

## XMC4100/XMC4200

### Microcontroller Series for Industrial Applications

ARM® Cortex®-M4

32-bit processor core

### About this Document

This datasheet is addressed to embedded hardware and software developers. It provides the reader with detailed descriptions about the ordering designations, available features, electrical and physical characteristics of the XMC4[12]00 series devices.

The document describes the characteristics of a superset of the XMC4[12]00 series devices. For simplicity, the various device types are referred to by the collective term XMC4[12]00 throughout this manual.

#### XMC4000 Family User Documentation

The set of user documentation includes:

- **Reference Manual**
  - describes the functionality of the superset of devices.
- **Datasheets**
  - list the complete ordering designations, available features and electrical characteristics of derivative devices.
- **Errata Sheets**
  - list deviations from the specifications given in the related Reference Manual or Datasheets. Errata Sheets are provided for the superset of devices.

**Attention:** Please consult all parts of the documentation set to attain consolidated knowledge about your device.

Application related guidance is provided by **Users Guides** and **Application Notes**.

Please refer to <http://www.infineon.com/xmc4000> to get access to the latest versions of those documents.

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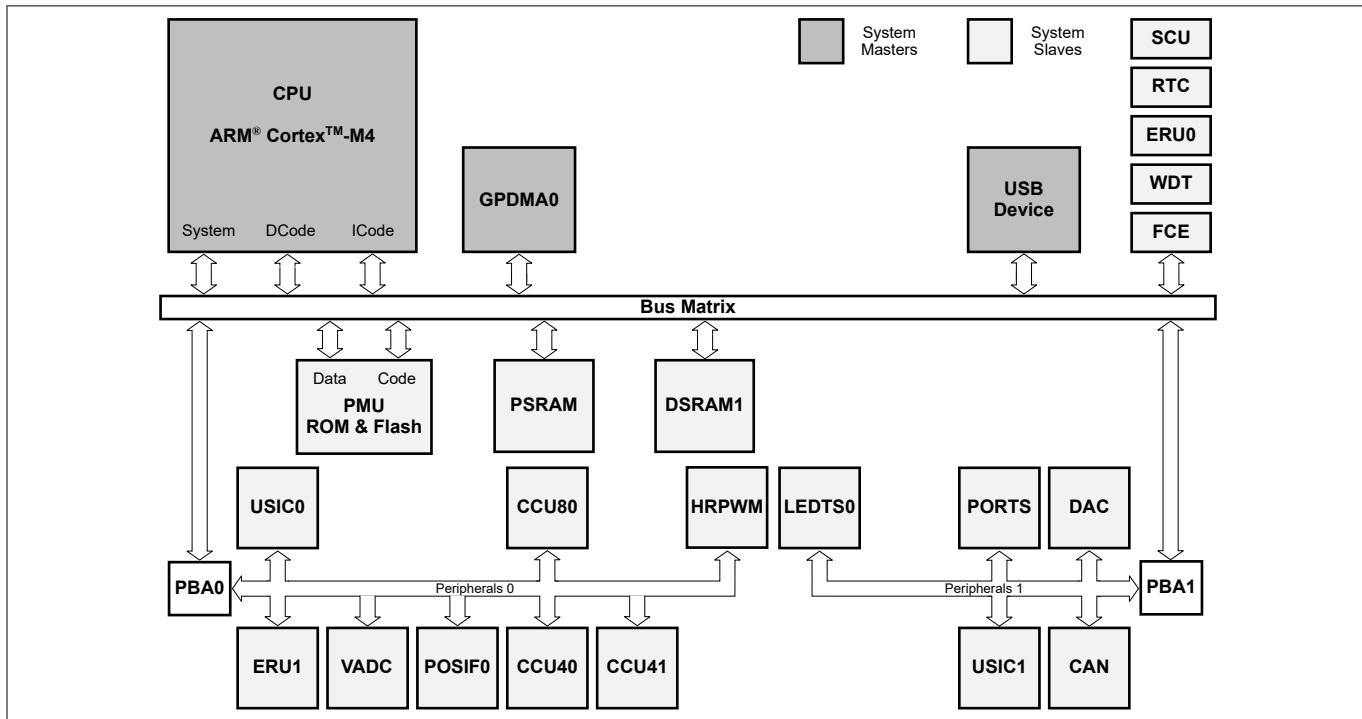
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## 1 Summary of Features

### 1 Summary of Features

The XMC4[12]00 devices are members of the XMC4000 Family of microcontrollers based on the ARM Cortex-M4 processor core. The XMC4000 is a family of high performance and energy efficient microcontrollers optimized for Industrial Connectivity, Industrial Control, Power Conversion, Sense & Control.



**Figure 1** System Block Diagram

#### CPU Subsystem

- CPU Core
  - High Performance 32-bit ARM Cortex-M4 CPU
  - 16-bit and 32-bit Thumb2 instruction set
  - DSP/MAC instructions
  - System timer (SysTick) for Operating System support
- Floating Point Unit
- Memory Protection Unit
- Nested Vectored Interrupt Controller
- One General Purpose DMA with up-to 8 channels
- Event Request Unit (ERU) for programmable processing of external and internal service requests
- Flexible CRC Engine (FCE) for multiple bit error detection

#### On-Chip Memories

- 16 KB on-chip boot ROM
- up to 16 KB on-chip high-speed program memory
- up to 24 KB on-chip high speed data memory
- up to 256 KB on-chip Flash Memory with 1 KB instruction cache

## 1 Summary of Features

### Communication Peripherals

- Universal Serial Bus, USB 2.0 device, with integrated PHY
- Controller Area Network interface (MultiCAN), Full-CAN/Basic-CAN with two nodes, 64 message objects (MO), data rate up to 1 MBit/s
- Four Universal Serial Interface Channels (USIC), providing four serial channels, usable as UART, double-SPI, quad-SPI, IIC, IIS and LIN interfaces
- LED and Touch-Sense Controller (LEDTS) for Human-Machine interface

### Analog Frontend Peripherals

- Two Analog-Digital Converters (VADC) of 12-bit resolution, 8 channels each, with input out-of-range comparators
- Digital-Analog Converter (DAC) with two channels of 12-bit resolution

### Industrial Control Peripherals

- Two Capture/Compare Units 4 (CCU4) for use as general purpose timers
- One Capture/Compare Units 8 (CCU8) for motor control and power conversion
- Four High Resolution PWM (HRPWM) channels
- One Position Interface (POSIF) for servo motor positioning
- Window Watchdog Timer (WDT) for safety sensitive applications
- Die Temperature Sensor (DTS)
- Real Time Clock module with alarm support
- System Control Unit (SCU) for system configuration and control

### Input/Output Lines

- Programmable port driver control module (PORTS)
- Individual bit addressability
- Tri-stated in input mode
- Push/pull or open drain output mode
- Boundary scan test support over JTAG interface

### On-Chip Debug Support

- Full support for debug features: 8 breakpoints, CoreSight, trace
- Various interfaces: ARM-JTAG, SWD, single wire trace

## 1 Summary of Features

### 1.1 Ordering Information

The ordering code for an Infineon microcontroller provides an exact reference to a specific product. The code “XMC4<DDD>-<Z><PPP><T><FFFF>” identifies:

- <DDD> the derivatives function set
- <Z> the package variant
  - E: LFBGA
  - F: LQFP, TQFP
  - Q: VQFN
- <PPP> package pin count
- <T> the temperature range:
  - F: -40°C to 85°C
  - K: -40°C to 125°C
- <FFFF> the Flash memory size.

For ordering codes for the XMC4[12]00 please contact your sales representative or local distributor.

This document describes several derivatives of the XMC4100 and XMC4200 series, some descriptions may not apply to a specific product. Please see [Table 1](#).

For simplicity the term **XMC4[12]00** is used for all derivatives throughout this document.

### 1.2 Device Types

These device types are available and can be ordered through Infineon's direct and/or distribution channels.

**Table 1 Synopsis of XMC4[12]00 Device Types**

Derivative <sup>1)</sup>	Package	Flash Kbytes	SRAM Kbytes
XMC4200-F64x256	PG-yQFP-64 <sup>2)</sup>	256	40
XMC4200-Q48x256	PG-VQFN-48	256	40
XMC4100-F64x128	PG-yQFP-64 <sup>2)</sup>	128	20
XMC4100-Q48x128	PG-VQFN-48	128	20
XMC4104-F64x64	PG-yQFP-64 <sup>2)</sup>	64	20
XMC4104-Q48x64	PG-VQFN-48	64	20
XMC4104-F64x128	PG-yQFP-64 <sup>2)</sup>	128	20
XMC4104-Q48x128	PG-VQFN-48	128	20
XMC4108-F64x64	PG-yQFP-64 <sup>2)</sup>	64	20
XMC4108-Q48x64	PG-VQFN-48	64	20

1) x is a placeholder for the supported temperature range.

2) y is a placeholder for the QFP package variant, LQFP or TQFP depending on the stepping, see [Section 1.3](#).

## 1 Summary of Features

### 1.3 Package Variants

Different markings of the XMC4[12]00 use different package variants. Details of those packages are given in the [Package Parameters](#) section of the datasheet.

**Table 2 XMC4[12]00 Package Variants**

Package Variant	Marking	Package
XMC4[12]00-F64	EES-AA, ES-AA, ES-AB, AB	PG-LQFP-64-19
XMC4[12]00-Q48		PG-VQFN-48-53
XMC4[12]00-F64	BA	PG-TQFP-64-19
XMC4[12]00-Q48		PG-VQFN-48-71

### 1.4 Device Type Features

The following table lists the available features per device type.

**Table 3 Features of XMC4[12]00 Device Types**

Derivative <sup>1)</sup>	LEDTS Intf.	USB Intf.	USIC Chan.	MultiCAN Nodes, MO
XMC4200-F64x256	1	1	2 x 2	N0, N1 MO[0..63]
XMC4200-Q48x256	1	1	2 x 2	N0, N1 MO[0..63]
XMC4100-F64x128	1	1	2 x 2	N0, N1 MO[0..63]
XMC4100-Q48x128	1	1	2 x 2	N0, N1 MO[0..63]
XMC4104-F64x64	1	-	2 x 2	-
XMC4104-Q48x64	1	-	2 x 2	-
XMC4104-F64x128	1	-	2 x 2	-
XMC4104-Q48x128	1	-	2 x 2	-
XMC4108-F64x64	-	-	2 x 2	N0, MO[0..31]
XMC4108-Q48x64	-	-	2 x 2	N0, MO[0..31]

1) x is a placeholder for the supported temperature range.

**Table 4 Features of XMC4[12]00 Device Types**

Derivative <sup>1)</sup>	ADC Chan.	DAC Chan.	CCU4 Slice	CCU8 Slice	POSIF Intf.	HRPWM Intf.
XMC4200-F64x256	10	2	2 x 4	1 x 4	1	1
XMC4200-Q48x256	9	2	2 x 4	1 x 4	1	1
XMC4100-F64x128	10	2	2 x 4	1 x 4	1	1

(table continues...)

## 1 Summary of Features

**Table 4 (continued) Features of XMC4[12]00 Device Types**

Derivative <sup>1)</sup>	ADC Chan.	DAC Chan.	CCU4 Slice	CCU8 Slice	POSIF Intf.	HRPWM Intf.
XMC4100-Q48x128	9	2	2 x 4	1 x 4	1	1
XMC4104-F64x64	10	2	2 x 4	1 x 4	1	1
XMC4104-Q48x64	9	2	2 x 4	1 x 4	1	1
XMC4104-F64x128	10	2	2 x 4	1 x 4	1	1
XMC4104-Q48x128	9	2	2 x 4	1 x 4	1	1
XMC4108-F64x64	10	2	2 x 4	1 x 4	1	-
XMC4108-Q48x64	9	2	2 x 4	1 x 4	1	-

1) x is a placeholder for the supported temperature range.

## 1.5 Definition of Feature Variants

The XMC4[12]00 types are offered with several memory sizes and number of available VADC channels. [Table 5](#) describes the location of the available Flash memory, [Table 6](#) describes the location of the available SRAMs, [Table 7](#) the available VADC channels.

**Table 5 Flash Memory Ranges**

Total Flash Size	Cached Range	Uncached Range
256 Kbytes	0800 0000 <sub>H</sub> – 0803 FFFF <sub>H</sub>	0C00 0000 <sub>H</sub> – 0C03 FFFF <sub>H</sub>
128 Kbytes	0800 0000 <sub>H</sub> – 0801 FFFF <sub>H</sub>	0C00 0000 <sub>H</sub> – 0C01 FFFF <sub>H</sub>
64 Kbytes	0800 0000 <sub>H</sub> – 0800 FFFF <sub>H</sub>	0C00 0000 <sub>H</sub> – 0C00 FFFF <sub>H</sub>

**Table 6 SRAM Memory Ranges**

Total SRAM Size	Program SRAM	System Data SRAM
40 Kbytes	1FFF C000 <sub>H</sub> – 1FFF FFFF <sub>H</sub>	2000 0000 <sub>H</sub> – 2000 5FFF <sub>H</sub>
20 Kbytes	1FFF E000 <sub>H</sub> – 1FFF FFFF <sub>H</sub>	2000 0000 <sub>H</sub> – 2000 2FFF <sub>H</sub>

**Table 7 ADC Channels<sup>1)</sup>**

Package	VADC G0	VADC G1
LQFP-64, TQFP-64	CH0, CH3..CH7	CH0, CH1, CH3, CH6
PG-VQFN-48	CH0, CH3..CH7	CH0, CH1, CH3

1) Some pins in a package may be connected to more than one channel. For the detailed mapping see the Port I/O Function table.

## 1 Summary of Features

### 1.6 Identification Registers

The identification registers allow software to identify the marking.

**Table 8 XMC4200 Identification Registers**

Register Name	Value	Marking
SCU_IDCHIP	0004 2001 <sub>H</sub>	EES-AA, ES-AA
SCU_IDCHIP	0004 2002 <sub>H</sub>	ES-AB, AB
SCU_IDCHIP	0004 2003 <sub>H</sub>	BA
JTAG IDCODE	101D D083 <sub>H</sub>	EES-AA, ES-AA
JTAG IDCODE	201D D083 <sub>H</sub>	ES-AB, AB
JTAG IDCODE	301D D083 <sub>H</sub>	BA

**Table 9 XMC4100 Identification Registers**

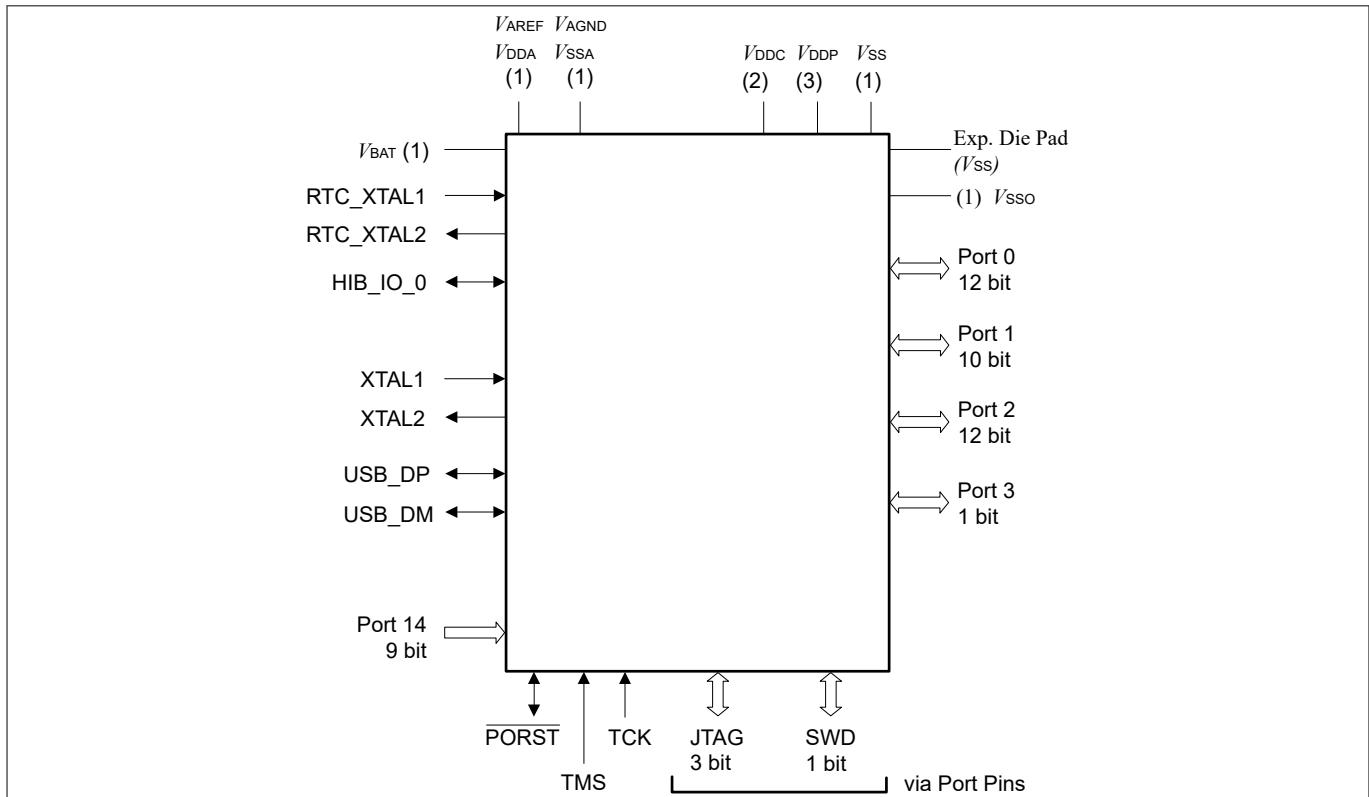
Register Name	Value	Marking
SCU_IDCHIP	0004 2001 <sub>H</sub>	EES-AA, ES-AA
SCU_IDCHIP	0004 2002 <sub>H</sub>	ES-AB, AB
SCU_IDCHIP	0004 1003 <sub>H</sub>	BA
JTAG IDCODE	101D D083 <sub>H</sub>	EES-AA, ES-AA
JTAG IDCODE	201D D083 <sub>H</sub>	ES-AB, AB
JTAG IDCODE	301D D083 <sub>H</sub>	BA

## 2 General Device Information

### 2 General Device Information

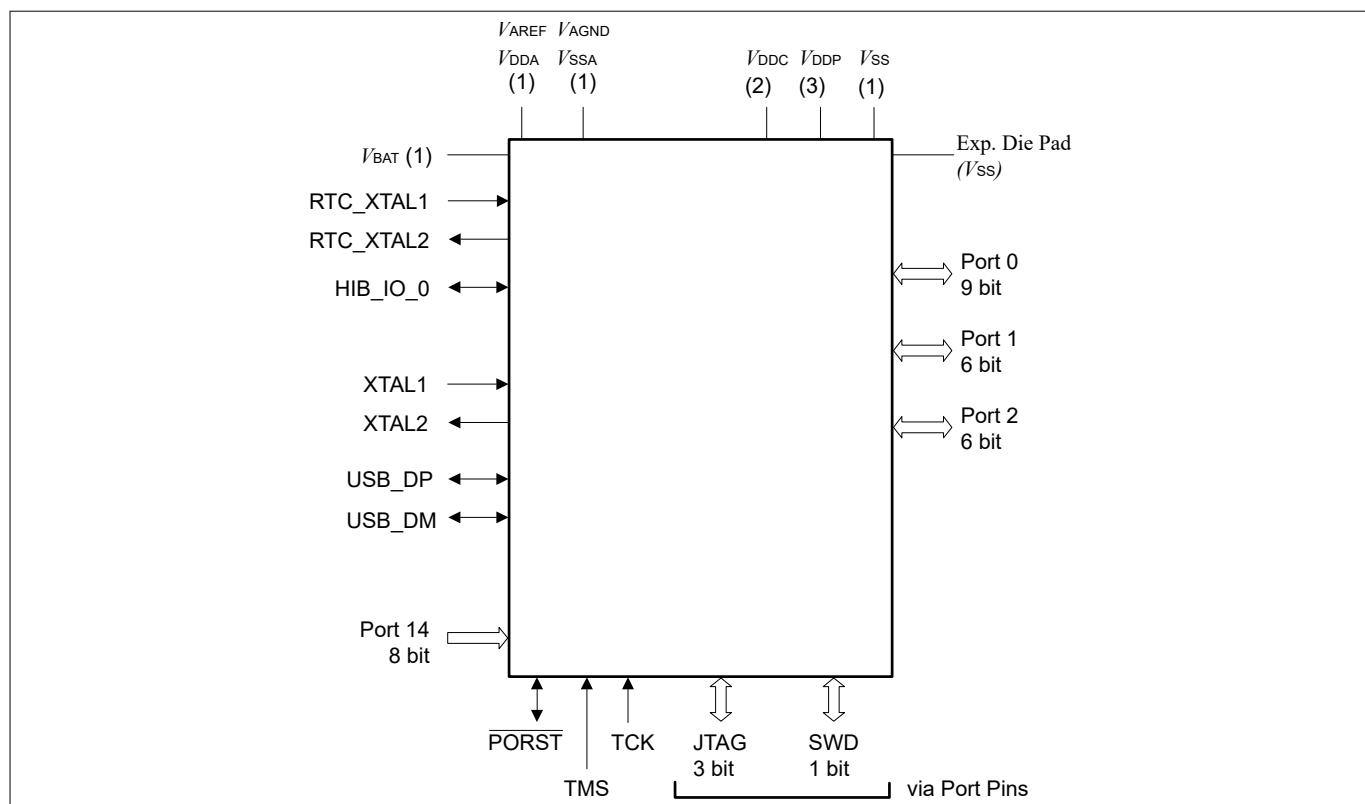
This section summarizes the logic symbols and package pin configurations with a detailed list of the functional I/O mapping.

