


Figure 12. Interface to LVPECL Outputs

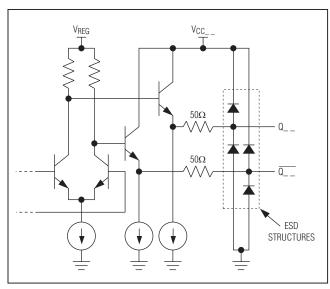


Figure 13. Equivalent LVDS Output Circuit

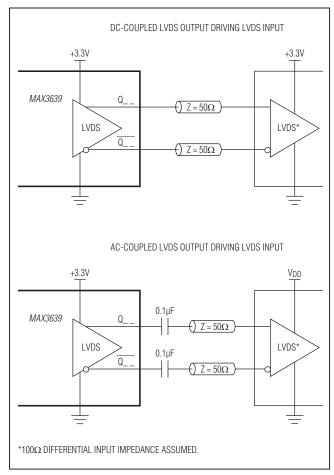


Figure 14. Interface to LVDS Outputs

Interfacing with LVDS Outputs

The equivalent LVDS output circuit is given in Figure 13. These outputs provide 100Ω differential output impedance designed to drive a 100Ω differential transmission line terminated with a 100Ω differential load. Example interface circuits are shown in Figure 14. For more information on LVDS terminations and how to interface with other logic families, refer to Application Note 291: *HFAN-01.0: Introduction to LVDS, PECL, and CML*.

Interfacing with LVCMOS Output

The equivalent LVCMOS output circuit is given in Figure 15. This output provides 15Ω output impedance and is designed to drive a high-impedance load. A series resistor of 33Ω is recommended at the LVCMOS output before the transmission line. An example interface circuit is shown in Figure 16.

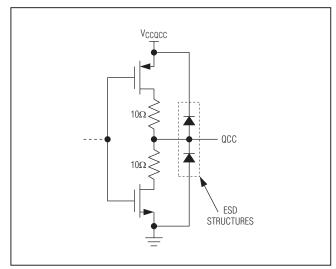


Figure 15. Equivalent LVCMOS Output Circuit

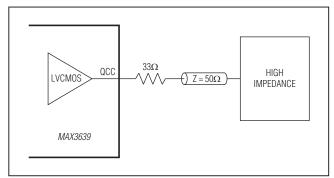


Figure 16. Interface to LVCMOS Output

Layout Considerations

The inputs and outputs are the most critical paths for the MAX3639; great care should be taken to minimize discontinuities on the transmission lines. Here are some suggestions for maximizing the performance of the MAX3639:

- An uninterrupted ground plane should be positioned beneath the clock outputs. The ground plane under the crystal should be removed to minimize capacitance.
- Supply decoupling capacitors should be placed close to the supply pins, preferably on the same side of the board as the MAX3639.
- Take care to isolate input traces from the MAX3639 outputs.

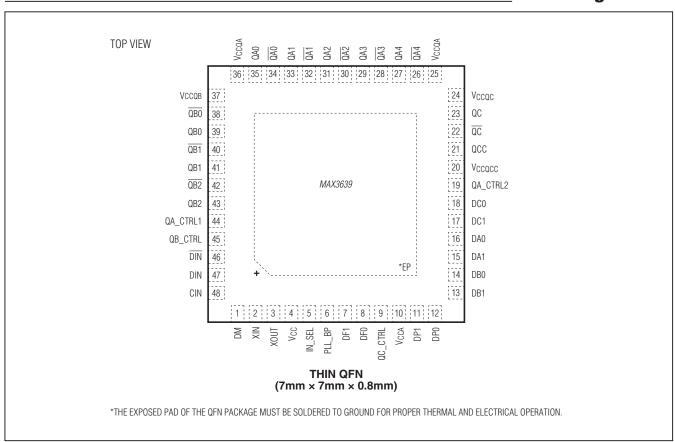
- The crystal, trace, and two external capacitors should be placed on the board as close as possible to the XIN and XOUT pins to reduce crosstalk of active signals into the oscillator.
- Maintain 100Ω differential (or 50Ω single-ended) transmission line impedance into and out of the part.
- Provide space between differential output pairs to reduce crosstalk, especially if the outputs are operating at different frequencies.
- Use multilayer boards with an uninterrupted ground plane to minimize EMI and crosstalk.

Refer to the MAX3639 evaluation kit for more information.