#### 2.1 Logic Symbols



**Figure 2 XMC4[12]00 Logic Symbol PG-LQFP-64 and PG-TQFP-64**

## 2 General Device Information

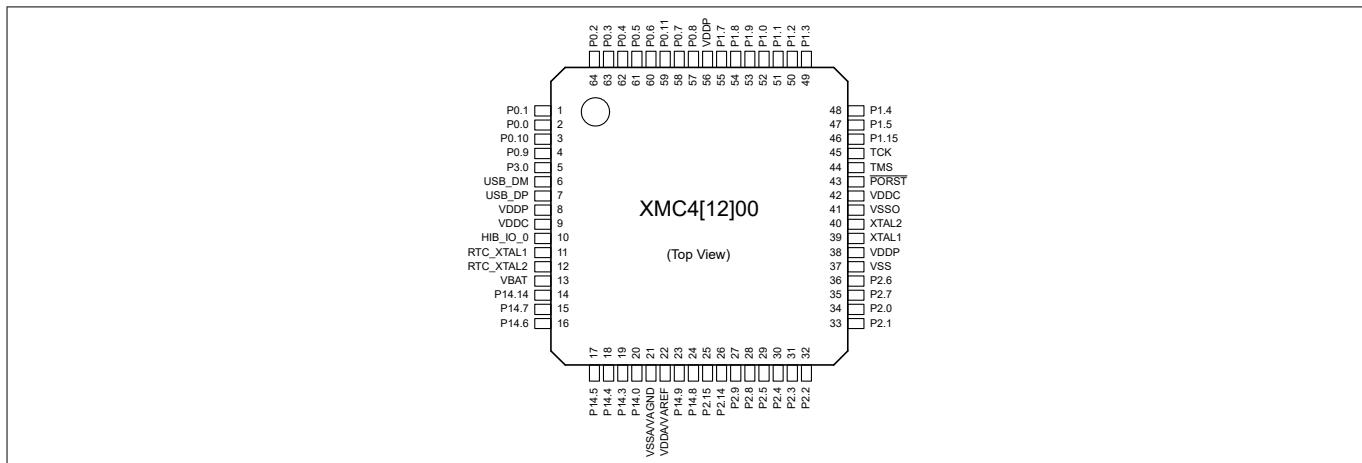


**Figure 3** XMC4[12]00 Logic Symbol PG-VQFN-48

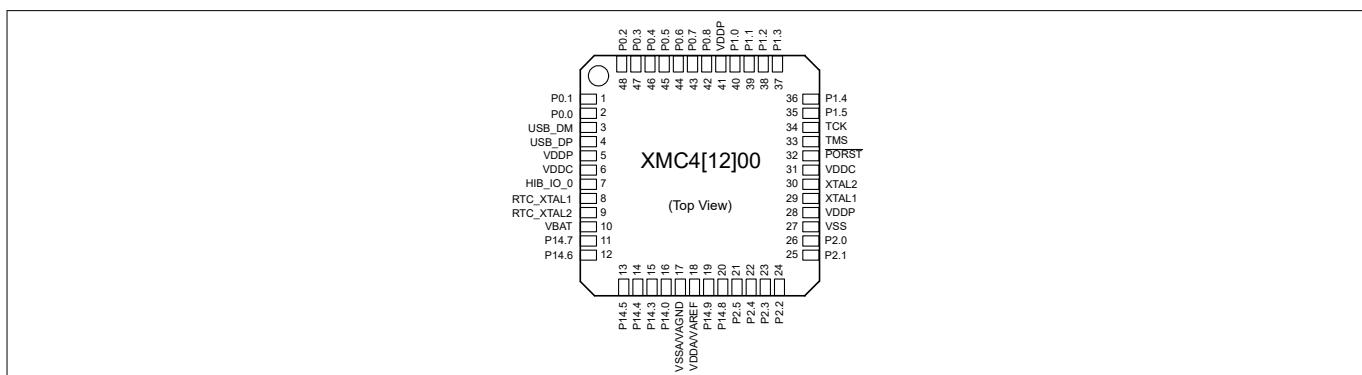
## 2 General Device Information

### 2.2 Pin Configuration and Definition

The following figures summarize all pins, showing their locations on the different packages.



**Figure 4 XMC4[12]00 PG-LQFP-64 and PG-TQFP-64 Pin Configuration (top view)**



**Figure 5 XMC4[12]00 PG-VQFN-48 Pin Configuration (top view)**

### 2.2.1 Package Pin Summary

The following general scheme is used to describe each pin:

**Table 10 Package Pin Mapping Description**

Function	Package A	Package B	...	Pad Type	Notes
Name	N	Ax	...	A1+	

The table is sorted by the “Function” column, starting with the regular Port pins (Px.y), followed by the dedicated pins (i.e. PORST) and supply pins.

The following columns, titled with the supported package variants, lists the package pin number to which the respective function is mapped in that package.

The “Pad Type” indicates the employed pad type (A1, A1+, special=special pad, In=input pad, AN/DIG\_IN=analog and digital input, Power=power supply). Details about the pad properties are defined in the Electrical Parameters.

In the “Notes”, special information to the respective pin/function is given, i.e. deviations from the default configuration after reset. Per default the regular Port pins are configured as direct input with no internal pull device active.

## 2 General Device Information

**Table 11 Package Pin Mapping**

Function	LQFP-64 TQFP-64	VQFN-48	Pad Type	Notes
P0.0	2	2	A1+	
P0.1	1	1	A1+	
P0.2	64	48	A1+	
P0.3	63	47	A1+	
P0.4	62	46	A1+	
P0.5	61	45	A1+	
P0.6	60	44	A1+	
P0.7	58	43	A1+	After a system reset, via HWSEL this pin selects the DB.TDI function.
P0.8	57	42	A1+	After a system reset, via HWSEL this pin selects the DB.TRST function, with a weak pull-down active.
P0.9	4	–	A1+	
P0.10	3	–	A1+	
P0.11	59	–	A1+	
P1.0	52	40	A1+	
P1.1	51	39	A1+	
P1.2	50	38	A1+	
P1.3	49	37	A1+	
P1.4	48	36	A1+	
P1.5	47	35	A1+	
P1.7	55	–	A1+	
P1.8	54	–	A1+	
P1.9	53	–	A1+	
P1.15	46	–	A1+	
P2.0	34	26	A1+	
P2.1	33	25	A1+	After a system reset, via HWSEL this pin selects the DB.TDO function.
P2.2	32	24	A1+	
P2.3	31	23	A1+	
P2.4	30	22	A1+	
P2.5	29	21	A1+	
P2.6	36	–	A1+	
P2.7	35	–	A1+	

(table continues...)

## 2 General Device Information

**Table 11** (continued) Package Pin Mapping

Function	LQFP-64 TQFP-64	VQFN-48	Pad Type	Notes
P2.8	28	–	A1+	
P2.9	27	–	A1+	
P2.14	26	–	A1+	
P2.15	25	–	A1+	
P3.0	5	–	A1+	
P14.0	20	16	AN/DIG_IN	
P14.3	19	15	AN/DIG_IN	
P14.4	18	14	AN/DIG_IN	
P14.5	17	13	AN/DIG_IN	
P14.6	16	12	AN/DIG_IN	
P14.7	15	11	AN/DIG_IN	
P14.8	24	20	AN/DAC/DIG_IN	
P14.9	23	19	AN/DAC/DIG_IN	
P14.14	14	–	AN/DIG_IN	
USB_DP	7	4	special	
USB_DM	6	3	special	
HIB_IO_0	10	7	A1 special	At the first power-up and with every reset of the hibernate domain this pin is configured as open-drain output and drives "0". As output the medium driver mode is active.
TCK	45	34	A1	Weak pull-down active.
TMS	44	33	A1+	Weak pull-up active. As output the strong-soft driver mode is active.
PORST	43	32	special	Strong pull-down controlled by EVR. Weak pull-up active while strong pull-down is not active.
XTAL1	39	29	clock_IN	
XTAL2	40	30	clock_O	
RTC_XTAL1	11	8	clock_IN	
RTC_XTAL2	12	9	clock_O	
VBAT	13	10	Power	When VDDP is supplied VBAT has to be supplied as well.

(table continues...)

## 2 General Device Information

**Table 11** (continued) Package Pin Mapping

Function	LQFP-64 TQFP-64	VQFN-48	Pad Type	Notes
VDDA/VAREF	22	18	AN_Power/ AN_Ref	Shared analog supply and reference voltage pin.
VSSA/VAGND	21	17	AN_Power/ AN_Ref	Shared analog supply and reference ground pin.
VDDC	9	6	Power	
VDDC	42	31	Power	
VDDP	8	5	Power	
VDDP	38	28	Power	
VDDP	56	41	Power	
VSS	37	27	Power	
VSSO	41	–	Power	
VSS	Exp. Pad	Exp. Pad	Power	<p><b>Exposed Die Pad</b></p> <p>The exposed die pad is connected internally to VSS. For proper operation, it is mandatory to connect the exposed pad directly to the common ground on the board.</p> <p>For thermal aspects, please refer to the Datasheet. Board layout examples are given in an application note.</p>

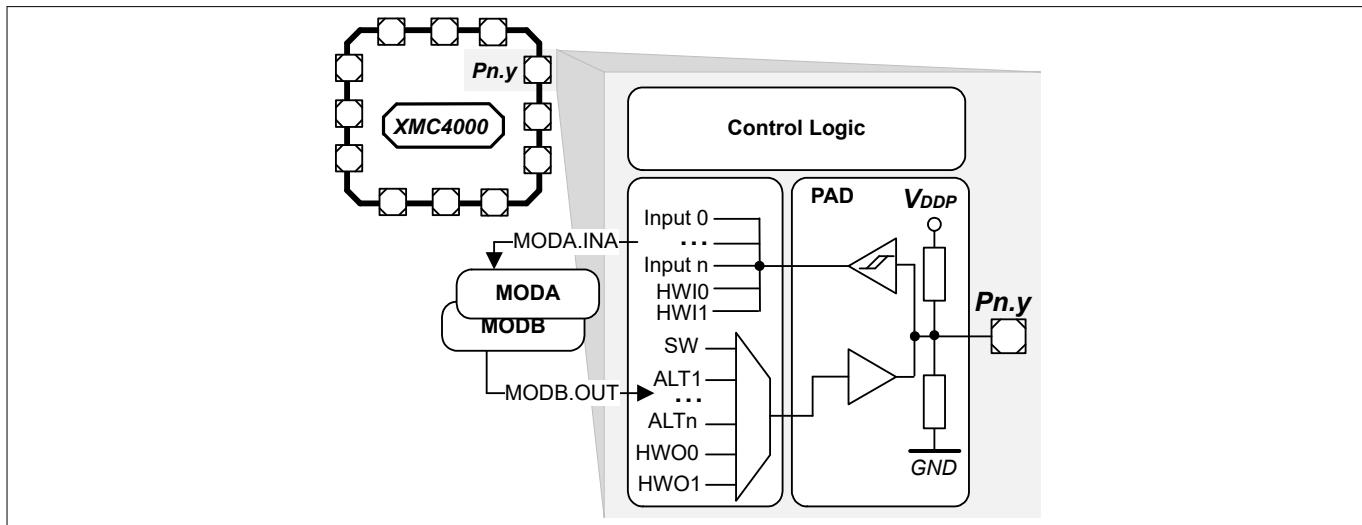
## 2 General Device Information

### 2.2.2 Port I/O Functions

The following general scheme is used to describe each PORT pin:

**Table 12** Port I/O Function Description

Function	Outputs			Inputs		
	ALT1	ALTn	HWO0	HWI0	Input	Input
P0.0		MODA.OUT	MODB.OUT	MODB.INA	MODC.INA	
Pn.y	MODA.OUT				MODA.INA	MODC.INB



**Figure 6** Simplified Port Structure

Pn.y is the port pin name, defining the control and data bits/registers associated with it. As GPIO, the port is under software control. Its input value is read via Pn\_IN.y, Pn\_OUT defines the output value.

Up to four alternate output functions (ALT1/2/3/4) can be mapped to a single port pin, selected by Pn\_IOCR.PC. The output value is directly driven by the respective module, with the pin characteristics controlled by the port registers (within the limits of the connected pad).

The port pin input can be connected to multiple peripherals. Most peripherals have an input multiplexer to select between different possible input sources.

The input path is also active while the pin is configured as output. This allows to feedback an output to on-chip resources without wasting an additional external pin.

By Pn\_HWSEL it is possible to select between different hardware “masters” (HWO0/HWI0). The selected peripheral can take control of the pin(s). Hardware control overrules settings in the respective port pin registers.

## 2 General Device Information

### 2.2.2.1 Port I/O Function Table

**Table 13** Port I/O Functions

Function	Output				Input				Input
	ALT1	ALT2	ALT3	ALT4	HWO0	HWI0	Input	Input	
P0.0	CAN. N0_TXD	CCU80. OUT21	LEDTS0. COL2			U1C1. DX0D		ERU0.0 B0	USB.VB USDETE CTA
P0.1	U1C1. DOUT0	CCU80. OUT11	LEDTS0. COL3					ERU0.0 A0	HRPWM 0.C1INB
P0.2	U1C1. SEL01	CCU80. OUT01	HRPWM 0.HROU T01		U1C0. DOUT3	U1C0. HWIN3		ERU0.3 B3	HRPWM 0.C2INB
P0.3		CCU80. OUT20	HRPWM 0.HROU T20		U1C0. DOUT2	U1C0. HWIN2			ERU1.3 B0
P0.4		CCU80. OUT10	HRPWM 0.HROU T21		U1C0. DOUT1	U1C0. HWIN1		U1C0.D X0A	ERU0.2 B3
P0.5	U1C0. DOUT0	CCU80. OUT00	HRPWM 0.HROU T00		U1C0. DOUT0	U1C0. HWINO		U1C0.D X0B	ERU1.3 A0
P0.6	U1C0. SEL00	CCU80. OUT30	HRPWM 0.HROU T30					U1C0.D X2A	CCU80.I N2B
P0.7	WWDT.S ERVICE _OUT	U0C0. SEL00			HRPWM 0.HROU T11	DB.TDI	U0C0. DX2B	ERU0.2 B1	CCU80.I N0A
P0.8	SCU.EX TCLK	U0C0. SCLK0 UT			HRPWM 0.HROU T10	<u>DB.TRST</u>	U0C0. DX1B	ERU0.2 A1	CCU80.I N1B

(table continues...)

## 2 General Device Information

**Table 13 (continued) Port I/O Functions**

Function	Output	Input										
		ALT1	ALT2	ALT3	ALT4	HWIO	Input	Input	Input	Input	Input	Input
P0.9	HRPWM 0.HROU T31	U1C1. SEL00	CCU80. OUT12	LEDTS0. COL0		U1C1. DX2A			ERU0.1 B0			
P0.10		U1C1. SCLK0 UT	CCU80. OUT02	LEDTS0. COL1		U1C1. DX1A			ERU0.1 A0			
P0.11		U1C0. SCLK0 UT	CCU80. OUT31				U1C0.D X1A	ERU0.3 A2				
P1.0		U0C0. SEL00	CCU40. OUT3	ERU1.P DOUT3		U0C0. DX2A			ERU0.3 B0		CCU40.I N3A	HRPWM 0.C0INA
P1.1		U0C0. SCLK0 UT	CCU40. OUT2	ERU1.P DOUT2		U0C0. DX1A			ERU0.3 A0		CCU40.I N2A	HRPWM 0.C1INA
P1.2			CCU40. OUT1	ERU1.P DOUT1	U0C0. DOUT3	U0C0. HWIN3			POSIFO. IN1A		ERU1.2 B0	CCU40.I N1A
P1.3		U0C0. MCLK0 UT	CCU40. OUT0	ERU1.P DOUT0	U0C0. DOUT2	U0C0. HWIN2			POSIFO. IN0A		ERU1.2 A0	CCU40.I N0A
P1.4	WWDT.S ERVICE _OUT	CAN. NO_TXD	CCU80. OUT33		U0C0. DOUT1	U0C0. HWIN1	U0C0. DX0B	CAN.N1 _RXDD	ERU0.2 B0		CCU41.I N0C	HRPWM 0.BL0A
P1.5	CAN.N1 _TXD	U0C0. DOUT0	CCU80. OUT23		U0C0. DOUT0	U0C0. HWIN0	U0C0. DX0A	CAN.NO _RXDA	ERU1.0 A0	CCU41.I N1C		

(table continues...)

## 2 General Device Information

**Table 13 (continued) Port I/O Functions**

Function	Output	Input										
		ALT1	ALT2	ALT3	ALT4	HWO0	HWI0	Input	Input	Input	Input	Input
P1.7	U0C0. DOUT0			U1C1.S EL02						USB.VB USDETE CTB		
P1.8	U0C0. SEL01			U1C1.S CLKOU T								
P1.9	U0C0.S CLKOU T			U1C1.D OUT0								
P1.15	SCU.EX TCLK			U1C0.D OUT0						ERU1.1 A0		
P2.0	CAN.N0 _TXD			LEDTS0. COL1						ERU0.0 B3	CCU40.I N1C	
P2.1				LEDTS0. COL0	DB.TDO / TRACES WO					ERU1.0 B0	CCU40.I N0C	
P2.2	VADC.E MUX00			CCU41. OUT3	LEDTS0. LINE0	LEDTS0. EXTEND ED0	LEDTS0. TSIN0A	U0C1.D X0A	ERU0.1 B2		CCU41.I N3A	
P2.3	VADC.E MUX01	U0C1. SEL00	CCU41. OUT2	LEDTS0. LINE1	LEDTS0. EXTEND ED1	LEDTS0. TSIN1A	U0C1.D X2A	ERU0.1 A2			CCU41.I N2A	
P2.4	VADC.E MUX02	U0C1. SCLKO UT	CCU41. OUT1	LEDTS0. LINE2	LEDTS0. EXTEND ED2	LEDTS0. TSIN2A	U0C1.D X1A	ERU0.0 B2			CCU41.I N1A	HRPWM 0.BL1A

(table continues...)

## 2 General Device Information

Table 13 (continued) Port I/O Functions

Function	Output	HW00				Input				Input	Input	Input
		ALT1	ALT2	ALT3	ALT4	Input	Input	Input	Input			
P2.5	U0C1. DOUT0	CCU41. OUT0	LEDTS0. LINE3	LEDTS0. EXTEND ED3	LEDTS0. TSIN3A	UOC1.D X0B	ERU0.0 A2	CCU41.I N0A	HRPWM 0.BL2A			
P2.6		CCU80. OUT13	LEDTS0. COL3			CAN.N1 _RXDA	ERU0.1 B3	CCU40.I N3C				
P2.7	CAN. N1_TxD	CCU80. OUT03	LEDTS0. COL2					ERU1.1 B0	CCU40.I N2C			
P2.8		CCU80. OUT32	LEDTS0. LINE4	LEDTS0. EXTEND ED4	LEDTS0. TSIN4A	DAC.TRI GGER5			CCU40.I N0B	CCU40.I N1B	CCU40.I N2B	CCU40.I N3B
P2.9		CCU80. OUT22	LEDTS0. LINE5	LEDTS0. EXTEND ED5	LEDTS0. TSIN5A	DAC.TRI GGER4			CCU41.I N0B	CCU41.I N1B	CCU41.I N2B	CCU41.I N3B
P2.14	VADC.E MUX11	U1C0. DOUT0	CCU80. OUT21				U1C0.D X0D					
P2.15	VADC.E MUX12		CCU80. OUT11	LEDTS0. LINE6	LEDTS0. EXTEND ED6	LEDTS0. TSIN6A	U1C0.D X0C					
P3.0	U0C1. SCLK0 UT					U0C1.D X1B			CCU80.I N2C			
P14.0						VADC.G 0CH0						
P14.3						VADC.G 0CH3	VADC.G 1CH3	CAN.N0 _RXDB				

(table continues...)

## 2 General Device Information

**Table 13** (continued) Port I/O Functions

Function	Output	Input											
		ALT1	ALT2	ALT3	ALT4	HWO0	HWI0	Input	Input	Input	Input	Input	Input
P14.4								VADC.G 0CH4					
P14.5								VADC.G 0CH5					
P14.6								VADC.G 0CH6					
P14.7								VADC.G 0CH7					
P14.8						DAC.OU T_0		VADC.G 1CH0					
P14.9						DAC.OU T_1		VADC.G 1CH1					
P14.14								VADC.G 1CH6					
USB_DP													
USB_DM													
HIB_IO_0	HIBOUT	WWDT. SERVIC E_OUT						WAKEU PA					
TCK								DB.TCK/ SWCLK					
TMS								DB.TMS /SWDIO					
<b>PORST</b>													

(table continues...)

## 2 General Device Information

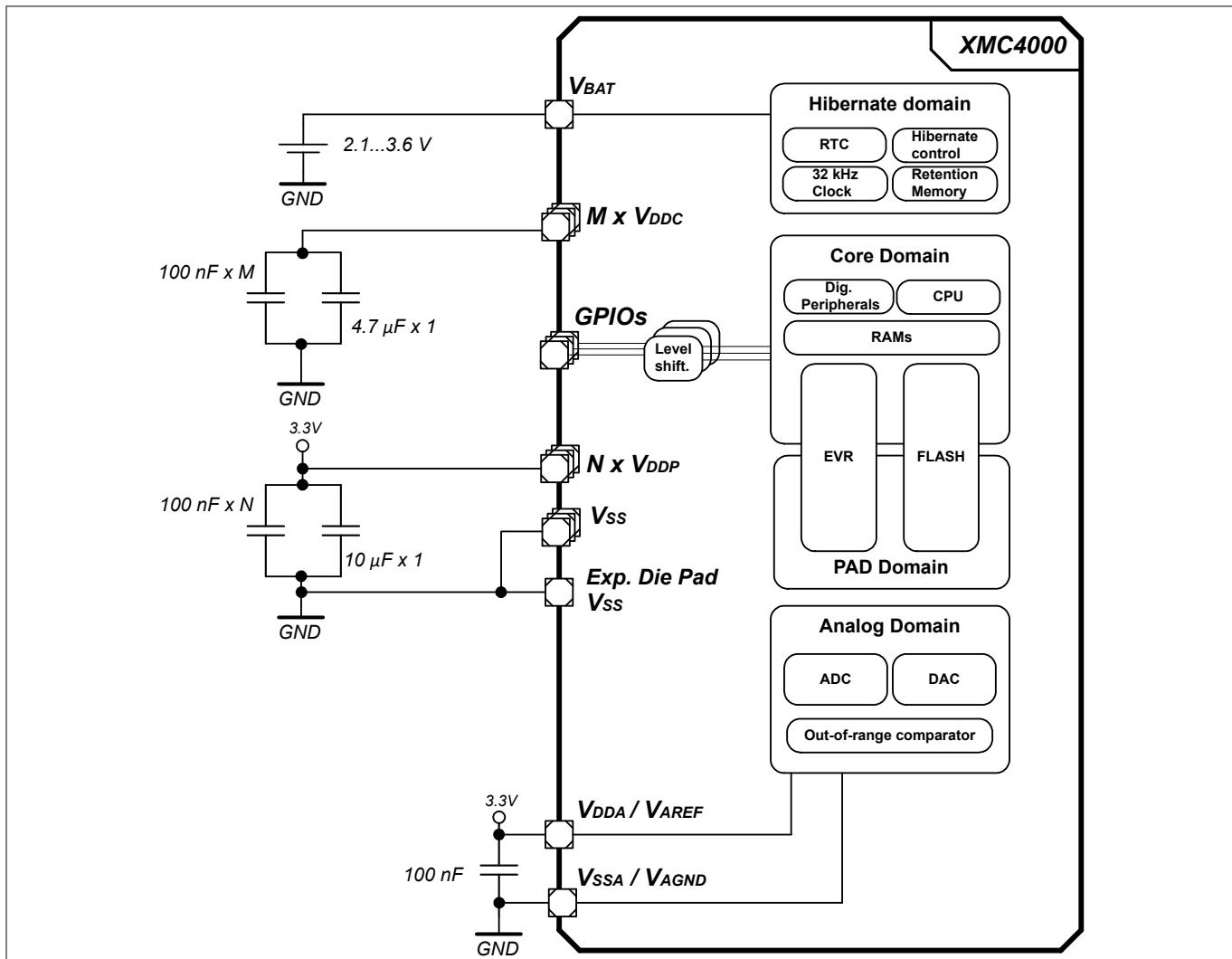
**Table 13** (continued) Port I/O Functions

Function	Output	Input													
		ALT1	ALT2	ALT3	ALT4	HWO0	HWO1	UOC0.D X0F	UOC1.D X0F	U1C0.D X0F	U1C1.D X0F	Input	Input	Input	Input
XTAL1															
XTAL2															
RTC_XTA L1												ERU0.1 B1			
RTC_XTA L2															

## 2 General Device Information

### 2.3 Power Connection Scheme

Figure 7 shows a reference power connection scheme for the XMC4[12]00.



**Figure 7** Power Connection Scheme

Every power supply pin needs to be connected. Different pins of the same supply need also to be externally connected. As example, all  $V_{DDP}$  pins must be connected externally to one  $V_{DDP}$  net. In this reference scheme one 100 nF capacitor is connected at each supply pin against  $V_{SS}$ . An additional 10  $\mu$ F capacitor is connected to the  $V_{DDP}$  nets and an additional 4.7  $\mu$ F capacitor to the  $V_{DDC}$  nets.

The XMC4[12]00 has a common ground concept, all  $V_{SS}$ ,  $V_{SSA}$  and  $V_{SSO}$  pins share the same ground potential. In packages with an exposed die pad it must be connected to the common ground as well.

There are no dedicated connections for the analog reference  $V_{AREF}$  and  $V_{AGND}$ . Instead, they share the same pins as the analog supply pins  $V_{DDA}$  and  $V_{SSA}$ . Some analog channels can optionally serve as “Alternate Reference”; further details on this operating mode are described in the Reference Manual.

When  $V_{DDP}$  is supplied,  $V_{BAT}$  must be supplied as well. If no other supply source (e.g. battery) is connected to  $V_{BAT}$ , the  $V_{BAT}$  pin can also be connected directly to  $V_{DDP}$ .

### 3 Electrical Parameters

## 3 Electrical Parameters

### 3.1 General Parameters

#### 3.1.1 Parameter Interpretation

The parameters listed in this section partly represent the characteristics of the XMC4[12]00 and partly its requirements on the system. To aid interpreting the parameters easily when evaluating them for a design, they are marked with an two-letter abbreviation in column “Symbol”:

- **CC**

Such parameters indicate **Controller Characteristics**, which are a distinctive feature of the XMC4[12]00 and must be regarded for system design.

- **SR**

Such parameters indicate **System Requirements**, which must be provided by the application system in which the XMC4[12]00 is designed in.

#### 3.1.2 Absolute Maximum Ratings

Stresses above the values listed under “Absolute Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions may affect device reliability.

**Table 14**      **Absolute Maximum Rating Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Storage temperature	$T_{ST}$ SR	-65	-	150	°C	-
Junction temperature	$T_J$ SR	-40	-	150	°C	-
Voltage at 3.3 V power supply pins with respect to $V_{SS}$	$V_{DDP}$ SR	-	-	4.3	V	-
Voltage on any Class A and dedicated input pin with respect to $V_{SS}$	$V_{IN}$ SR	-1.0	-	$V_{DDP} + 1.0$ or max. 4.3	V	whichever is lower
Voltage on any analog input pin with respect to $V_{AGND}$	$V_{AIN}$ $V_{AREF}$ SR	-1.0	-	$V_{DDP} + 1.0$ or max. 4.3	V	whichever is lower
Input current on any pin during overload condition	$I_{IN}$ SR	-10	-	+10	mA	
Absolute maximum sum of all input circuit currents for one port group during overload condition <sup>1)</sup>	$\Sigma I_{IN}$ SR	-25	-	+25	mA	

(table continues...)

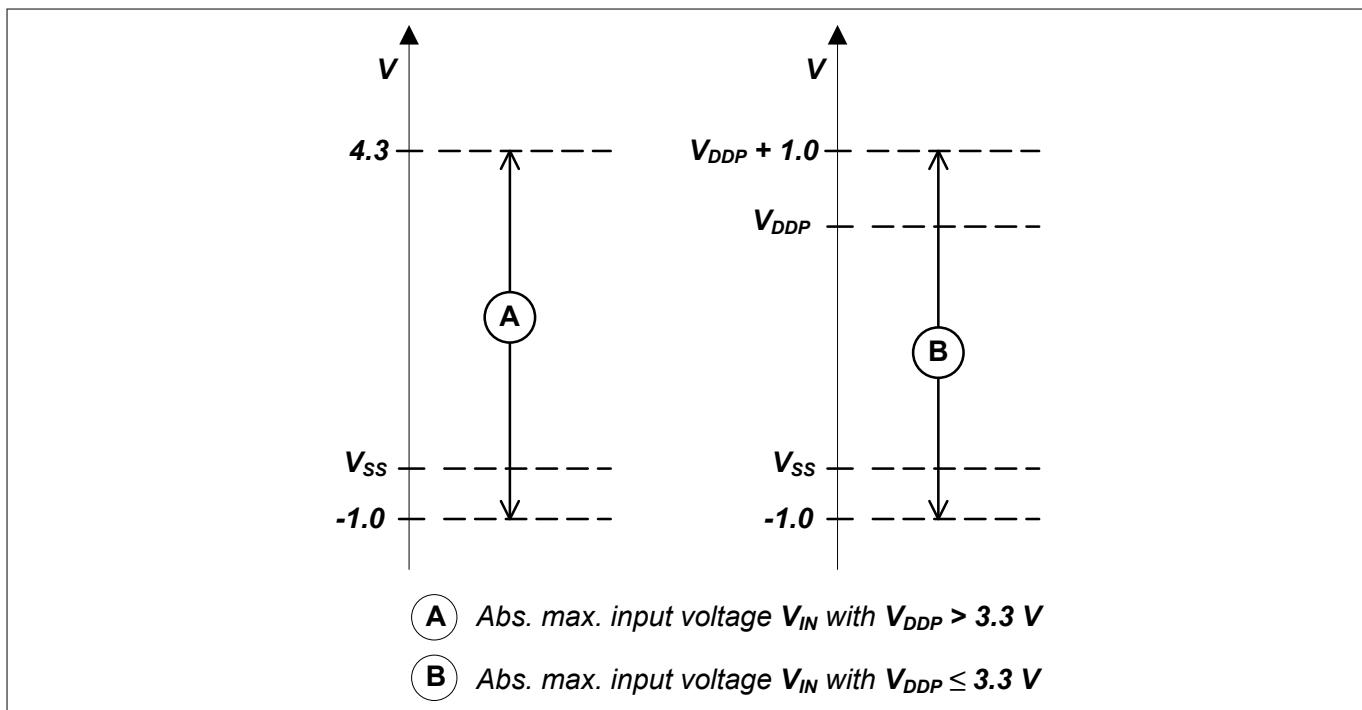
### 3 Electrical Parameters

**Table 14** (continued) Absolute Maximum Rating Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Absolute maximum sum of all input circuit currents during overload condition	$\Sigma I_{IN SR}$	-100	-	+100	mA	

1) The port groups are defined in [Table 18](#).

[Figure 8](#) explains the input voltage ranges of  $V_{IN}$  and  $V_{AIN}$  and its dependency to the supply level of  $V_{DDP}$ . The input voltage must not exceed 4.3 V, and it must not be more than 1.0 V above  $V_{DDP}$ . For the range up to  $V_{DDP} + 1.0$  V also see the definition of the overload conditions in [Pin Reliability in Overload](#).



**Figure 8**

**Absolute Maximum Input Voltage Ranges**

### 3 Electrical Parameters

#### 3.1.3 Pin Reliability in Overload

When receiving signals from higher voltage devices, low-voltage devices experience overload currents and voltages that go beyond their own IO power supplies specification.

[Table 15](#) defines overload conditions that will not cause any negative reliability impact if all the following conditions are met:

- full operation life-time is not exceeded
- [Operating Conditions](#) are met for
  - pad supply levels ( $V_{DDP}$  or  $V_{DDA}$ )
  - temperature

If a pin current is outside of the [Operating Conditions](#) but within the overload parameters, then the parameters functionality of this pin as stated in the Operating Conditions can no longer be guaranteed. Operation is still possible in most cases but with relaxed parameters.

**Note:** An overload condition on one or more pins does not require a reset.

**Note:** A series resistor at the pin to limit the current to the maximum permitted overload current is sufficient to handle failure situations like short to battery.

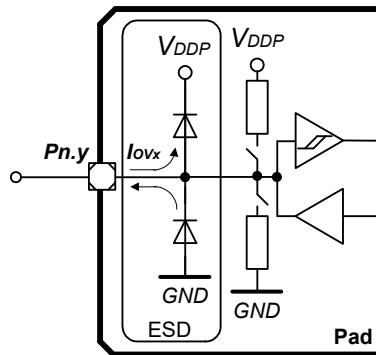
**Table 15** Overload Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Input current on any port pin during overload condition	$I_{OV}$ SR	-5	-	5	mA	
Absolute sum of all input circuit currents for one port group during overload condition <sup>1)</sup>	$I_{OVG}$ SR	-	-	20	mA	$\sum I_{OVX} $ , for all $I_{OVX} < 0$ mA
		-	-	20	mA	$\sum I_{OVX} $ , for all $I_{OVX} > 0$ mA
Absolute sum of all input circuit currents during overload condition	$I_{OVS}$ SR	-	-	80	mA	$\Sigma I_{OVG}$

1) The port groups are defined in [Table 18](#).

[Figure 9](#) shows the path of the input currents during overload via the ESD protection structures. The diodes against  $V_{DDP}$  and ground are a simplified representation of these ESD protection structures.

### 3 Electrical Parameters



**Figure 9** Input Overload Current via ESD structures

Table 16 and Table 17 list input voltages that can be reached under overload conditions. Note that the absolute maximum input voltages as defined in the [Absolute Maximum Ratings](#) must not be exceeded during overload.

**Table 16** PN-Junction Characteristics for positive Overload

Pad Type	$I_{OV} = 5 \text{ mA}, T_J = -40^\circ\text{C}$	$I_{OV} = 5 \text{ mA}, T_J = 150^\circ\text{C}$
A1/A1+	$V_{IN} = V_{DDP} + 1.0 \text{ V}$	$V_{IN} = V_{DDP} + 0.75 \text{ V}$
AN/DIG_IN	$V_{IN} = V_{DDP} + 1.0 \text{ V}$	$V_{IN} = V_{DDP} + 0.75 \text{ V}$

**Table 17** PN-Junction Characteristics for negative Overload

Pad Type	$I_{OV} = 5 \text{ mA}, T_J = -40^\circ\text{C}$	$I_{OV} = 5 \text{ mA}, T_J = 150^\circ\text{C}$
A1/A1+	$V_{IN} = V_{SS} - 1.0 \text{ V}$	$V_{IN} = V_{SS} - 0.75 \text{ V}$
AN/DIG_IN	$V_{IN} = V_{DDP} - 1.0 \text{ V}$	$V_{IN} = V_{DDP} - 0.75 \text{ V}$

**Table 18** Port Groups for Overload and Short-Circuit Current Sum Parameters

Group	Pins
1	P0.[12:0], P3.0
2	P14.[8:0]
3	P2.[15:0]
4	P1.[15:0]

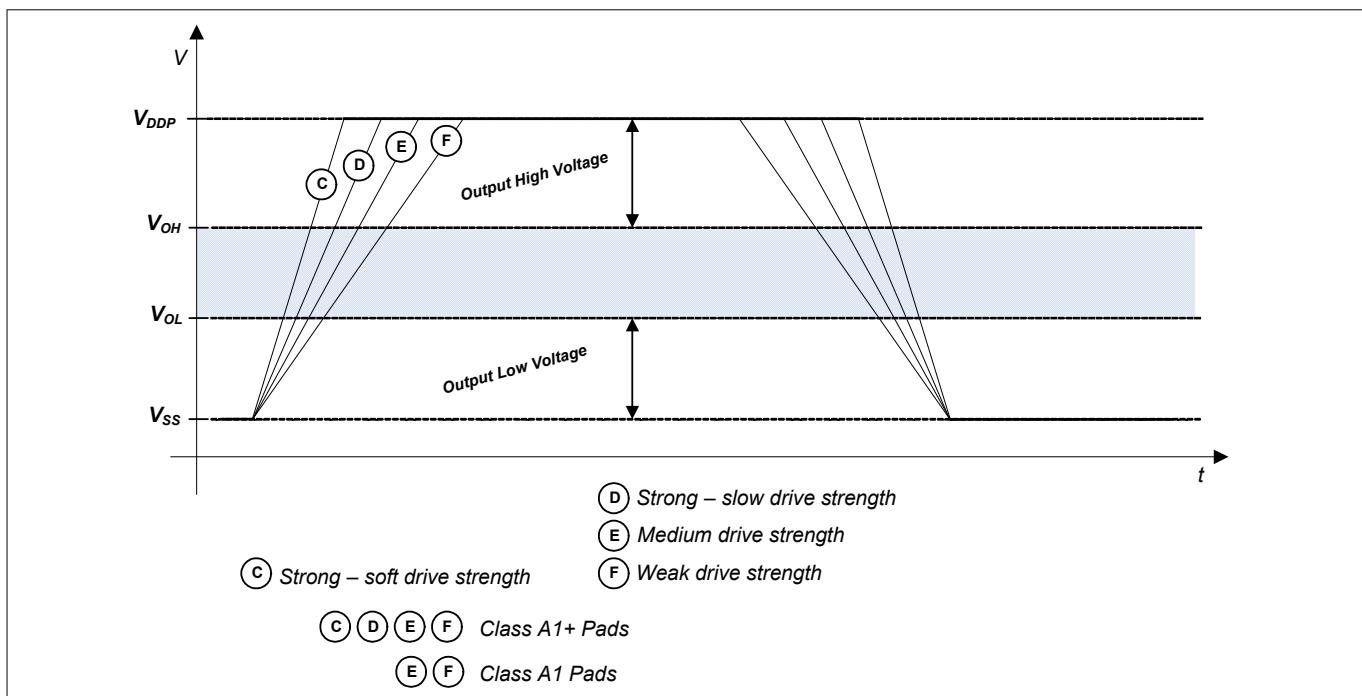
### 3 Electrical Parameters

#### 3.1.4 Pad Driver and Pad Classes Summary

This section gives an overview on the different pad driver classes and its basic characteristics. More details (mainly DC parameters) are defined in the [Input/Output Pins](#).

**Table 19 Pad Driver and Pad Classes Overview**

Class	Power Supply	Type	Sub-Class	Speed Grade	Load	Termination
A	3.3 V	LVTTI I/O, LVTTI outputs	A1 (e.g. GPIO)	6 MHz	100 pF	No
			A1+ (e.g. serial I/Os)	25 MHz	50 pF	Series termination recommended



**Figure 10 Output Slopes with different Pad Driver Modes**

Figure 10 is a qualitative display of the resulting output slope performance with different output driver modes. The detailed input and output characteristics are listed in [Input/Output Pins](#).

### 3 Electrical Parameters

#### 3.1.5 Operating Conditions

The following operating conditions must not be exceeded in order to ensure correct operation and reliability of the XMC4[12]00. All parameters specified in the following tables refer to these operating conditions, unless noted otherwise.

**Table 20** Operating Conditions Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Ambient Temperature	$T_A$ SR	-40	–	85	°C	Temp. Range F
		-40	–	125	°C	Temp. Range K
Digital supply voltage	$V_{DDP}$ SR	3.13 <sup>1)</sup>	3.3	3.63 <sup>2)</sup>	V	
Core Supply Voltage	$V_{DDC}$ CC	– <sup>1)</sup>	1.3	–	V	Generated internally
Digital ground voltage	$V_{SS}$ SR	0	–	–	V	
ADC analog supply voltage	$V_{DDA}$ SR	3.0	3.3	3.6 <sup>2)</sup>	V	
Analog ground voltage for $V_{DDA}$	$V_{SSA}$ SR	-0.1	0	0.1	V	
Battery Supply Voltage for Hibernate Domain <sup>3)</sup>	$V_{BAT}$ SR	1.95 <sup>4)</sup>	–	3.63	V	When $V_{DDP}$ is supplied $V_{BAT}$ has to be supplied as well.
System Frequency	$f_{SYS}$ SR	–	–	80	MHz	
Short circuit current of digital outputs	$I_{SC}$ SR	-5	–	5	mA	
Absolute sum of short circuit currents per pin group <sup>5)</sup>	$\Sigma I_{SC\_PG}$ SR	–	–	20	mA	
Absolute sum of short circuit currents of the device	$\Sigma I_{SC\_D}$ SR	–	–	100	mA	

1) See also the Supply Monitoring thresholds, [Section 3.3.2](#).

2) Voltage overshoot to 4.0 V is permissible at Power-Up and PORST low, provided the pulse duration is less than 100 µs and the cumulated sum of the pulses does not exceed 1 h over lifetime.

3) Different limits apply for LPAC operation, [Section 3.2.6](#).

4) To start the hibernate domain it is required that  $V_{BAT} \geq 2.1$  V, for a reliable start of the oscillation of RTC\_XTAL in crystal mode it is required that  $V_{BAT} \geq 3.0$  V.

5) The port groups are defined in [Table 18](#).

### 3 Electrical Parameters

## 3.2 DC Parameters

### 3.2.1 Input/Output Pins

The digital input stage of the shared analog/digital input pins is identical to the input stage of the standard digital input/output pins.

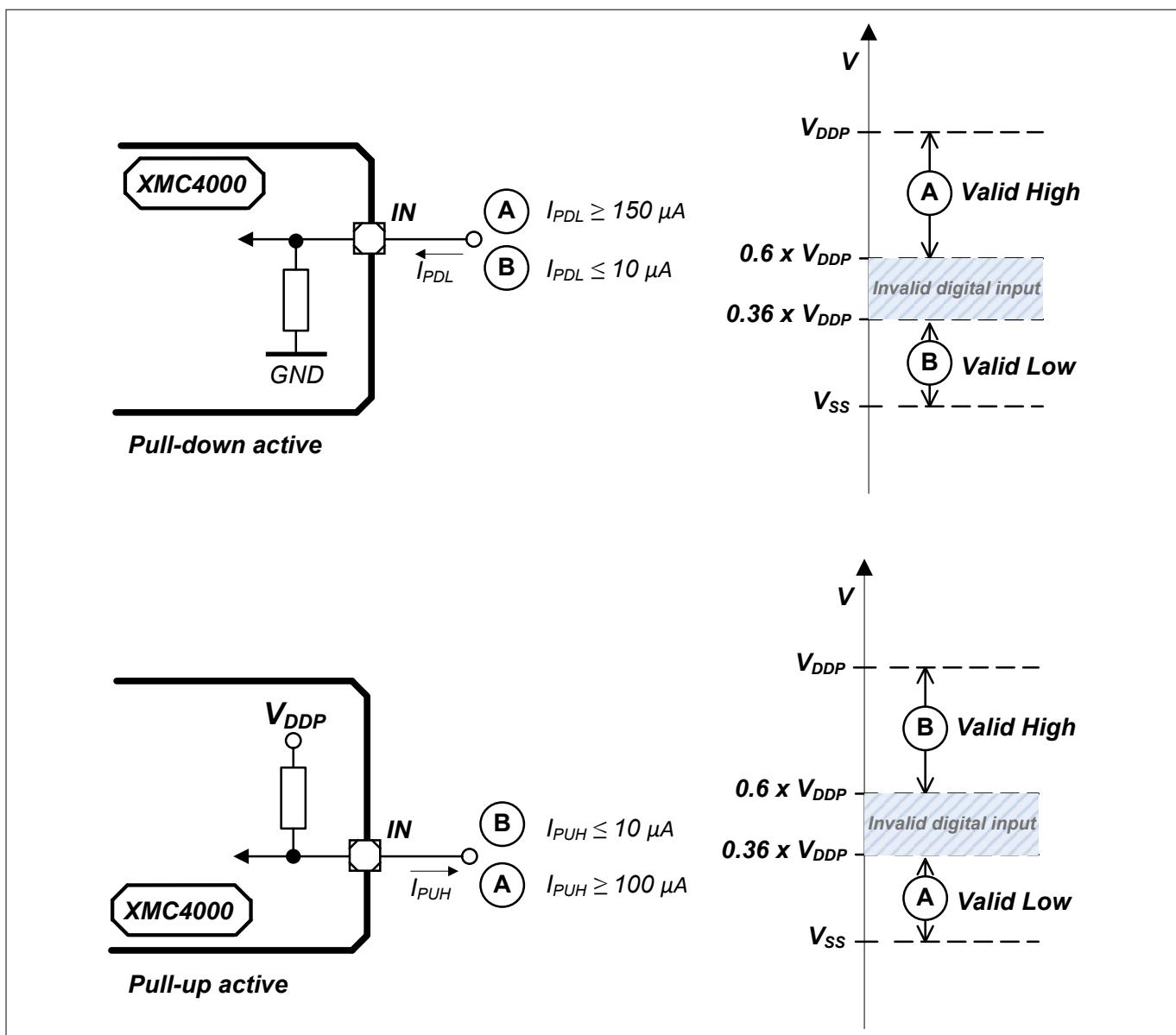
The pull-up characteristics ( $I_{PUH}$ ) and the input high and low voltage levels ( $V_{IH}$  and  $V_{IL}$ ) of the  $\overline{PORST}$  pin are identical to the respective values of the standard digital input/output pins.

**Table 21 Standard Pad Parameters**

Parameter	Symbol	Values		Unit	Note/Test Condition
		Min.	Max.		
Pin capacitance (digital inputs/outputs)	$C_{IO}$ CC	–	10	pF	
Pull-down current	$ I_{PDL} $ SR	150	–	$\mu A$	<sup>1)</sup> $V_{IN} \geq 0.6 \times V_{DDP}$
		–	10	$\mu A$	<sup>2)</sup> $V_{IN} \leq 0.36 \times V_{DDP}$
Pull-up current	$ I_{PUH} $ SR	–	10	$\mu A$	<sup>2)</sup> $V_{IN} \geq 0.6 \times V_{DDP}$
		100	–	$\mu A$	<sup>1)</sup> $V_{IN} \leq 0.36 \times V_{DDP}$
Input Hysteresis for pads of all A classes <sup>3)</sup>	HYSA CC	$0.1 \times V_{DDP}$	–	V	
PORST spike filter always blocked pulse duration	$t_{SF1}$ CC	–	10	ns	
PORST spike filter pass-through pulse duration	$t_{SF2}$ CC	100	–	ns	
PORST pull-down current	$ I_{PPD} $ CC	13	–	mA	$V_i = 1.0$ V

- 1) Current required to override the pull device with the opposite logic level (“force current”). With active pull device, at load currents between force and keep current the input state is undefined.
- 2) Load current at which the pull device still maintains the valid logic level (“keep current”). With active pull device, at load currents between force and keep current the input state is undefined.
- 3) Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It can not be guaranteed that it suppresses switching due to external system noise.

### 3 Electrical Parameters



**Figure 11** Pull Device Input Characteristics

Figure 11 visualizes the input characteristics with an active internal pull device:

- in the cases "A" the internal pull device is overridden by a strong external driver;
- in the cases "B" the internal pull device defines the input logical state against a weak external load.

**Table 22** Standard Pads Class\_A1

Parameter	Symbol	Values		Unit	Note/Test Condition
		Min.	Max.		
Input leakage current	$I_{OZA1}$ CC	-500	500	nA	$0 \text{ V} \leq V_{IN} \leq V_{DDP}$
Input high voltage	$V_{IHA1}$ SR	$0.6 \times V_{DDP}$	$V_{DDP} + 0.3$	V	max. 3.6 V
Input low voltage	$V_{ILA1}$ SR	-0.3	$0.36 \times V_{DDP}$	V	
Output high voltage, POD <sup>1)</sup> = weak	$V_{OHA1}$ CC	$V_{DDP} - 0.4$	-	V	$I_{OH} \geq -400 \mu\text{A}$
		2.4	-	V	$I_{OH} \geq -500 \mu\text{A}$

(table continues...)