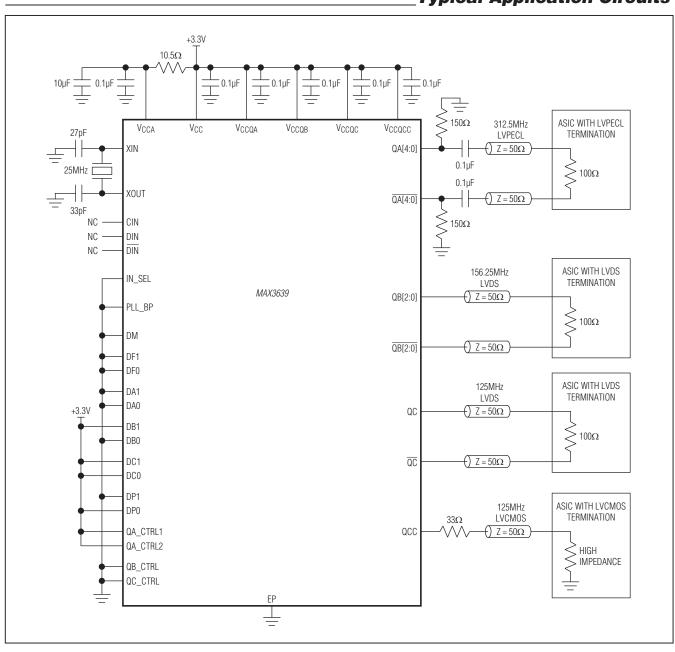
Chip Information

PROCESS: BICMOS

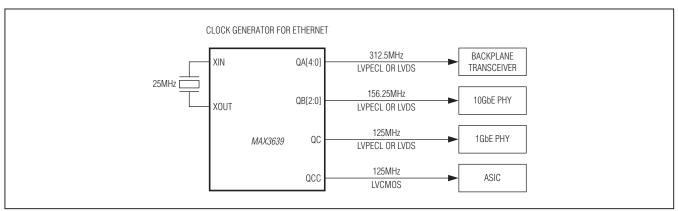
Pin Configuration

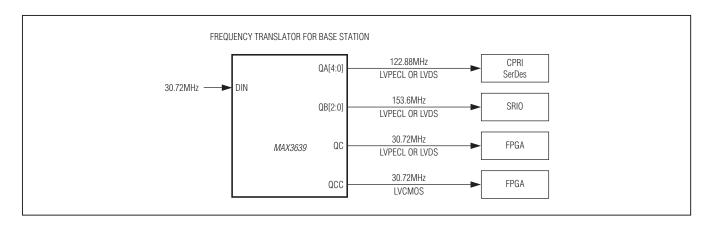


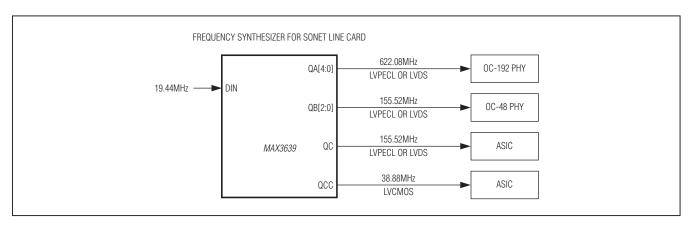
Typical Application Circuits



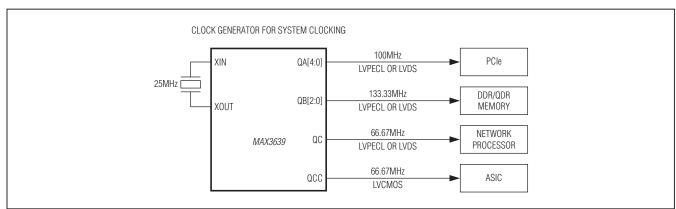
Typical Application Circuits (continued)

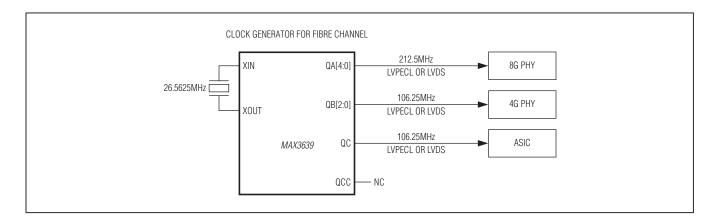






Typical Application Circuits (continued)





Package Information

For the latest package outline information and land patterns, go to http://www.microsemi.com. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

| PACKAGE TYPE | PACKAGE CODE | DOCUMENT NO. |
|--------------|--------------|----------------|
| 48 TQFN-EP | T4877+4 | <u>21-0144</u> |



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