### 3 Electrical Parameters

**Table 22 (continued) Standard Pads Class\_A1**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>		<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Max.</b>		
Output high voltage, POD <sup>1)</sup> = medium		$V_{DDP} - 0.4$	-	V	$I_{OH} \geq -1.4 \text{ mA}$
		2.4	-	V	$I_{OH} \geq -2 \text{ mA}$
Output low voltage	$V_{OLA1} \text{ CC}$	-	0.4	V	$I_{OL} \leq 500 \mu\text{A}$ ; POD <sup>1)</sup> = weak
		-	0.4	V	$I_{OL} \leq 2 \text{ mA}$ ; POD <sup>1)</sup> = medium
Fall time	$t_{FA1} \text{ CC}$	-	150	ns	$C_L = 20 \text{ pF}$ ; POD <sup>1)</sup> = weak
		-	50	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = medium
Rise time	$t_{RA1} \text{ CC}$	-	150	ns	$C_L = 20 \text{ pF}$ ; POD <sup>1)</sup> = weak
		-	50	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = medium

1) POD = Pin Out Driver

**Table 23 Standard Pads Class\_A1+**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>		<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Max.</b>		
Input leakage current	$I_{OZA1+} \text{ CC}$	-1	1	$\mu\text{A}$	$0 \text{ V} \leq V_{IN} \leq V_{DDP}$
Input high voltage	$V_{IHA1+} \text{ SR}$	$0.6 \times V_{DDP}$	$V_{DDP} + 0.3$	V	max. 3.6 V
Input low voltage	$V_{ILA1+} \text{ SR}$	-0.3	$0.36 \times V_{DDP}$	V	
Output high voltage, POD <sup>1)</sup> = weak	$V_{OHA1+} \text{ CC}$	$V_{DDP} - 0.4$	-	V	$I_{OH} \geq -400 \mu\text{A}$
		2.4	-	V	$I_{OH} \geq -500 \mu\text{A}$
Output high voltage, POD <sup>1)</sup> = medium		$V_{DDP} - 0.4$	-	V	$I_{OH} \geq -1.4 \text{ mA}$
		2.4	-	V	$I_{OH} \geq -2 \text{ mA}$
Output high voltage, POD <sup>1)</sup> = strong		$V_{DDP} - 0.4$	-	V	$I_{OH} \geq -1.4 \text{ mA}$
		2.4	-	V	$I_{OH} \geq -2 \text{ mA}$
Output low voltage	$V_{OLA1+} \text{ CC}$	-	0.4	V	$I_{OL} \leq 500 \mu\text{A}$ ; POD <sup>1)</sup> = weak
		-	0.4	V	$I_{OL} \leq 2 \text{ mA}$ ; POD <sup>1)</sup> = medium
		-	0.4	V	$I_{OL} \leq 2 \text{ mA}$ ; POD <sup>1)</sup> = strong

(table continues...)

### 3 Electrical Parameters

**Table 23** (continued) Standard Pads Class\_A1+

Parameter	Symbol	Values		Unit	Note/Test Condition
		Min.	Max.		
Fall time	$t_{FA1+}$ CC	-	150	ns	$C_L = 20 \text{ pF}$ ; POD <sup>1)</sup> = weak
		-	50	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = medium
		-	28	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = strong; edge = slow
		-	16	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = strong; edge = soft;
Rise time	$t_{RA1+}$ CC	-	150	ns	$C_L = 20 \text{ pF}$ ; POD <sup>1)</sup> = weak
		-	50	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = medium
		-	28	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = strong; edge = slow
		-	16	ns	$C_L = 50 \text{ pF}$ ; POD <sup>1)</sup> = strong; edge = soft

1) POD = Pin Out Driver

**Table 24** HIB\_IO Class\_A1 special Pads

Parameter	Symbol	Values		Unit	Note/Test Condition
		Min.	Max.		
Input leakage current	$I_{OZHIB}$ CC	-500	500	nA	$0 \text{ V} \leq V_{IN} \leq V_{BAT}$
Input high voltage	$V_{IHHIB}$ SR	$0.6 \times V_{BAT}$	$V_{BAT} + 0.3$	V	max. 3.6 V
Input low voltage	$V_{ILHIB}$ SR	-0.3	$0.36 \times V_{BAT}$	V	
Input Hysteresis for HIB_IO pins <sup>1)</sup>	$HYSHIB$ CC	$0.1 \times V_{BAT}$	-	V	$V_{BAT} \geq 3.13 \text{ V}$
		$0.06 \times V_{BAT}$	-	V	$V_{BAT} < 3.13 \text{ V}$
Output high voltage, POD <sup>1)</sup> = medium	$V_{OHHIB}$ CC	$V_{BAT} - 0.4$	-	V	$I_{OH} \geq -1.4 \text{ mA}$
Output low voltage	$V_{OLHIB}$ CC	-	0.4	V	$I_{OL} \leq 2 \text{ mA}$
Fall time	$t_{FHIB}$ CC	-	50	ns	$V_{BAT} \geq 3.13 \text{ V}$ $C_L = 50 \text{ pF}$

(table continues...)

### 3 Electrical Parameters

**Table 24 (continued) HIB\_IO Class\_A1 special Pads**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>		<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Max.</b>		
		-	100	ns	$V_{BAT} < 3.13 \text{ V}$ $C_L = 50 \text{ pF}$
Rise time	$t_{RHIB\_CC}$	-	50	ns	$V_{BAT} \geq 3.13 \text{ V}$ $C_L = 50 \text{ pF}$
		-	100	ns	$V_{BAT} < 3.13 \text{ V}$ $C_L = 50 \text{ pF}$

1) Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It can not be guaranteed that it suppresses switching due to external system noise.

### 3.2.2 Analog to Digital Converters (ADCx)

**Table 25 ADC Parameters (Operating Conditions apply)**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Analog reference voltage	$V_{AREF\_SR}$	-	-	-	V	$V_{AREF} = V_{DDA}$ shared analog supply and reference input pin
Alternate reference voltage <sup>1)</sup>	$V_{AREF\_SR}$	$V_{AGND} + 1$	-	$V_{DDA} + 0.05$ <sup>2)</sup>	V	
Analog reference ground	$V_{AGND\_SR}$	-	-	-	V	$V_{AGND} = V_{SSA}$ shared analog supply and reference input pin
Alternate reference voltage range <sup>1) 3)</sup>	$V_{AREF} - V_{AGND}$ SR	1	-	$V_{DDA} + 0.1$	V	
Analog input voltage	$V_{AIN\_SR}$	$V_{AGND}$	-	$V_{DDA}$	V	
Input leakage at analog inputs <sup>4)</sup>	$I_{OZ1\_CC}$	-100	-	200	nA	$0.03 \times V_{DDA} < V_{AIN} < 0.97 \times V_{DDA}$
		-500	-	100	nA	$0 \text{ V} \leq V_{AIN} \leq 0.03 \times V_{DDA}$
		-100	-	500	nA	$0.97 \times V_{DDA} \leq V_{AIN} \leq V_{DDA}$
Internal ADC clock	$f_{ADCI\_CC}$	2	-	30	MHz	$V_{DDA} = 3.3 \text{ V}$
Switched capacitance at the analog voltage inputs <sup>5)</sup>	$C_{AINSW\_CC}$	-	4	6.5	pF	

(table continues...)

### 3 Electrical Parameters

**Table 25 (continued) ADC Parameters (Operating Conditions apply)**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Total capacitance of an analog input	$C_{AINTOT}$ CC	-	12	20	pF	
Switched capacitance at the alternate reference voltage input <sup>1) 6)</sup>	$C_{AREFSW}$ CC	-	15	30	pF	
Total capacitance of the alternate reference inputs <sup>1)</sup>	$C_{AREFTOT}$ CC	-	20	40	pF	
Total Unadjusted Error	$TUE$ CC	-6	-	6	LSB	12-bit resolution; $V_{DDA} = 3.3$ V; $V_{AREF} = V_{DDA}$ <sup>7)</sup>
Differential Non-Linearity Error <sup>8)</sup>	$EA_{DNL}$ CC	-4.5	-	4.5	LSB	
Gain Error <sup>8)</sup>	$EA_{GAIN}$ CC	-6	-	6	LSB	
Integral Non-Linearity <sup>8)</sup>	$EA_{INL}$ CC	-4.5	-	4.5	LSB	
Offset Error <sup>8)</sup>	$EA_{OFF}$ CC	-6	-	6	LSB	
RMS Noise <sup>9)</sup>	$EN_{RMS}$ CC	-	1	2 <sup>10) 11)</sup>	LSB	
Worst case ADC $V_{DDA}$ power supply current per active converter	$I_{DDAA}$ CC	-	1.5	2	mA	during conversion $V_{DDP} = 3.6$ V, $T_J = 150^\circ\text{C}$
Charge consumption on alternate reference per conversion <sup>1)</sup>	$Q_{CONV}$ CC	-	30	-	pC	$0 \text{ V} \leq V_{AREF} \leq V_{DDA}$ <sup>12)</sup>
ON resistance of the analog input path	$R_{AIN}$ CC	-	600	1200	Ohm	
ON resistance for the ADC test (pull down for AIN7)	$R_{AIN7T}$ CC	180	550	900	Ohm	

1) Applies to AINx, when used as alternate reference input.

2) A running conversion may become imprecise in case the normal conditions are violated (voltage overshoot).

3) If the analog reference voltage is below  $V_{DDA}$ , then the ADC converter errors increase. If the reference voltage is reduced by the factor k ( $k < 1$ ), TUE, DNL, INL, Gain, and Offset errors increase also by the factor  $1/k$ .

4) The leakage current definition is a continuous function, as shown in figure ADCx Analog Inputs Leakage. The numerical values defined determine the characteristic points of the given continuous linear approximation - they do not define step function (see Figure 14).

5) The sampling capacity of the conversion C-network is pre-charged to  $V_{AREF}/2$  before the sampling moment. Because of the parasitic elements, the voltage measured at AINx can deviate from  $V_{AREF}/2$ .

6) This represents an equivalent switched capacitance. This capacitance is not switched to the reference voltage at once. Instead, smaller capacitances are successively switched to the reference voltage.

7) For 10-bit conversions, the errors are reduced to 1/4; for 8-bit conversions, the errors are reduced to 1/16. Never less than  $\pm 1$  LSB.

8) The sum of DNL/INL/GAIN/OFF errors does not exceed the related total unadjusted error TUE.

9) This parameter is valid for soldered devices and requires careful analog board design.

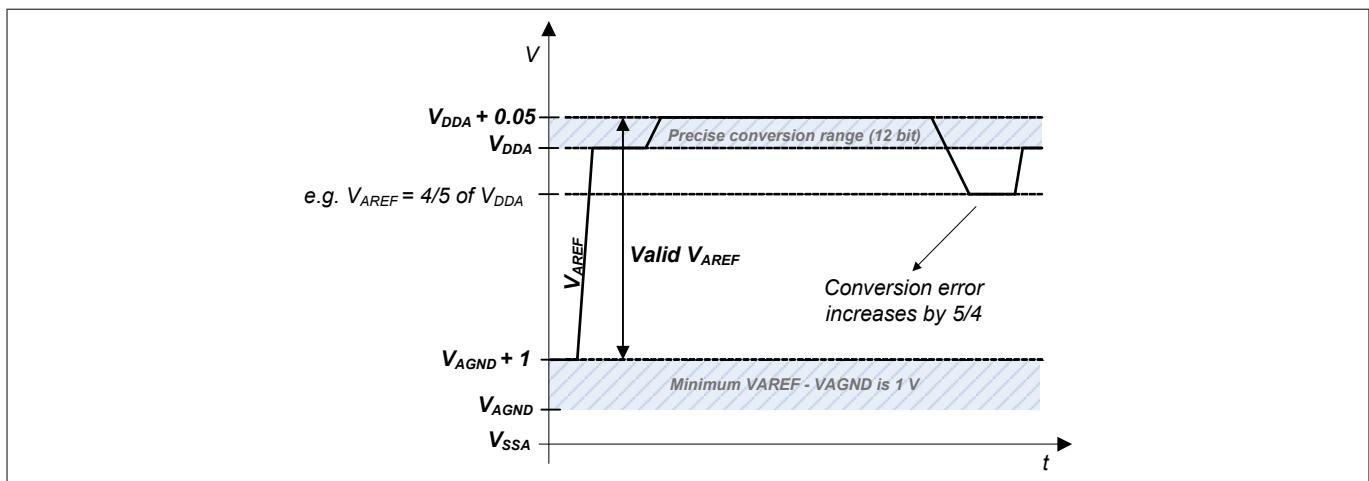
10) Resulting worst case combined error is arithmetic combination of TUE and  $EN_{RMS}$ .

11) Value is defined for one sigma Gauss distribution.

12) The resulting current for a conversion can be calculated with  $I_{AREF} = Q_{CONV}/t_c$ .

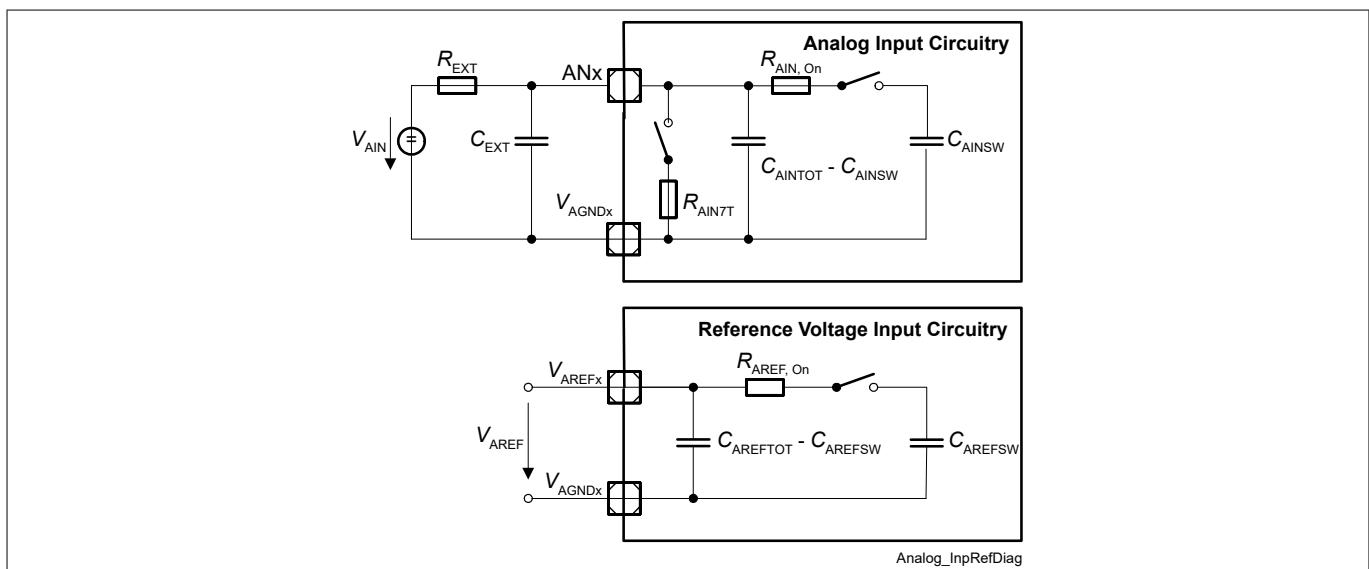
The fastest 12-bit post-calibrated conversion of  $t_c = 566$  ns results in a typical average current of  $I_{AREF} = 53 \mu\text{A}$ .

### 3 Electrical Parameters

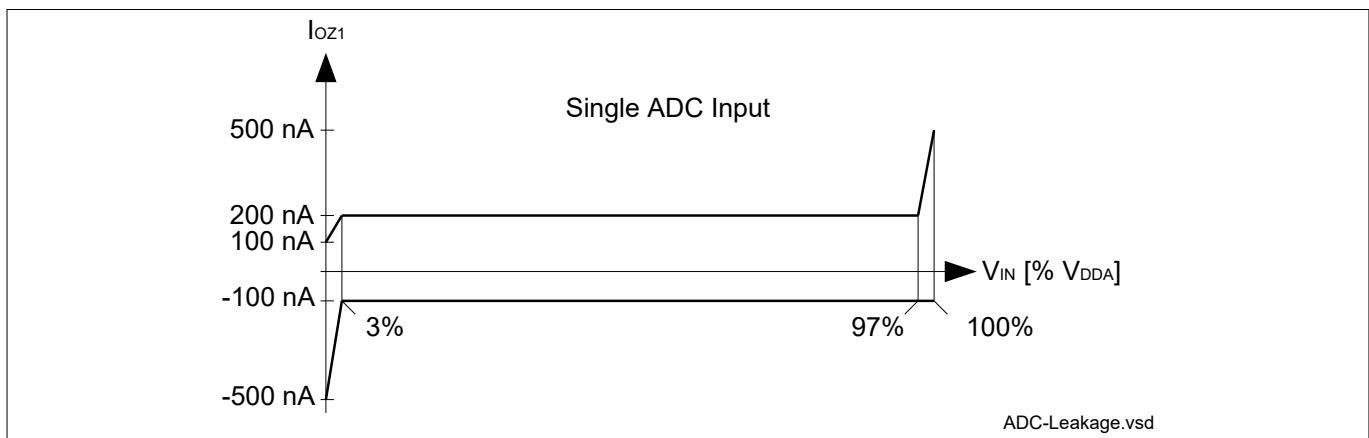


**Figure 12** VADC Reference Voltage Range

The power-up calibration of the ADC requires a maximum number of 4352  $f_{\text{ADCI}}$  cycles.



**Figure 13** ADCx Input Circuits



**Figure 14** ADCx Analog Input Leakage Current

### 3 Electrical Parameters

#### Conversion Time

**Table 26 Conversion Time (Operating Conditions apply)**

Parameter	Symbol	Values	Unit	Note
Conversion time	$t_{CC}$	$2 \times T_{ADC} + (2 + N + STC + PC + DM) \times T_{ADCI}$	μs	$N = 8, 10, 12$ for N-bit conversion $T_{ADC} = 1/f_{PERIPH}$ $T_{ADCI} = 1/f_{ADCI}$

- STC defines additional clock cycles to extend the sample time
- PC adds two cycles if post-calibration is enabled
- DM adds one cycle for an extended conversion time of the MSB

#### Conversion Time Examples

System assumptions (max.  $f_{ADC}$ ):

$f_{ADC} = 80$  MHz i.e.  $t_{ADC} = 12.5$  ns, DIVA = 2,  $f_{ADCI} = 26.7$  MHz i.e.  $t_{ADCI} = 37.5$  ns

According to the given formulas the following minimum conversion times can be achieved (STC = 0, DM = 0):

12-bit post-calibrated conversion (PC = 2):

$$t_{CN12C} = (2 + 12 + 2) \times t_{ADCI} + 2 \times t_{ADC} = 16 \times 37.5 \text{ ns} + 2 \times 12.5 \text{ ns} = 625 \text{ ns}$$

12-bit uncalibrated conversion:

$$t_{CN12} = (2 + 12) \times t_{ADCI} + 2 \times t_{ADC} = 14 \times 37.5 \text{ ns} + 2 \times 12.5 \text{ ns} = 550 \text{ ns}$$

10-bit uncalibrated conversion:

$$t_{CN10} = (2 + 10) \times t_{ADCI} + 2 \times t_{ADC} = 12 \times 37.5 \text{ ns} + 2 \times 12.5 \text{ ns} = 475 \text{ ns}$$

8-bit uncalibrated:

$$t_{CN8} = (2 + 8) \times t_{ADCI} + 2 \times t_{ADC} = 10 \times 37.5 \text{ ns} + 2 \times 12.5 \text{ ns} = 400 \text{ ns}$$

System assumptions (max.  $f_{ADCI}$ ):

$f_{ADC} = 60$  MHz i.e.  $t_{ADC} = 16.67$  ns, DIVA = 1,  $f_{ADCI} = 30$  MHz i.e.  $t_{ADCI} = 33.33$  ns

12-bit post-calibrated conversion (PC = 2):

$$t_{CN12C} = (2 + 12 + 2) \times t_{ADCI} + 2 \times t_{ADC} = 16 \times 33.33 \text{ ns} + 2 \times 16.67 \text{ ns} = 566 \text{ ns}$$

### 3 Electrical Parameters

#### 3.2.3 Digital to Analog Converters (DACx)

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 27 DAC Parameters (Operating Conditions apply)**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
RMS supply current	$I_{DD\ CC}$	–	2.5	4	mA	per active DAC channel, without load currents of DAC outputs
Resolution	$RES\ CC$	–	12	–	Bit	
Update rate	$f_{URATE\_A}\ CC$	–		2	Msample/s	data rate, where DAC can follow 64 LSB code jumps to $\pm 1$ LSB accuracy
Update rate	$f_{URATE\_F}\ CC$	–		5	Msample/s	data rate, where DAC can follow 64 LSB code jumps to $\pm 4$ LSB accuracy
Settling time	$t_{SETTLE}\ CC$	–	1	2	$\mu s$	at full scale jump, output voltage reaches target value $\pm 20$ LSB
Slew rate	$SR\ CC$	2	5	–	V/ $\mu s$	
Minimum output voltage	$V_{OUT\_MIN}\ CC$	–	0.3	–	V	code value unsigned: 000 <sub>H</sub> ; signed: 800 <sub>H</sub>
Maximum output voltage	$V_{OUT\_MAX}\ CC$	–	2.5	–	V	code value unsigned: FFF <sub>H</sub> ; signed: 7FF <sub>H</sub>
Integral non-linearity <sup>1)</sup>	$INL\ CC$	-5.5	$\pm 2.5$	5.5	LSB	$R_L \geq 5$ kOhm, $C_L \leq 50$ pF
Differential non-linearity	$DNL\ CC$	-2	$\pm 1$	2	LSB	$R_L \geq 5$ kOhm, $C_L \leq 50$ pF
Offset error	$ED_{OFF}\ CC$		$\pm 20$		mV	
Gain error	$ED_{G\_IN}\ CC$	-5	0	5	%	
Startup time	$t_{STARTUP}\ CC$	–	15	30	$\mu s$	time from output enabling till code valid $\pm 16$ LSB
3dB Bandwidth of Output Buffer	$f_{C1}\ CC$	2.5	5	–	MHz	verified by design

(table continues...)

### 3 Electrical Parameters

**Table 27 (continued) DAC Parameters (Operating Conditions apply)**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Output sourcing current	$I_{OUT\_SOURCE}$ CC	-	-30	-	mA	
Output sinking current	$I_{OUT\_SINK}$ CC	-	0.6	-	mA	
Output resistance	$R_{OUT}$ CC	-	50	-	Ohm	
Load resistance	$R_L$ SR	5	-	-	kOhm	
Load capacitance	$C_L$ SR	-	-	50	pF	
Signal-to-Noise Ratio	SNR CC	-	70	-	dB	examination bandwidth < 25 kHz
Total Harmonic Distortion	THD CC	-	70	-	dB	examination bandwidth < 25 kHz
Power Supply Rejection Ratio	PSRR CC	-	56	-	dB	to $V_{DDA}$ verified by design

1) According to best straight line method.

### Conversion Calculation

Unsigned:

$$DACxDATA = 4095 \times (V_{OUT} - V_{OUT\_MIN}) / (V_{OUT\_MAX} - V_{OUT\_MIN})$$

Signed:

$$DACxDATA = 4095 \times (V_{OUT} - V_{OUT\_MIN}) / (V_{OUT\_MAX} - V_{OUT\_MIN}) - 2048$$

### 3 Electrical Parameters

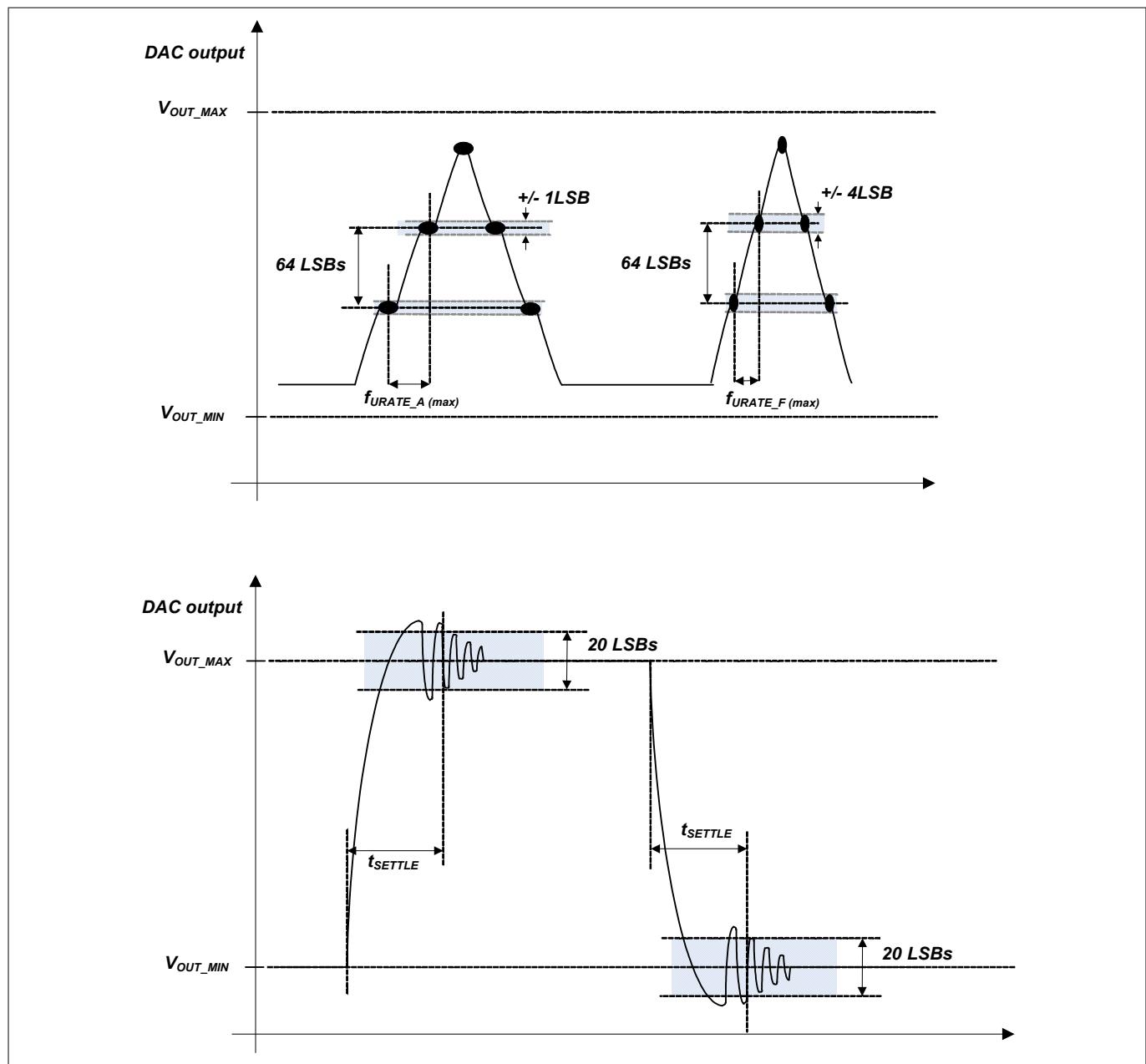


Figure 15 DAC Conversion Examples

### 3 Electrical Parameters

#### 3.2.4 Out-of-Range Comparator (ORC)

The Out-of-Range Comparator (ORC) triggers on analog input voltages ( $V_{AIN}$ ) above the analog reference<sup>1)</sup> ( $V_{AREF}$ ) on selected input pins (GxORC<sub>C</sub>) and generates a service request trigger (GxORCOUT<sub>y</sub>).

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

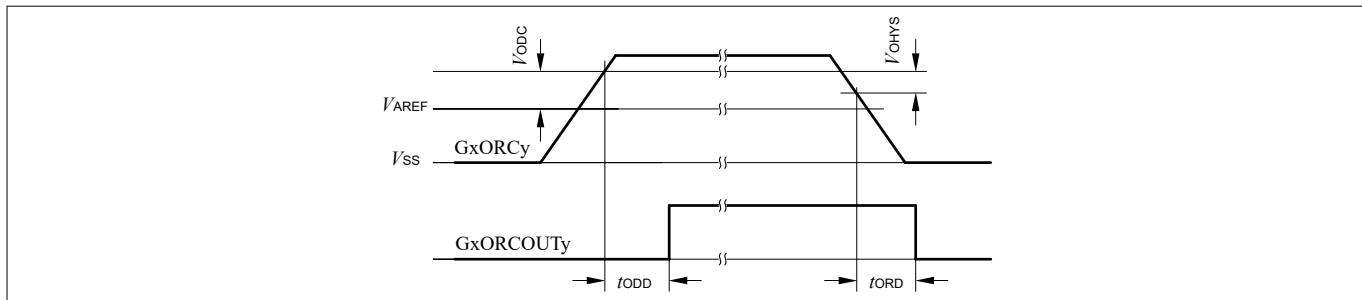
The parameters in [Table 28](#) apply for the maximum reference voltage  $V_{AREF} = V_{DDA} + 50 \text{ mV}$ .

**Table 28 ORC Parameters (Operating Conditions apply)**

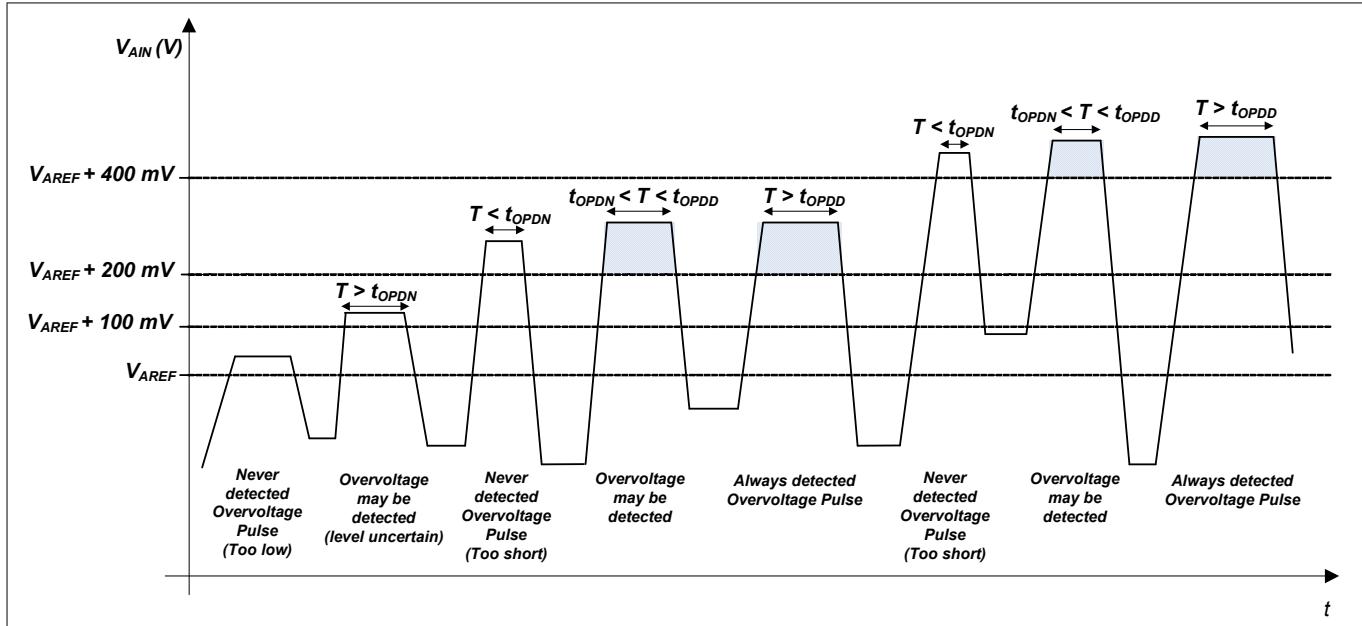
Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
DC Switching Level	$V_{ODC}$ CC	100	125	200	mV	Ax-marking devices $V_{AIN} \geq V_{AREF} + V_{ODC}$
Hysteresis	$V_{OHYS}$ CC	50	-	$V_{ODC}$	mV	
Detection Delay of a persistent Overvoltage	$t_{ODD}$ CC	55	-	450	ns	Ax-marking devices $V_{AIN} \geq V_{AREF} + 200 \text{ mV}$
		45	-	105	ns	$V_{AIN} \geq V_{AREF} + 400 \text{ mV}$
Always detected Overvoltage Pulse	$t_{OPDD}$ CC	440	-	-	ns	Ax-marking devices $V_{AIN} \geq V_{AREF} + 200 \text{ mV}$
		90	-	-	ns	$V_{AIN} \geq V_{AREF} + 400 \text{ mV}$
Never detected Overvoltage Pulse	$t_{OPDN}$ CC	-	-	49	ns	Ax-marking devices $V_{AIN} \geq V_{AREF} + 200 \text{ mV}$
		-	-	30	ns	$V_{AIN} \geq V_{AREF} + 400 \text{ mV}$
Release Delay	$t_{ORD}$ CC	65	-	105	ns	$V_{AIN} \leq V_{AREF}$
Enable Delay	$t_{OED}$ CC	-	100	200	ns	

<sup>1</sup> Always the standard VADC reference, alternate references do not apply to the ORC.

### 3 Electrical Parameters



**Figure 16** GxORCOUTy Trigger Generation



**Figure 17** ORC Detection Ranges

### 3 Electrical Parameters

#### 3.2.5 High Resolution PWM (HRPWM)

The following chapters describe the operating conditions, characteristics and timing requirements, for all the components inside the HRPWM module. Each description is given for just one sub unit, e.g., one CSG or one HRC.

All the timing information is related to the module clock,  $f_{hrpwm}$ .

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

##### 3.2.5.1 HRC characteristics

Table 29 summarizes the characteristics of the HRC units.

**Table 29 HRC characteristics (Operating Conditions apply)**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
High resolution step size <sup>1) 2)</sup>	$t_{HRS\ CC}$	–	150	–	ps	
Startup time (after reset release)	$t_{start\ CC}$	–	–	2	μs	

1) The step size for clock frequencies equal to 180, 120 and 80 MHz is 150 ps.

2) The step size for clock frequencies different from 180, 120 and 80 MHz but within the range from 180 to 64 MHz can be between 118 to 180 ps (fixed over process and operating conditions)

### 3 Electrical Parameters

#### 3.2.5.2 CMP and 10-bit DAC characteristics

The [Table 30](#) summarizes the characteristics of the CSG unit.

The specified characteristics require that the setup of the HRPWM follows the initialization sequence as documented in the Reference Manual.

**Table 30** **CMP and 10-bit DAC characteristics (Operating Conditions apply)**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
DAC Resolution	$RES_{CC}$		10		bits	
DAC differential nonlinearity	$DNL_{CC}$	-1	-	1.5	LSB	Monotonic behavior See <a href="#">Figure 18</a>
DAC integral nonlinearity	$INL_{CC}$	-3	-	3	LSB	See <a href="#">Figure 18</a>
CSG Output Jitter	$D_{CSG}_{CC}$	-	-	1	clk	
Bias startup time	$t_{start}_{CC}$	-	-	98	us	
Bias supply current	$I_{DDbias}_{CC}$	-	-	400	$\mu A$	
CSGy startup time	$t_{CSGs}_{CC}$	-	-	2	$\mu s$	
Input operation current <sup>1)</sup>	$I_{DDCIN}_{CC}$	-10	-	33	$\mu A$	See <a href="#">Figure 19</a>
<b>High Speed Mode</b>						
DAC output voltage range	$V_{DOUT}_{CC}$	$V_{SS}$	-	$V_{DDP}$	V	
DAC propagation delay - Full scale	$t_{FShs}_{CC}$	-	-	80	ns	See <a href="#">Figure 20</a>
Input Selector propagation delay - Full scale	$t_{Dhs}_{CC}$	-	-	100	ns	See <a href="#">Figure 20</a>
Comparator bandwidth	$t_{Dhs}_{CC}$	20	-	-	ns	
DAC CLK frequency	$f_{clk}_{SR}$	-	-	30	MHz	
Supply current	$I_{DDhs}_{CC}$	-	-	940	$\mu A$	
<b>Low Speed Mode</b>						
DAC output voltage range	$V_{DOUT}_{CC}$	$0.1 \times V_{DDP}^{2)}$	-	$V_{DDP}$	V	
DAC propagation delay - Full Scale	$t_{FSls}_{CC}$	-	-	160	ns	See <a href="#">Figure 20</a>
Input Selector propagation delay - Full Scale	$t_{Dls}_{CC}$	-	-	200	ns	See <a href="#">Figure 20</a>
Comparator bandwidth	$t_{Dls}_{CC}$	20	-	-	ns	

(table continues...)

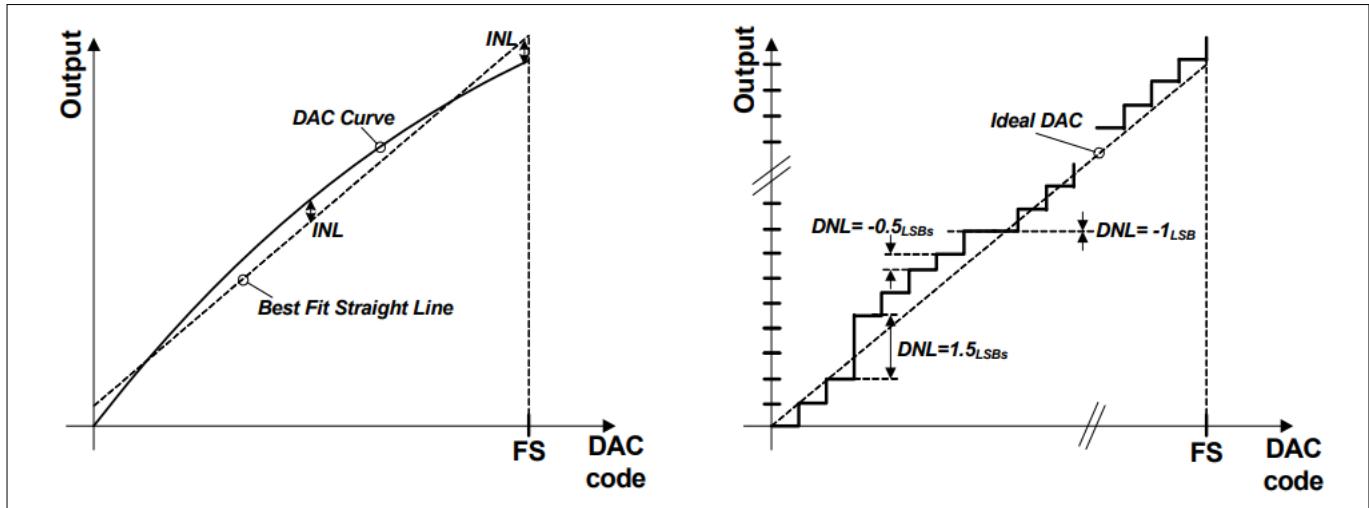
### 3 Electrical Parameters

**Table 30** (continued) CMP and 10-bit DAC characteristics (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
DAC CLK frequency	$f_{\text{clk SR}}$	-	-	30	MHz	
Supply current	$I_{\text{DDIs CC}}$	-	-	300	$\mu\text{A}$	

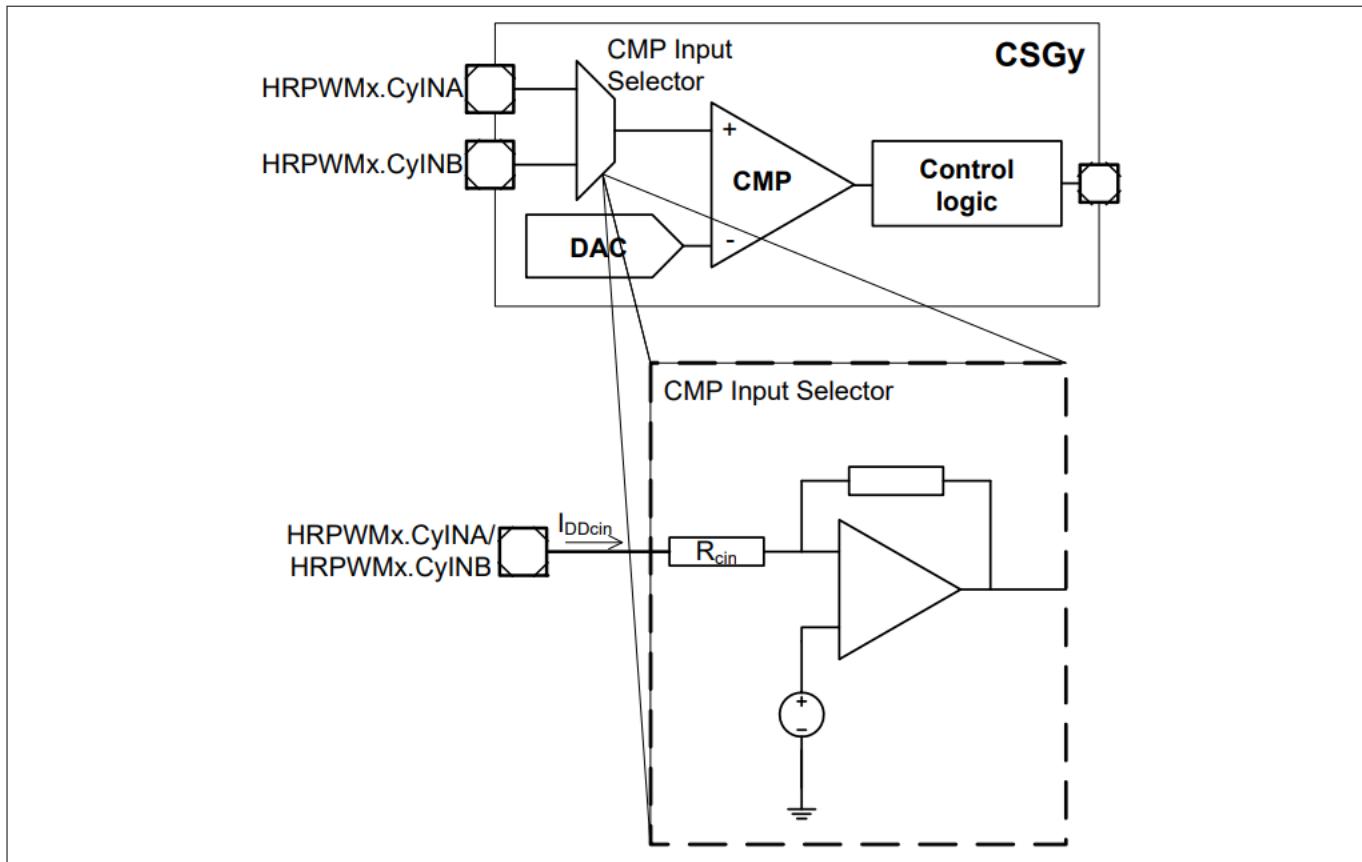
1) Typical input resistance  $R_{\text{CIN}} = 100 \text{ kOhm}$ .

2) The INL error increases for DAC output voltages below this limit.

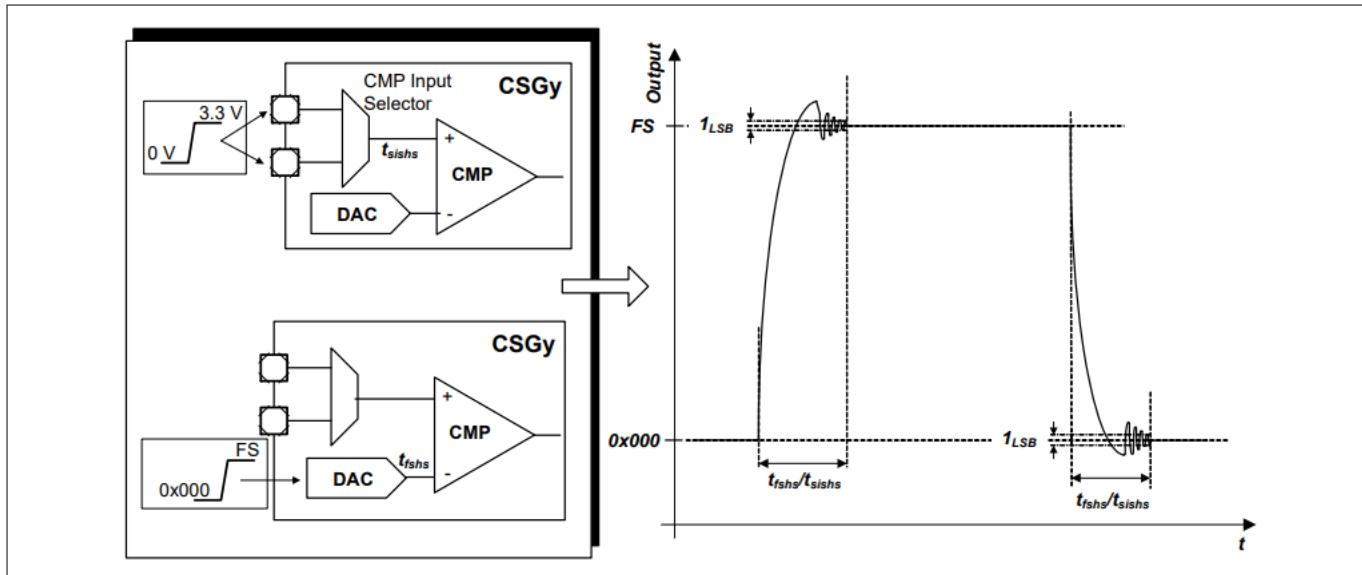


**Figure 18** CSG DAC INL and DNL example

### 3 Electrical Parameters



**Figure 19** Input operation current



**Figure 20** DAC and Input Selector Propagation Delay

### 3 Electrical Parameters

#### 3.2.5.3 Clocks

##### HRPWM DAC Conversion Clock

The DAC conversion clock can be generated internally or it can be controlled via a HRPWM module pin.

**Table 31 External DAC conversion trigger operating conditions**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Frequency	$f_{etrg}$ SR	–	–	30 <sup>2)</sup>	MHz	
ON time	$t_{onetrq}$ SR	$2T_{ccu}$ <sup>1) 2)</sup>	–	–	ns	
OFF time	$t_{offetrq}$ SR	$2T_{ccu}$ <sup>1) 2)</sup>	–	–	ns	

1) 50% duty cycle is not obligatory

2) Only valid if the signal was not previously synchronized/generated with the fccu clock (or a synchronous clock)

##### CSG External Clock

It is possible to select an external source, that can be used as a clock for the slope generation, HRPWMx.ECLKy. This clock is synchronized internally with the module clock and therefore the external clock needs to meet the criterion described on [Table 32](#).

**Table 32 External clock operating conditions**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Frequency	$f_{eclk}$ SR	–	–	$f_{hrpwm}/4$	MHz	
ON time	$t_{oneclk}$ SR	$2T_{ccu}$ <sup>1) 2)</sup>	–	–	ns	
OFF time	$t_{offeclk}$ SR	$2T_{ccu}$ <sup>1) 2)</sup>	–	–	ns	Only the rising edge is used

1) 50% duty cycle is not obligatory

2) Only valid if the signal was not previously synchronized/generated with the fccu clock (or a synchronous clock)

#### 3.2.6 Low Power Analog Comparator (LPAC)

The Low Power Analog Comparator (LPAC) triggers a wake-up event from Hibernate state or an interrupt trigger during normal operation. It does so by comparing  $V_{BAT}$  or another external sensor voltage  $V_{LPS}$  with a pre-programmed threshold voltage.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 33 Low Power Analog Comparator Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
$V_{BAT}$ supply voltage range for LPAC operation	$V_{BAT}$ SR	2.1	–	3.6	V	
Sensor voltage range	$V_{LPS}$ CC	0	–	1.2	V	
<b>(table continues...)</b>						

### 3 Electrical Parameters

**Table 33** (continued) Low Power Analog Comparator Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Threshold step size	$V_{th}$ CC	–	18.75	–	mV	
Threshold trigger accuracy	$\Delta V_{th}$ CC	–	–	$\pm 10$	%	for $V_{th} > 0.4$ V
Conversion time	$t_{LPCC}$ CC	–	–	250	$\mu s$	
Average current consumption over time	$I_{LPCAC}$ CC	–	–	15	$\mu A$	conversion interval 10 ms <sup>1)</sup>
Current consumption during conversion	$I_{LPCC}$ CC	–	150	–	$\mu A$	<sup>1)</sup>

1) Single channel conversion, measuring  $V_{BAT} = 3.3$  V, 8 cycles settling time

### 3.2.7 Die Temperature Sensor

The Die Temperature Sensor (DTS) measures the junction temperature  $T_J$ .

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 34** Die Temperature Sensor Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Temperature sensor range	$T_{SR}$ SR	-40	–	150	°C	
Linearity Error (to the below defined formula)	$\Delta T_{LE}$ CC	–	$\pm 1$	–	°C	per $\Delta T_J \leq 30$ °C
Offset Error	$\Delta T_{OE}$ CC	–	$\pm 6$	–	°C	$\Delta T_{OE} = T_J - T_{DTS}$ $V_{DDP} \leq 3.3$ V <sup>1)</sup>
Measurement time	$t_M$ CC	–	–	100	$\mu s$	
Start-up time after reset inactive	$t_{TSST}$ SR	–	–	10	$\mu s$	

1) At  $V_{DDP\_max} = 3.63$  V the typical offset error increases by an additional  $\Delta T_{OE} = \pm 1$  °C.

The following formula calculates the temperature measured by the DTS in [°C] from the RESULT bit field of the DTSSTAT register.

$$\text{Temperature } T_{DTS} = (\text{RESULT} - 605) / 2.05 \text{ [°C]}$$

This formula and the values defined in Table 34 apply with the following calibration values:

- DTSCON.BGTRIM = 8<sub>H</sub>
- DTSCON.REFTRIM = 4<sub>H</sub>

### 3 Electrical Parameters

#### 3.2.8 USB Device Interface DC Characteristics

The Universal Serial Bus (USB) Interface is compliant to the USB Rev. 2.0 Specification. High-Speed Mode is not supported.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 35 USB Device Data Line (USB\_DP, USB\_DM) Parameters (Operating Conditions apply)**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Input low voltage	$V_{IL\ SR}$	–	–	0.8	V	
Input high voltage (driven)	$V_{IH\ SR}$	2.0	–	–	V	
Input high voltage (floating) <sup>1)</sup>	$V_{IHZ\ SR}$	2.7	–	3.6	V	
Differential input sensitivity	$V_{DIS\ CC}$	0.2	–	–	V	
Differential common mode range	$V_{CM\ CC}$	0.8	–	2.5	V	
Output low voltage	$V_{OL\ CC}$	0.0	–	0.3	V	1.5 kOhm pull-up to 3.6 V
Output high voltage	$V_{OH\ CC}$	2.8	–	3.6	V	15 kOhm pull-down to 0 V
DP pull-up resistor (idle bus)	$R_{PUI\ CC}$	900	–	1575	Ohm	
DP pull-up resistor (upstream port receiving)	$R_{PUA\ CC}$	1 425	–	3090	Ohm	
Input impedance DP, DM	$Z_{INP\ CC}$	300	–	–	kOhm	$0\text{ V} \leq V_{IN} \leq V_{DDP}$
Driver output resistance DP, DM	$Z_{DRV\ CC}$	28	–	44	Ohm	

1) Measured at A-connector with 1.5 kOhm  $\pm$  5% to 3.3 V  $\pm$  0.3 V connected to USB\_DP or USB\_DM and at B-connector with 15 kOhm  $\pm$  5% to ground connected to USB\_DP and USB\_DM.

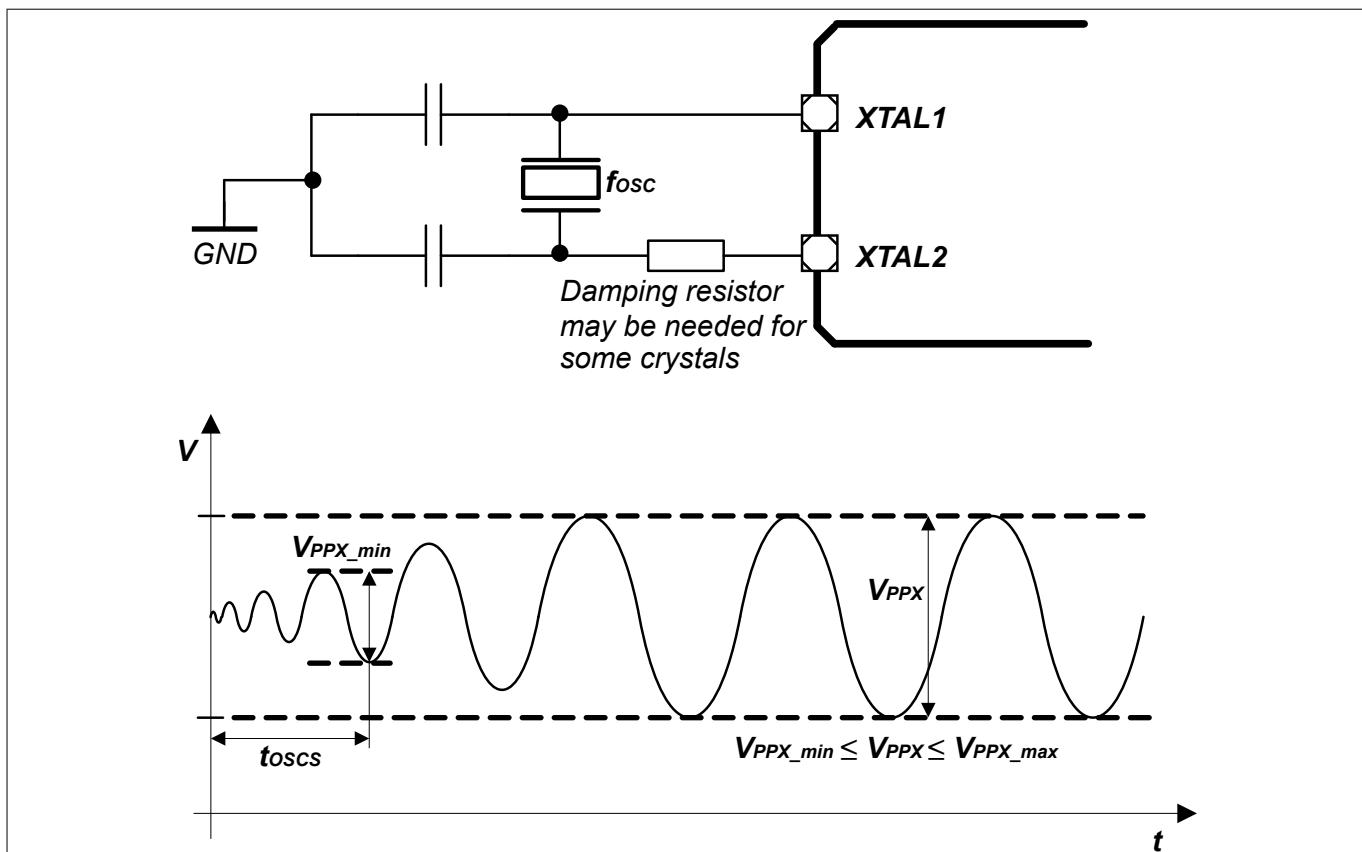
#### 3.2.9 Oscillator Pins

**Note:** It is strongly recommended to measure the oscillation allowance (negative resistance) in the final target system (layout) to determine the optimal parameters for the oscillator operation. Please refer to the limits specified by the crystal or ceramic resonator supplier.

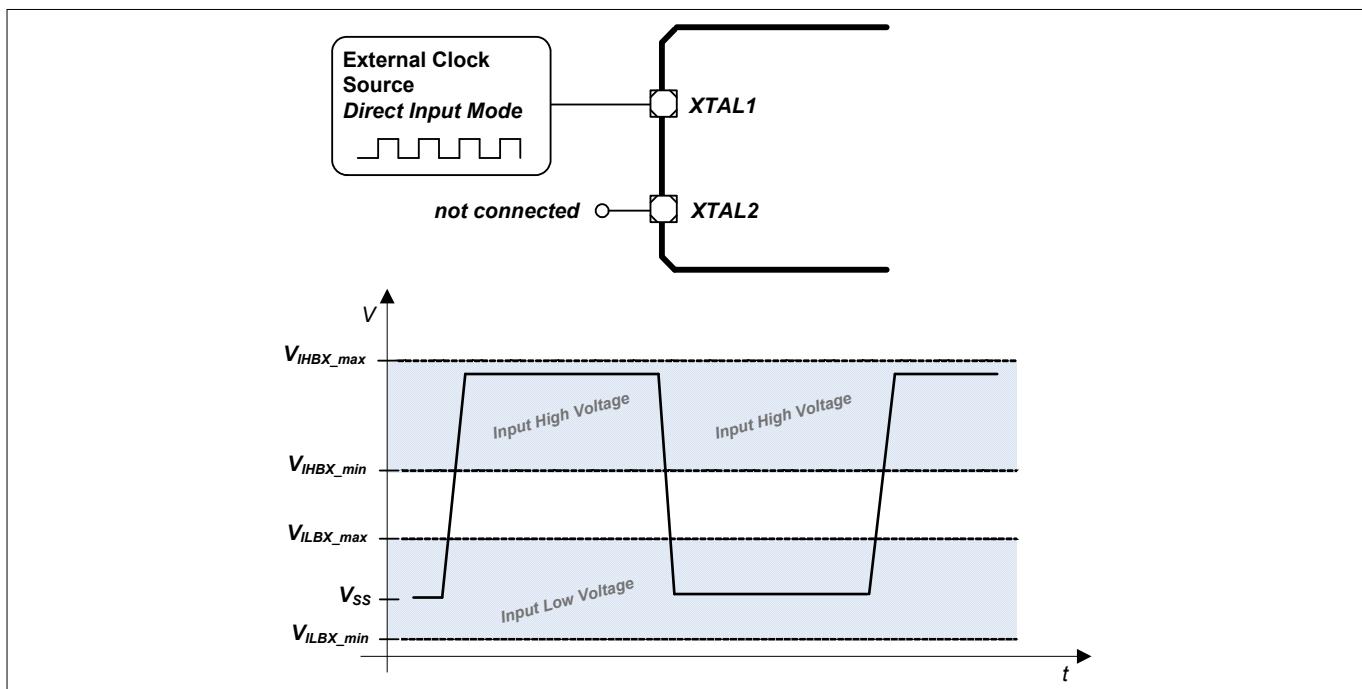
**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

The oscillator pins can be operated with an external crystal (see Figure 21) or in direct input mode (see Figure 22).

### 3 Electrical Parameters



**Figure 21** Oscillator in Crystal Mode



**Figure 22** Oscillator in Direct Input Mode

### 3 Electrical Parameters

**Table 36 OSC\_XTAL Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Input frequency	$f_{\text{OSC SR}}$	4	-	40	MHz	Direct Input Mode selected
		4	-	25	MHz	External Crystal Mode selected
Oscillator start-up time <sup>1) 2)</sup>	$t_{\text{oscS CC}}$	-	-	10	ms	
Input voltage at XTAL1	$V_{IX \text{ SR}}$	-0.5	-	$V_{DDP} + 0.5$	V	
Input amplitude (peak-to-peak) at XTAL1 <sup>2) 3)</sup>	$V_{PPX \text{ SR}}$	$0.4 \times V_{DDP}$	-	$V_{DDP} + 1.0$	V	
Input high voltage at XTAL1 <sup>4)</sup>	$V_{IHBX \text{ SR}}$	1.0	-	$V_{DDP} + 0.5$	V	
Input low voltage at XTAL1 <sup>4)</sup>	$V_{ILBX \text{ SR}}$	-0.5	-	0.4	V	
Input leakage current at XTAL1	$I_{ILX1 \text{ CC}}$	-100	-	100	nA	Oscillator power down $0 \text{ V} \leq V_{IX} \leq V_{DDP}$

- 1)  $t_{\text{oscS}}$  is defined from the moment the oscillator is enabled with SCU\_OSCHPCTRL.MODE until the oscillations reach an amplitude at XTAL of  $0.4 * V_{DDP}$ .
- 2) The external oscillator circuitry must be optimized by the customer and checked for negative resistance and amplitude as recommended and specified by crystal suppliers.
- 3) If the shaper unit is enabled and not bypassed.
- 4) If the shaper unit is bypassed, dedicated DC-thresholds have to be met.

**Table 37 RTC\_XTAL Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Input frequency	$f_{\text{OSC SR}}$	-	32.768	-	kHz	
Oscillator start-up time <sup>1) 2) 3)</sup>	$t_{\text{oscS CC}}$	-	-	5	s	
Input voltage at RTC_XTAL1	$V_{IX \text{ SR}}$	-0.3	-	$V_{BAT} + 0.3$	V	
Input amplitude (peak-to-peak) at RTC_XTAL1 <sup>2) 4)</sup>	$V_{PPX \text{ SR}}$	0.4	-	-	V	
Input high voltage at RTC_XTAL1 <sup>5)</sup>	$V_{IHBX \text{ SR}}$	$0.6 \times V_{BAT}$	-	$V_{BAT} + 0.3$	V	
Input low voltage at RTC_XTAL1 <sup>5)</sup>	$V_{ILBX \text{ SR}}$	-0.3	-	$0.36 \times V_{BAT}$	V	

(table continues...)

### 3 Electrical Parameters

**Table 37 (continued) RTC\_XTAL Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Input Hysteresis for RTC_XTAL1 <sup>3) 6)</sup>	$V_{HYSX}$ CC	$0.1 \times V_{BAT}$		–	V	$3.0 \text{ V} \leq V_{BAT} < 3.6 \text{ V}$
		$0.03 \times V_{BAT}$		–	V	$V_{BAT} < 3.0 \text{ V}$
Input leakage current at RTC_XTAL1	$I_{ILX1}$ CC	-100	–	100	nA	Oscillator power down $0 \text{ V} \leq V_{IX} \leq V_{BAT}$

- 1)  $t_{osc}$  is defined from the moment the oscillator is enabled by the user with SCU\_OSCULCTRL.MODE until the oscillations reach an amplitude at RTC\_XTAL1 of 400 mV.
- 2) The external oscillator circuitry must be optimized by the customer and checked for negative resistance and amplitude as recommended and specified by crystal suppliers.
- 3) For a reliable start of the oscillation in crystal mode it is required that  $V_{BAT} \geq 3.0 \text{ V}$ . A running oscillation is maintained across the full  $V_{BAT}$  voltage range.
- 4) If the shaper unit is enabled and not bypassed.
- 5) If the shaper unit is bypassed, dedicated DC-thresholds have to be met.
- 6) Hysteresis is implemented to avoid metastable states and switching due to internal ground bounce. It can not be guaranteed that it suppresses switching due to external system noise.

### 3.2.10 Power Supply Current

The total power supply current defined below consists of a leakage and a switching component.

Application relevant values are typically lower than those given in the following tables, and depend on the customer's system operating conditions (e.g. thermal connection or used application configurations).

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

If not stated otherwise, the operating conditions for the parameters in the following table are:

$V_{DDP} = 3.3 \text{ V}$ ,  $T_A = 25^\circ\text{C}$

**Table 38 Power Supply Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Active supply current <sup>1)</sup> Peripherals enabled Frequency: $f_{CPU}/f_{PERIPH}/f_{CCU}$ in MHz	$I_{DDPA}$ CC	–	80	–	mA	80/80/80
		–	75	–		80/40/40
		–	73	–		40/40/80
		–	59	–		24/24/24
		–	50	–		1/1/1
Active supply current Code execution from RAM Flash in Sleep mode Frequency: $f_{CPU}/f_{PERIPH}/f_{CCU}$ in MHz	$I_{DDPA}$ CC	–	24	–	mA	80/80/80
		–	19	–		80/40/40
<b>(table continues...)</b>						

### 3 Electrical Parameters

**Table 38 (continued) Power Supply Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Active supply current <sup>2)</sup> Peripherals disabled Frequency: $f_{CPU}/f_{PERIPH}$ in MHz	$I_{DDPA\ CC}$	–	63	–	mA	80/80/80
		–	62	–		80/40/40
		–	60	–		40/40/80
		–	54	–		24/24/24
		–	50	–		1/1/1
Sleep supply current <sup>3)</sup> Peripherals enabled Frequency: $f_{CPU}/f_{PERIPH}/f_{CCU}$ in MHz	$I_{DDPS\ CC}$	–	76	–	mA	80/80/80
		–	73	–		80/40/40
		–	70	–		40/40/80
		–	56	–		24/24/24
		–	47	–		1/1/1
$f_{CPU}/f_{PERIPH}/f_{CCU}$ in kHz		–	46	–		100/100/100
Sleep supply current <sup>4)</sup> Peripherals disabled Frequency: $f_{CPU}/f_{PERIPH}/f_{CCU}$ in MHz	$I_{DDPS\ CC}$	–	59	–	mA	80/80/80
		–	58	–		80/40/40
		–	57	–		40/40/80
		–	51	–		24/24/24
		–	46	–		1/1/1
$f_{CPU}/f_{PERIPH}/f_{CCU}$ in kHz		–	46	–		100/100/100
Deep Sleep supply current <sup>5)</sup> Flash in Sleep mode Frequency: $f_{CPU}/f_{PERIPH}/f_{CCU}$ in MHz	$I_{DDPD\ CC}$	–	6.9	–	mA	24/24/24
		–	4.3	–		4/4/4
		–	3.8	–		1/1/1
		–	4.5	–		100/100/100 <sup>6)</sup>
Hibernate supply current RTC on <sup>7)</sup>	$I_{DDPH\ CC}$	–	10.8	–	$\mu A$	$V_{BAT} = 3.3\text{ V}$
		–	8.0	–		$V_{BAT} = 2.4\text{ V}$
		–	6.8	–		$V_{BAT} = 2.0\text{ V}$
Hibernate supply current RTC off <sup>8)</sup>	$I_{DDPH\ CC}$	–	10.3	–	$\mu A$	$V_{BAT} = 3.3\text{ V}$
		–	7.5	–		$V_{BAT} = 2.4\text{ V}$
		–	6.3	–		$V_{BAT} = 2.0\text{ V}$
Worst case active supply current <sup>9)</sup>	$I_{DDPA\ CC}$	–	–	140 <sup>10)</sup>	mA	$V_{DDP} = 3.6\text{ V}, T_J = 150^\circ\text{C}$
$V_{DDA}$ power supply current	$I_{DDA\ CC}$	–	–	– <sup>11)</sup>	mA	

(table continues...)

### 3 Electrical Parameters

**Table 38 (continued) Power Supply Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
$I_{DDP}$ current at PORST Low	$I_{DDP\_PORST\ CC}$	–	–	24	mA	$V_{DDP} = 3.6\text{ V}$ , $T_J = 150^\circ\text{C}$
Power Dissipation	$P_{DISS\ CC}$	–	–	1	W	$V_{DDP} = 3.6\text{ V}$ , $T_J = 150^\circ\text{C}$
Wake-up time from Sleep to Active mode	$t_{SSA\ CC}$	–	6	–	cycles	
Wake-up time from Deep Sleep to Active mode		–	–	–	ms	Defined by the wake-up of the Flash module, see <a href="#">Section 3.2.11</a>
Wake-up time from Hibernate mode		–	–	–	ms	Wake-up via power-on reset event, see <a href="#">Section 3.3.2</a>

- 1) CPU executing code from Flash, all peripherals idle.
- 2) CPU executing code from Flash. USB and CCU clock off.
- 3) CPU in sleep, all peripherals idle, Flash in Active mode.
- 4) CPU in sleep, Flash in Active mode.
- 5) CPU in sleep, peripherals disabled, after wake-up code execution from RAM.
- 6) To wake-up the Flash from its Sleep mode,  $f_{CPU} \geq 1\text{ MHz}$  is required.
- 7) OSC\_ULP operating with external crystal on RTC\_XTAL.
- 8) OSC\_ULP off, Hibernate domain operating with OSC\_SI clock.
- 9) Test Power Loop:  $f_{SYS} = 80\text{ MHz}$ , CPU executing benchmark code from Flash, all CCUs in 100 kHz timer mode, all ADC groups in continuous conversion mode, USICs as SPI in internal loop-back mode, CAN in 500 kHz internal loop-back mode, interrupt triggered DMA block transfers to parity protected RAMs and FCE, DTS measurements and FPU calculations.  
The power consumption of each customer application will most probably be lower than this value, but must be evaluated separately.
- 10)  $I_{DDP}$  decreases typically by 3.5 mA when  $f_{SYS}$  decreases by 10 MHz, at constant  $T_J$ .
- 11) Sum of currents of all active converters (ADC and DAC).

### Peripheral Idle Currents

Test conditions:

- $f_{SYS}$  and derived clocks at 80 MHz
- $V_{DDP} = 3.3\text{ V}$ ,  $T_a = 25^\circ\text{C}$
- all peripherals are held in reset (see the PRSTAT registers in the Reset Control Unit of the SCU)
- the peripheral clocks are disabled (see CGATSTAT registers in the Clock Control Unit of the SCU)
- no I/O activity
- the given values are a result of differential measurements with asserted and deasserted peripheral reset and enabled clock of the peripheral under test

The tested peripheral is left in the state after the peripheral reset is deasserted, no further initialisation or configuration is done. For example no timer is running in the CCUs, no communication active in the USICs, etc.

### 3 Electrical Parameters

**Table 39 Peripheral Idle Currents**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
PORTS	$I_{PER\ CC}$	–	$\leq 0.3$	–	mA	
USB						
FCE						
WDT						
POSIFx <sup>1)</sup>						
MultiCAN		–	$\leq 1.0$	–		
ERU						
LEDTSCU0						
CCU4x <sup>1)</sup>						
CCU8x <sup>1)</sup>						
DAC (digital) <sup>2)</sup>		–	1.3	–		
USICx		–	3.0	–		
VADC (digital) <sup>2)</sup>		–	4.5	–		
DMax		–	6.0	–		

1) Enabling the  $f_{CCU}$  clock for the POSIFx/CCU4x/CCU8x modules adds approximately  $I_{PER} = 1.8$  mA, disregarding which and how many of those peripherals are enabled.

2) The current consumption of the analog components are given in the dedicated Datasheet sections of the respective peripheral.

### 3.2.11 Flash Memory Parameters

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 40 Flash Memory Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Erase Time per 256 Kbyte Sector	$t_{ERP\ CC}$	–	5	5.5	s	
Erase Time per 64 Kbyte Sector	$t_{ERP\ CC}$	–	1.2	1.4	s	
Erase Time per 16 Kbyte Logical Sector	$t_{ERP\ CC}$	–	0.3	0.4	s	
Program time per page <sup>1)</sup>	$t_{PRP\ CC}$	–	5.5	11	ms	
Erase suspend delay	$t_{FL\_ErSusp\ CC}$	–	–	15	ms	
Wait time after margin change	$t_{FL\_MarginDel\ CC}$	10	–	–	μs	
Wake-up time	$t_{WU\ CC}$	–	–	270	μs	

(table continues...)

### 3 Electrical Parameters

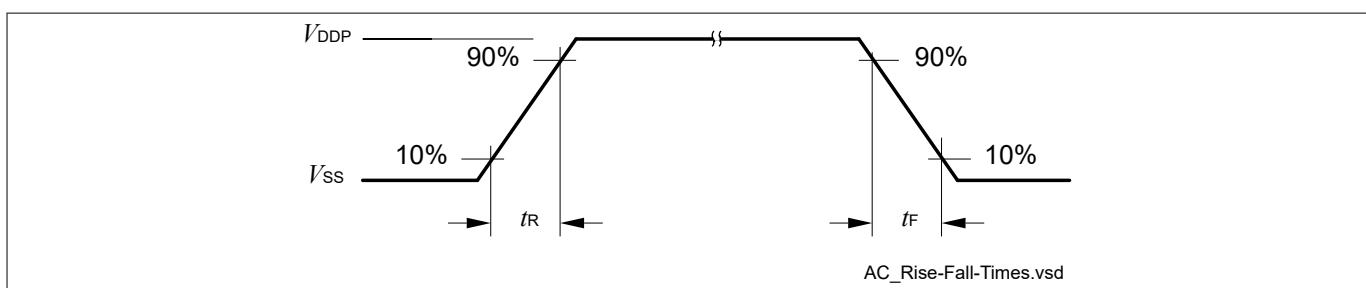
**Table 40** (continued) Flash Memory Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Read access time	$t_a$ CC	20	–	–	ns	For operation with $1/f_{CPU} < t_a$ wait states must be configured <sup>2)</sup>
Data Retention Time, Physical Sector <sup>3) 4)</sup>	$t_{RET}$ CC	20	–	–	years	Max. 1000 erase/program cycles
Data Retention Time, Logical Sector <sup>3) 4)</sup>	$t_{RETL}$ CC	20	–	–	years	Max. 100 erase/program cycles
Data Retention Time, User Configuration Block (UCB) <sup>3) 4)</sup>	$t_{RTU}$ CC	20	–	–	years	Max. 4 erase/program cycles per UCB
Endurance on 64 Kbyte Physical Sector PS4	$N_{EPS4}$ CC	10000	–	–	cycles	BA-marking devices only! Cycling distributed over life time <sup>5)</sup>

- 1) In case the Program Verify feature detects weak bits, these bits will be programmed once more. The reprogramming takes an additional time of 5.5 ms.
- 2) The following formula applies to the wait state configuration:  $FCON.WSPFLASH \times (1/f_{CPU}) \geq t_a$
- 3) Storage and inactive time included.
- 4) Values given are valid for an average weighted junction temperature of  $T_J = 110^{\circ}\text{C}$ .
- 5) Only valid with robust EEPROM emulation algorithm, equally cycling the logical sectors. For more details see the Reference Manual.

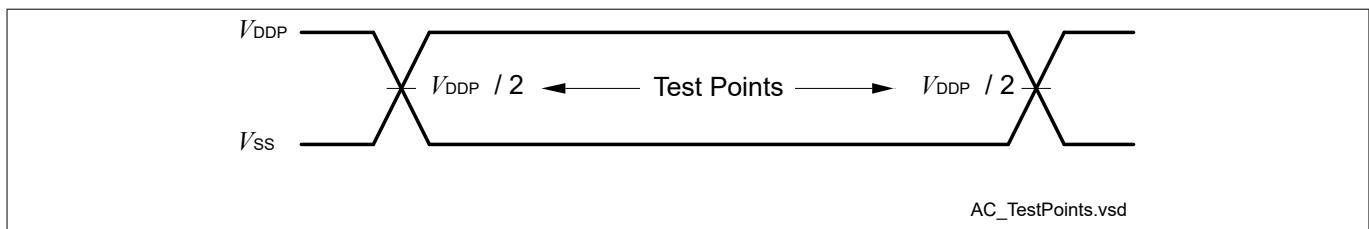
### 3.3 AC Parameters

#### 3.3.1 Testing Waveforms

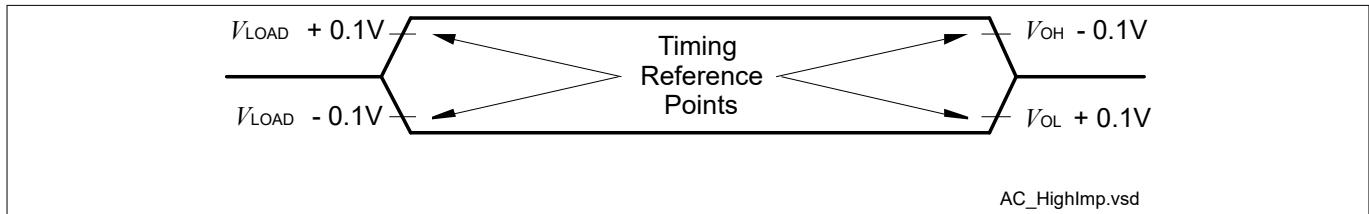


**Figure 23** Rise/Fall Time Parameters

### 3 Electrical Parameters



**Figure 24** Testing Waveform, Output Delay

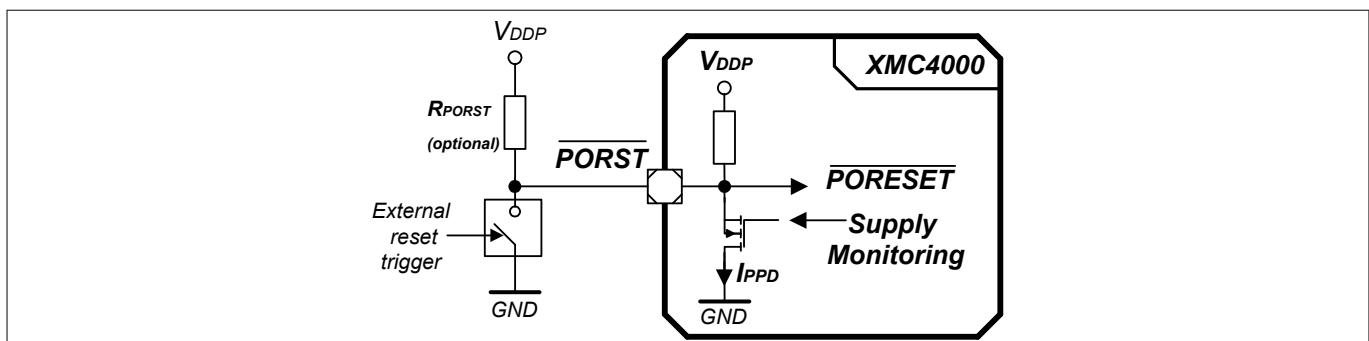


**Figure 25** Testing Waveform, Output High Impedance

#### 3.3.2 Power-Up and Supply Monitoring

PORST is always asserted when  $V_{DDP}$  and/or  $V_{DDC}$  violate the respective thresholds.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.



**Figure 26** PORST Circuit

**Table 41** Supply Monitoring Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Digital supply voltage reset threshold	$V_{POR\ CC}$	2.79 <sup>1)</sup>	-	3.05 <sup>2)</sup>	V	<sup>3)</sup>
Core supply voltage reset threshold	$V_{PV\ CC}$	-	-	1.17	V	
$V_{DDP}$ voltage to ensure defined pad states	$V_{DDPPA\ CC}$	-	1.0	-	V	
PORST rise time	$t_{PR\ SR}$	-	-	2	$\mu s$	

(table continues...)

### 3 Electrical Parameters

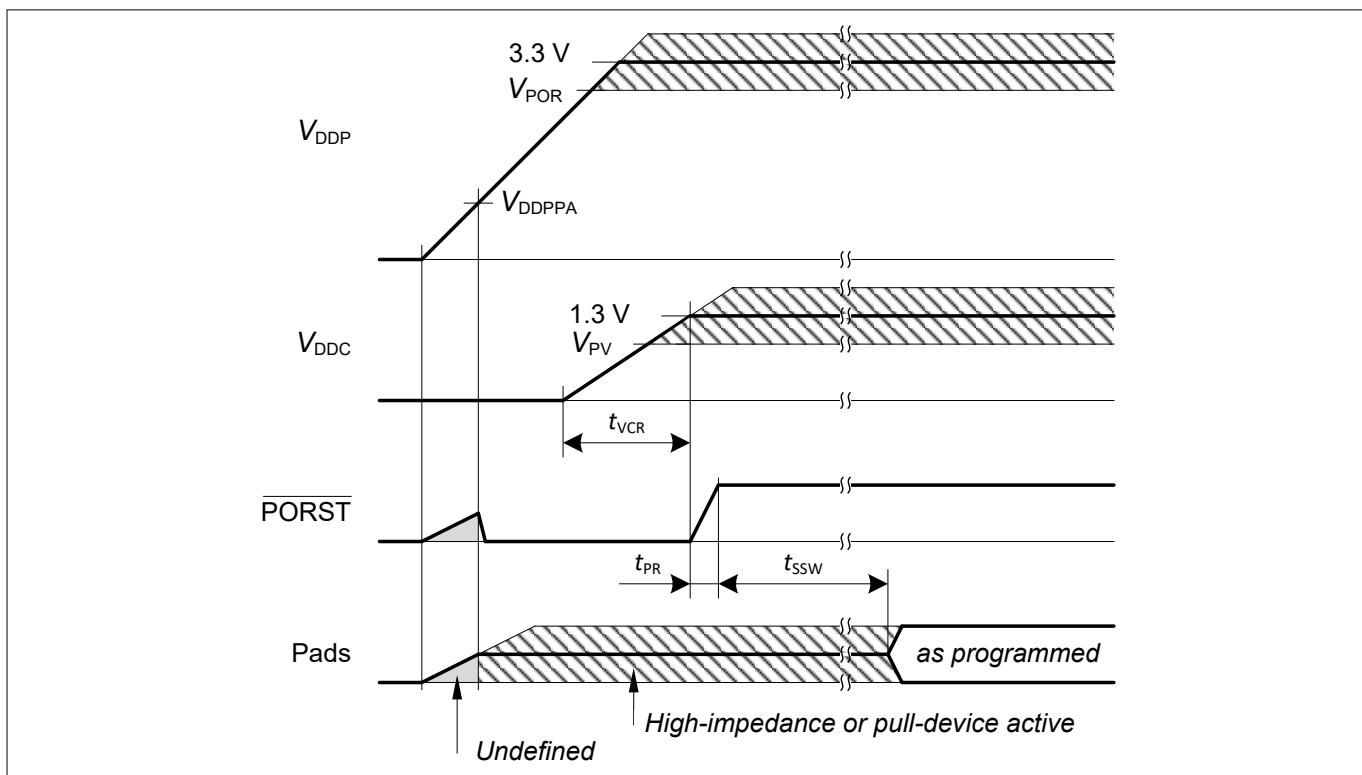
**Table 41 (continued) Supply Monitoring Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Startup time from power-on reset with code execution from Flash	$t_{SSW\ CC}$	–	2.5	3.5	ms	Time to the first user code instruction
$V_{DDC}$ ramp up time	$t_{VCR\ CC}$	–	550	–	μs	Ramp up after power-on or after a reset triggered by a violation of $V_{POR}$ or $V_{PV}$

1) Minimum threshold for reset assertion.

2) Maximum threshold for reset deassertion.

3) The  $V_{DDP}$  monitoring has a typical hysteresis of  $V_{PORHYS} = 180$  mV.



**Figure 27 Power-Up Behavior**

### 3 Electrical Parameters

#### 3.3.3 Power Sequencing

While starting up and shutting down as well as when switching power modes of the system it is important to limit the current load steps. A typical cause for such load steps is changing the CPU frequency  $f_{CPU}$ . Load steps exceeding the below defined values may cause a power on reset triggered by the supply monitor.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 42 Power Sequencing Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Positive Load Step Current	$\Delta I_{PLS\ SR}$	–	–	50	mA	Load increase on $V_{DDP}$ $\Delta t \leq 10$ ns
Negative Load Step Current	$\Delta I_{NLS\ SR}$	–	–	150	mA	Load decrease on $V_{DDP}$ $\Delta t \leq 10$ ns
$V_{DDC}$ Voltage Over-/ Undershoot from Load Step	$\Delta V_{LS\ CC}$	–	–	$\pm 100$	mV	For maximum positive or negative load step
Positive Load Step Settling Time	$t_{PLSS\ SR}$	50	–	–	$\mu s$	
Negative Load Step Settling Time	$t_{NLSS\ SR}$	100	–	–	$\mu s$	
External Buffer Capacitor on $V_{DDC}$	$C_{EXT\ SR}$	3	4.7	6	$\mu F$	In addition $C = 100$ nF capacitor on each $V_{DDC}$ pin

#### Positive Load Step Examples

System assumptions:

$f_{CPU} = f_{SYS}$ , target frequency  $f_{CPU} = 80$  MHz, main PLL  $f_{VCO} = 480$  MHz, stepping done by K2 divider,  $t_{PLSS}$  between individual steps:

24 MHz - 48 MHz - 80 MHz (K2 steps 20 - 10 - 6)

24 MHz - 60 MHz - 80 MHz (K2 steps 20 - 8 - 6)

### 3 Electrical Parameters

#### 3.3.4 Phase Locked Loop (PLL) Characteristics

##### Main and USB PLL

**Table 43 PLL Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Accumulated Jitter	$D_P$ CC	–	–	$\pm 5$	ns	accumulated over 300 cycles $f_{SYS} = 80$ MHz
Duty Cycle <sup>1)</sup>	$D_{DC}$ CC	46	50	54	%	Low pulse to total period, assuming an ideal input clock source
PLL base frequency	$f_{PLLBASE}$ CC	30	–	140	MHz	
VCO input frequency	$f_{REF}$ CC	4	–	16	MHz	
VCO frequency range	$f_{VCO}$ CC	260	–	520	MHz	
PLL lock-in time	$t_L$ CC	–	–	400	$\mu$ s	

1) 50% for even K2 divider values,  $50 \pm (10/K2)$  for odd K2 divider values.

### 3 Electrical Parameters

#### 3.3.5 Internal Clock Source Characteristics

##### Fast Internal Clock Source

**Table 44** Fast Internal Clock Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Nominal frequency	$f_{OFINC\ CC}$	-	36.5	-	MHz	not calibrated
		-	24	-	MHz	calibrated
Accuracy	$\Delta f_{OFI\ CC}$	-0.5	-	0.5	%	automatic calibration <sup>1) 2)</sup>
		-15	-	15	%	factory calibration, $V_{DDP} = 3.3\text{ V}$
		-25	-	25	%	no calibration, $V_{DDP} = 3.3\text{ V}$
		-7	-	7	%	Variation over voltage range <sup>3)</sup> $3.13\text{ V} \leq V_{DDP} \leq 3.63\text{ V}$
Start-up time	$t_{OFIS\ CC}$	-	50	-	$\mu\text{s}$	

1) Error in addition to the accuracy of the reference clock.

2) Automatic calibration compensates variations of the temperature and in the  $V_{DDP}$  supply voltage.

3) Deviations from the nominal  $V_{DDP}$  voltage induce an additional error to the uncalibrated and/or factory calibrated oscillator frequency.

##### Slow Internal Clock Source

**Table 45** Slow Internal Clock Parameters

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Nominal frequency	$f_{OSI\ CC}$	-	32.768	-	kHz	
Accuracy	$\Delta f_{OSI\ CC}$	-4	-	4	%	$V_{BAT} = \text{const.}$ $0^\circ\text{C} \leq T_A \leq 85^\circ\text{C}$
		-5	-	5	%	$V_{BAT} = \text{const.}$ $T_A < 0^\circ\text{C}$ or $T_A > 85^\circ\text{C}$
		-5	-	5	%	$2.4\text{ V} \leq V_{BAT},$ $T_A = 25^\circ\text{C}$
		-10	-	10	%	$1.95\text{ V} \leq V_{BAT} < 2.4\text{ V},$ $T_A = 25^\circ\text{C}$
Start-up time	$t_{OSIS\ CC}$	-	50	-	$\mu\text{s}$	

### 3 Electrical Parameters

#### 3.3.6 JTAG Interface Timing

The following parameters are applicable for communication through the JTAG debug interface. The JTAG module is fully compliant with IEEE1149.1-2000.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

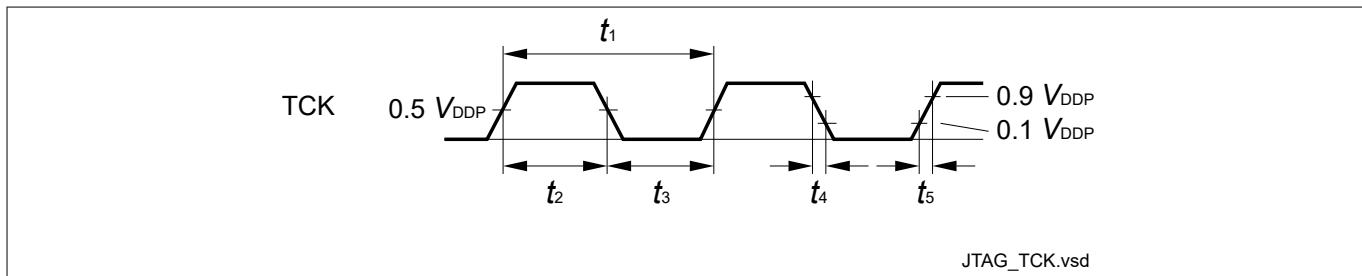
**Note:** Operating conditions apply.

**Table 46 JTAG Interface Timing Parameters**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
TCK clock period	$t_1$ SR	30	-	-	ns	For $C_L = 20 \text{ pF}$ on TDO
TCK clock period	$t_1$ SR	40	-	-	ns	For $C_L = 50 \text{ pF}$ on TDO
TCK high time	$t_2$ SR	10	-	-	ns	
TCK low time	$t_3$ SR	10	-	-	ns	
TCK clock rise time	$t_4$ SR	-	-	4	ns	
TCK clock fall time	$t_5$ SR	-	-	4	ns	
TDI/TMS setup to TCK rising edge	$t_6$ SR	6	-	-	ns	
TDI/TMS hold after TCK rising edge	$t_7$ SR	6	-	-	ns	
TDO valid after TCK falling edge <sup>1)</sup> (propagation delay)	$t_8$ CC	-	-	17	ns	$C_L = 50 \text{ pF}$
		3	-	-	ns	$C_L = 20 \text{ pF}$
TDO hold after TCK falling edge <sup>1)</sup>	$t_{18}$ CC	2	-	-	ns	
TDO high imped. to valid from TCK falling edge <sup>1) 2)</sup>	$t_9$ CC	-	-	14	ns	$C_L = 50 \text{ pF}$
TDO valid to high imped. from TCK falling edge <sup>1)</sup>	$t_{10}$ CC	-	-	13.5	ns	$C_L = 50 \text{ pF}$

1) The falling edge on TCK is used to generate the TDO timing.

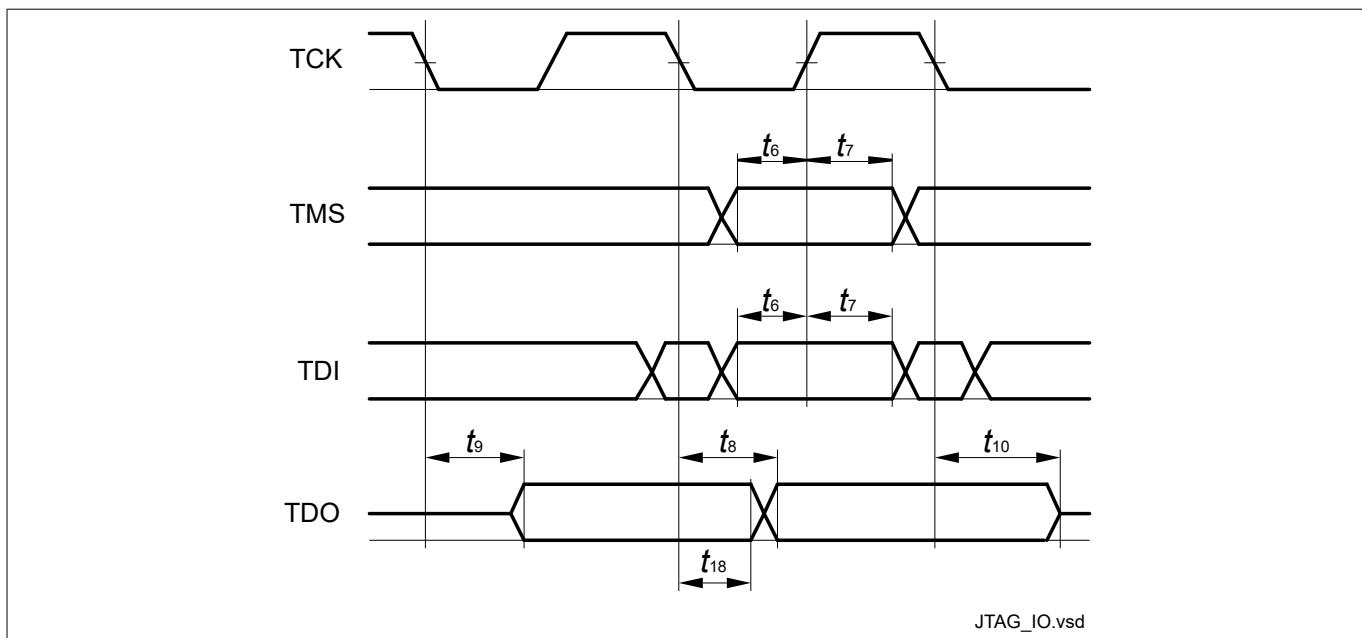
2) The setup time for TDO is given implicitly by the TCK cycle time.



**Figure 28**

**Test Clock Timing (TCK)**

### 3 Electrical Parameters



**Figure 29** JTAG Timing

#### 3.3.7 Serial Wire Debug Port (SW-DP) Timing

The following parameters are applicable for communication through the SW-DP interface.

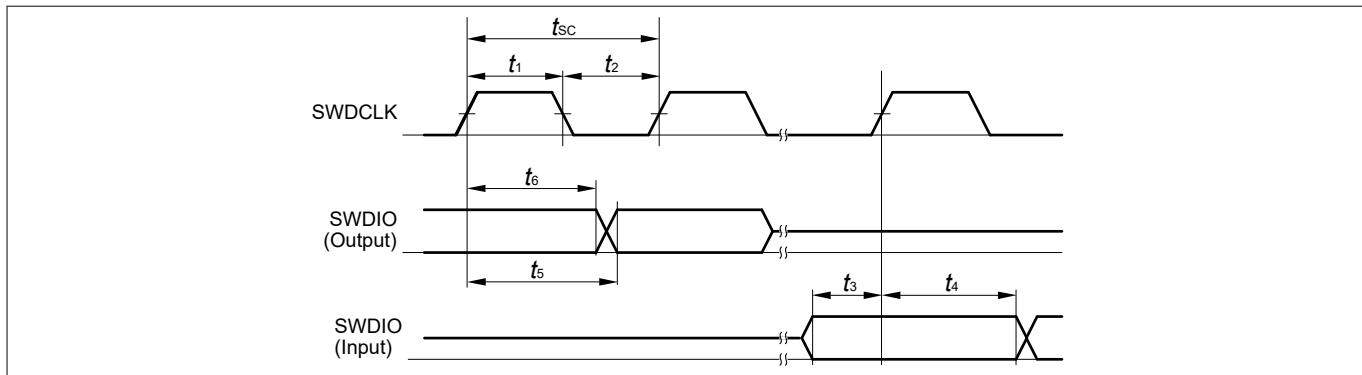
**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Note:** Operating conditions apply.

**Table 47** SWD Interface Timing Parameters (Operating Conditions apply)

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
SWDCLK clock period	$t_{SC}$ SR	25	–	–	ns	$C_L = 30 \text{ pF}$
		40	–	–	ns	$C_L = 50 \text{ pF}$
SWDCLK high time	$t_1$ SR	10	–	500000	ns	
SWDCLK low time	$t_2$ SR	10	–	500000	ns	
SWDIO input setup to SWDCLK rising edge	$t_3$ SR	6	–	–	ns	
SWDIO input hold after SWDCLK rising edge	$t_4$ SR	6	–	–	ns	
SWDIO output valid time after SWDCLK rising edge	$t_5$ CC	–	–	17	ns	$C_L = 50 \text{ pF}$
		–	–	13	ns	$C_L = 30 \text{ pF}$
SWDIO output hold time from SWDCLK rising edge	$t_6$ CC	3	–	–	ns	

### 3 Electrical Parameters



**Figure 30** SWD Timing

#### 3.3.8 Peripheral Timing

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Note:** Operating conditions apply.

##### 3.3.8.1 Synchronous Serial Interface (USIC SSC) Timing

The following parameters are applicable for a USIC channel operated in SSC mode.

**Note:** Operating Conditions apply.

**Table 48** USIC SSC Master Mode Timing

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
SCLKOUT master clock period	$t_{CLK\ CC}$	40	-	-	ns	
Slave select output SELO active to first SCLKOUT transmit edge	$t_1\ CC$	$t_{SYS} - 6.5^{(1)}$	-	-	ns	
Slave select output SELO inactive after last SCLKOUT receive edge	$t_2\ CC$	$t_{SYS} - 8.5^{(1)}$	-	-	ns	
Data output DOUT[3:0] valid time	$t_3\ CC$	-6	-	8	ns	
Receive data input DX0/DX[5:3] setup time to SCLKOUT receive edge	$t_4\ SR$	23	-	-	ns	
Data input DX0/DX[5:3] hold time from SCLKOUT receive edge	$t_5\ SR$	1	-	-	ns	

(1)  $t_{SYS} = 1/f_{PB}$

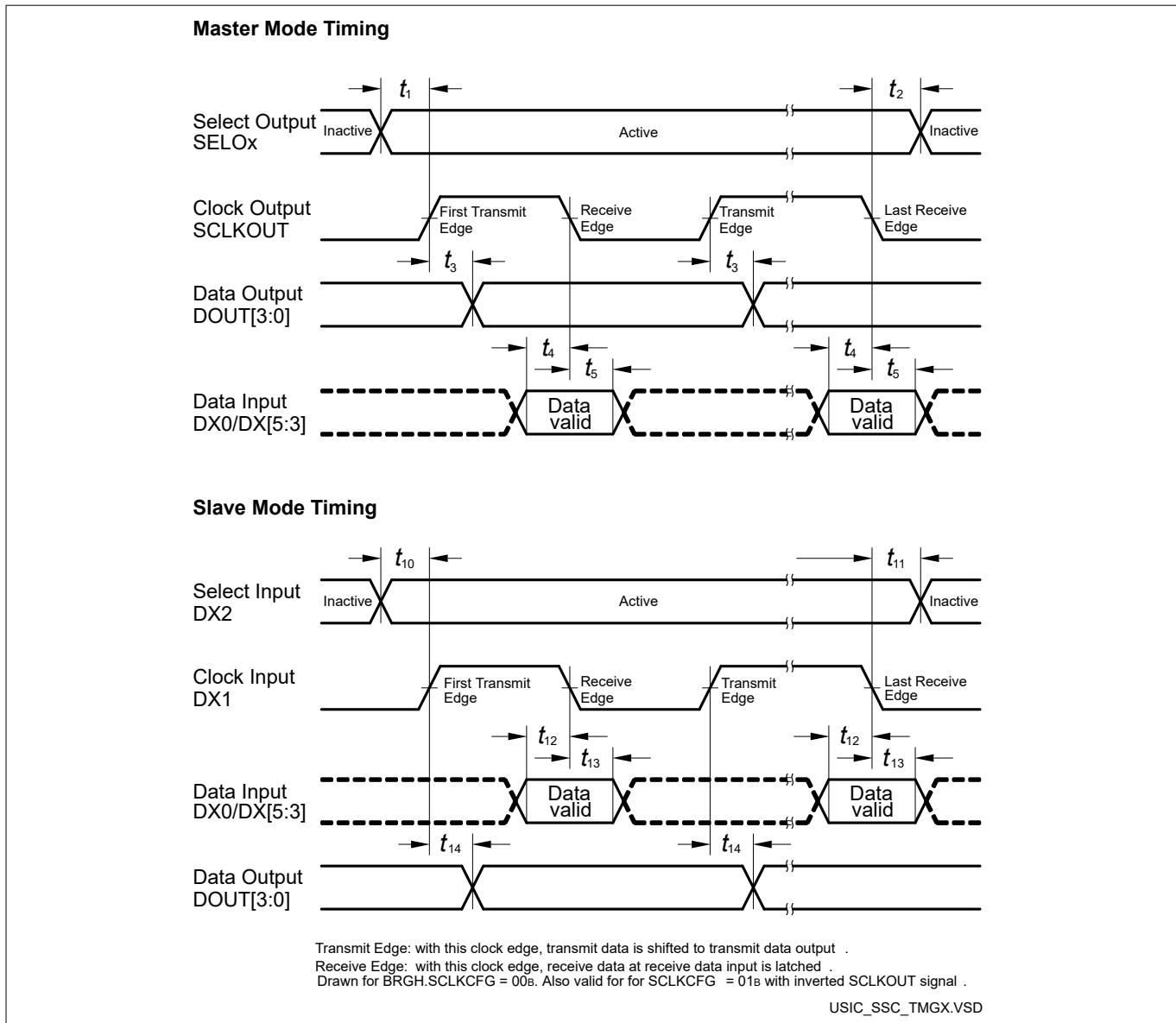
### 3 Electrical Parameters

**Table 49 USIC SSC Slave Mode Timing**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
DX1 slave clock period	$t_{CLK\ SR}$	66.6	–	–	ns	
Select input DX2 setup to first clock input DX1 transmit edge <sup>1)</sup>	$t_{10\ SR}$	3	–	–	ns	
Select input DX2 hold after last clock input DX1 receive edge <sup>1)</sup>	$t_{11\ SR}$	4	–	–	ns	
Receive data input DX0/DX[5:3] setup time to shift clock receive edge <sup>1)</sup>	$t_{12\ SR}$	6	–	–	ns	
Data input DX0/DX[5:3] hold time from clock input DX1 receive edge <sup>1)</sup>	$t_{13\ SR}$	4	–	–	ns	
Data output DOUT[3:0] valid time	$t_{14\ CC}$	0	–	24	ns	

1) These input timing are valid for asynchronous input signal handling of slave select input, shift clock input, and receive data input (bits DXnCR.DSEN = 0).

### 3 Electrical Parameters



**Figure 31      USIC - SSC Master/Slave Mode Timing**

**Note:** This timing diagram shows a standard configuration, for which the slave select signal is low-active, and the serial clock signal is not shifted and not inverted.

### 3 Electrical Parameters

#### 3.3.8.2 Inter-IC (IIC) Interface Timing

The following parameters are applicable for a USIC channel operated in IIC mode.

**Note:** Operating Conditions apply.

**Table 50** **USIC IIC Standard Mode Timing<sup>1)</sup>**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Fall time of both SDA and SCL	$t_1$ CC/SR	–	–	300	ns	
Rise time of both SDA and SCL	$t_2$ CC/SR	–	–	1000	ns	
Data hold time	$t_3$ CC/SR	0	–	–	μs	
Data set-up time	$t_4$ CC/SR	250	–	–	ns	
LOW period of SCL clock	$t_5$ CC/SR	4.7	–	–	μs	
HIGH period of SCL clock	$t_6$ CC/SR	4.0	–	–	μs	
Hold time for (repeated) START condition	$t_7$ CC/SR	4.0	–	–	μs	
Set-up time for repeated START condition	$t_8$ CC/SR	4.7	–	–	μs	
Set-up time for STOP condition	$t_9$ CC/SR	4.0	–	–	μs	
Bus free time between a STOP and START condition	$t_{10}$ CC/SR	4.7	–	–	μs	
Capacitive load for each bus line	$C_b$ SR	–	–	400	pF	

- 1) Due to the wired-AND configuration of an IIC bus system, the port drivers of the SCL and SDA signal lines need to operate in open-drain mode. The high level on these lines must be held by an external pull-up device, approximately 10 kOhm for operation at 100 kbit/s, approximately 2 kOhm for operation at 400 kbit/s.

**Table 51** **USIC IIC Fast Mode Timing<sup>1)</sup>**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Fall time of both SDA and SCL	$t_1$ CC/SR	$20 + 0.1 * C_b$ <sup>2)</sup>	–	300	ns	
Rise time of both SDA and SCL	$t_2$ CC/SR	$20 + 0.1 * C_b$ <sup>2)</sup>	–	300	ns	
Data hold time	$t_3$ CC/SR	0	–	–	μs	
Data set-up time	$t_4$ CC/SR	100	–	–	ns	
LOW period of SCL clock	$t_5$ CC/SR	1.3	–	–	μs	

(table continues...)

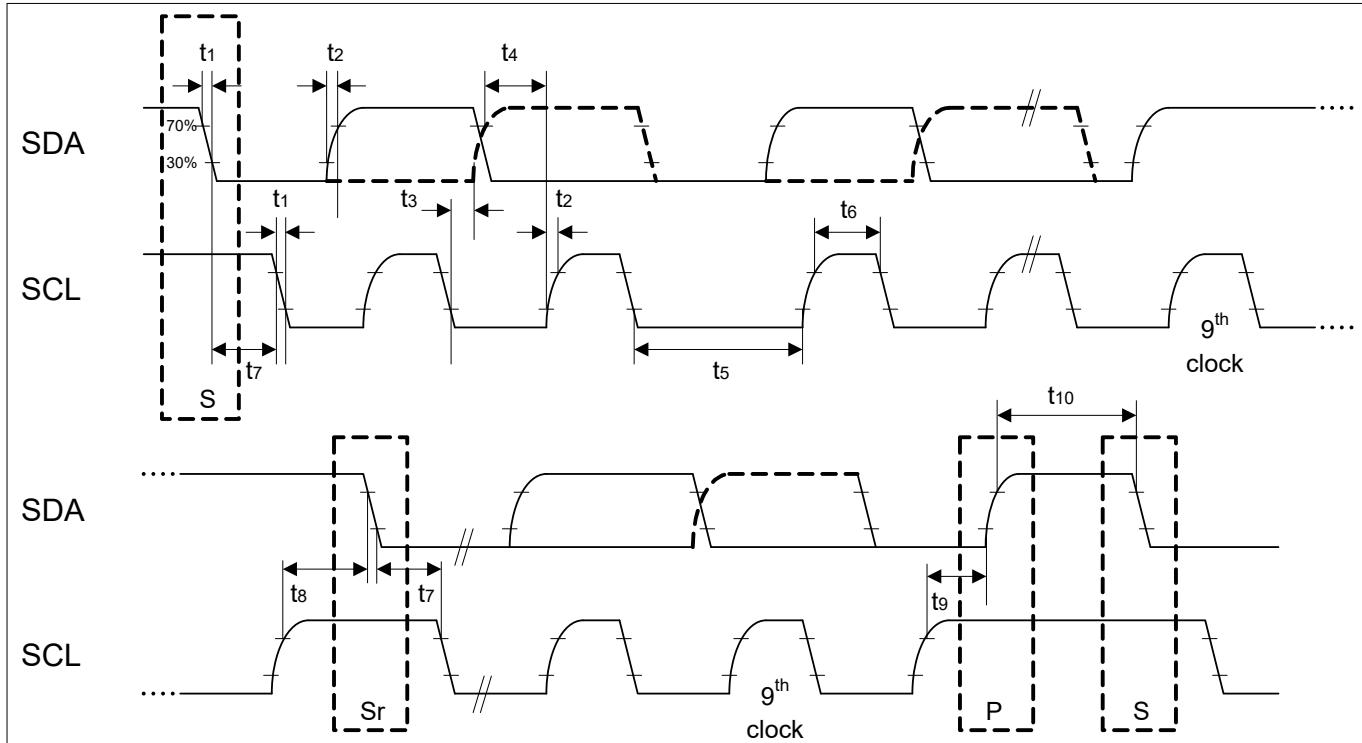
### 3 Electrical Parameters

**Table 51** (continued) USIC IIC Fast Mode Timing<sup>1)</sup>

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
HIGH period of SCL clock	$t_6$ CC/SR	0.6	-	-	μs	
Hold time for (repeated) START condition	$t_7$ CC/SR	0.6	-	-	μs	
Set-up time for repeated START condition	$t_8$ CC/SR	0.6	-	-	μs	
Set-up time for STOP condition	$t_9$ CC/SR	0.6	-	-	μs	
Bus free time between a STOP and START condition	$t_{10}$ CC/SR	1.3	-	-	μs	
Capacitive load for each bus line	$C_b$ SR	-	-	400	pF	

1) Due to the wired-AND configuration of an IIC bus system, the port drivers of the SCL and SDA signal lines need to operate in open-drain mode. The high level on these lines must be held by an external pull-up device, approximately 10 kOhm for operation at 100 kbit/s, approximately 2 kOhm for operation at 400 kbit/s.

2)  $C_b$  refers to the total capacitance of one bus line in pF.



**Figure 32** USIC IIC Stand and Fast Mode Timing

### 3 Electrical Parameters

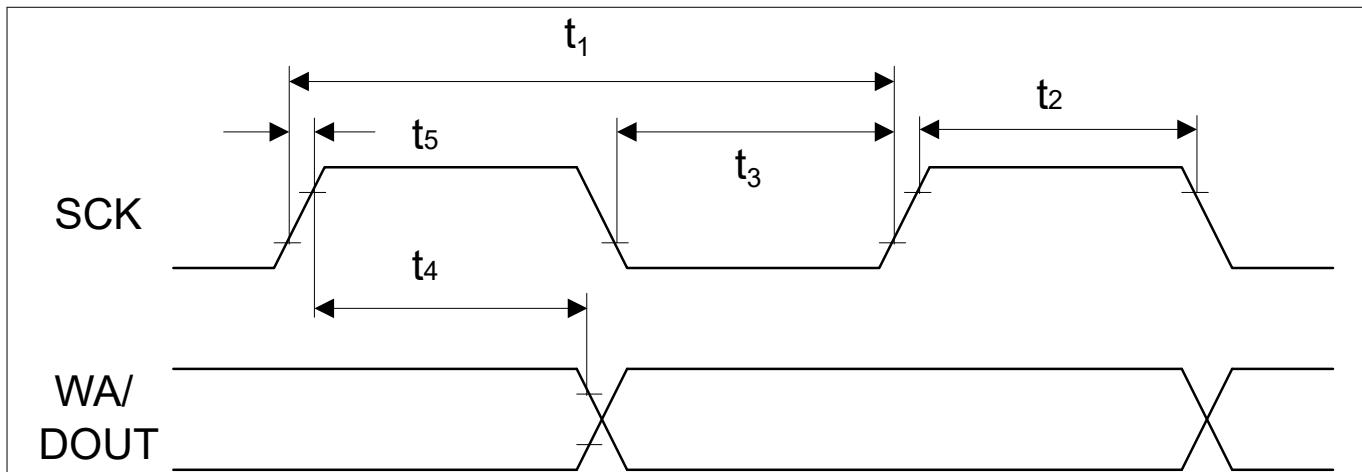
#### 3.3.8.3 Inter-IC Sound (IIS) Interface Timing

The following parameters are applicable for a USIC channel operated in IIS mode.

**Note:** Operating Conditions apply.

**Table 52** USIC IIS Master Transmitter Timing

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Clock period	$t_1$ CC	33.3	–	–	ns	
Clock high time	$t_2$ CC	$0.35 \times t_{1\min}$	–	–	ns	
Clock low time	$t_3$ CC	$0.35 \times t_{1\min}$	–	–	ns	
Hold time	$t_4$ CC	0	–	–	ns	
Clock rise time	$t_5$ CC	–	–	$0.15 \times t_{1\min}$	ns	



**Figure 33** USIC IIS Master Transmitter Timing

**Table 53** USIC IIS Slave Receiver Timing

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Clock period	$t_6$ SR	66.6	–	–	ns	
Clock high time	$t_7$ SR	$0.35 \times t_{6\min}$	–	–	ns	
Clock low time	$t_8$ SR	$0.35 \times t_{6\min}$	–	–	ns	
Set-up time	$t_9$ SR	$0.2 \times t_{6\min}$	–	–	ns	
Hold time	$t_{10}$ SR	0	–	–	ns	

### 3 Electrical Parameters

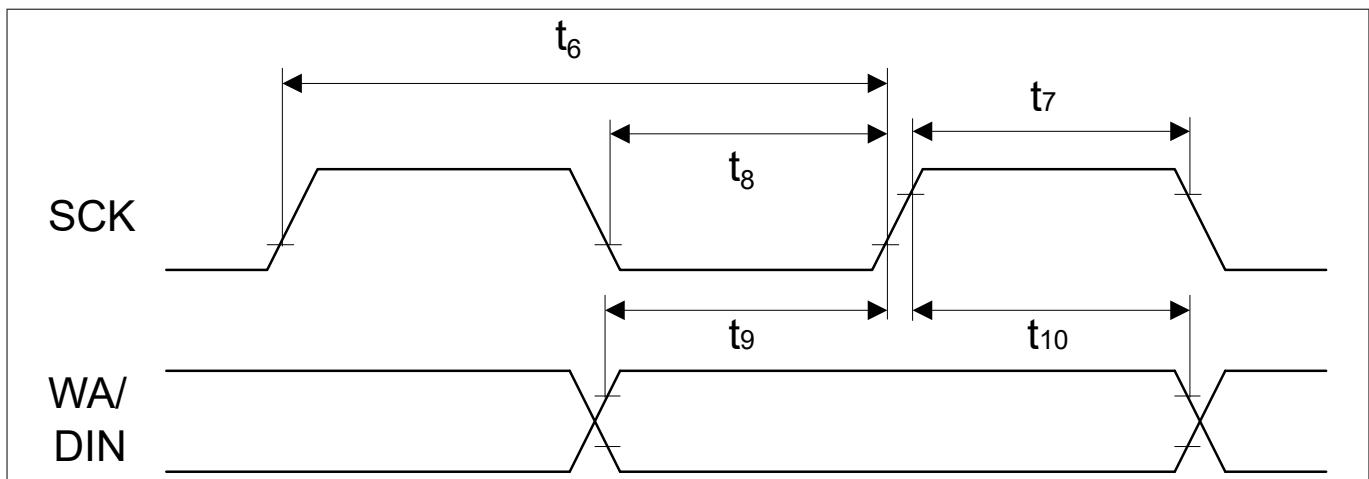


Figure 34      **USIC IIS Slave Receiver Timing**

#### 3.3.9      **USB Interface Characteristics**

The Universal Serial Bus (USB) Interface is compliant to the USB Rev. 2.0 Specification. High-Speed Mode is not supported.

**Note:** These parameters are not subject to production test, but verified by design and/or characterization.

**Table 54      USB Timing Parameters (operating conditions apply)**

Parameter	Symbol	Values			Unit	Note/Test Condition
		Min.	Typ.	Max.		
Rise time	$t_R$ CC	4	-	20	ns	$C_L = 50 \text{ pF}$
Fall time	$t_F$ CC	4	-	20	ns	$C_L = 50 \text{ pF}$
Rise/Fall time matching	$t_R/t_F$ CC	90	-	111.11	%	$C_L = 50 \text{ pF}$
Crossover voltage	$V_{CRS}$ CC	1.3	-	2.0	V	$C_L = 50 \text{ pF}$

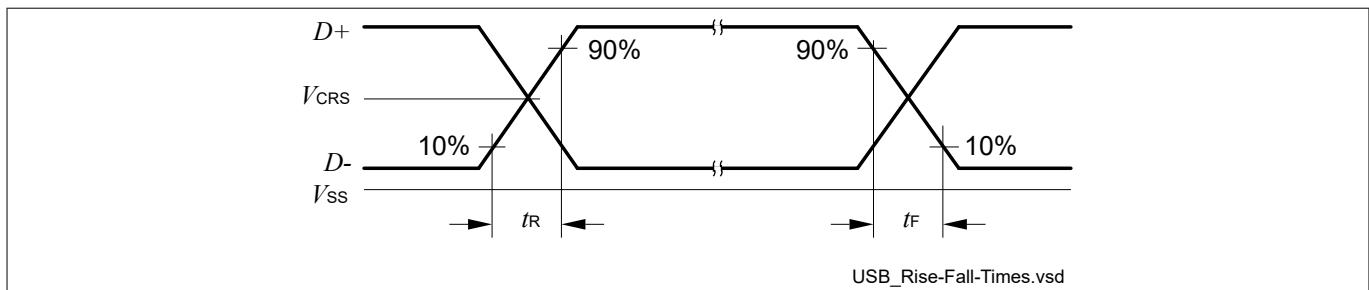


Figure 35      **USB Signal Timing**

## 4 Package and Reliability

### 4.1 Package and Reliability

The XMC4[12]00 is a member of the XMC4000 Family of microcontrollers. It is also compatible to a certain extent with members of similar families or subfamilies.

Each package is optimized for the device it houses. Therefore, there may be slight differences between packages of the same pin-count but for different device types. In particular, the size of the Exposed Die Pad may vary.

If different device types are considered or planned for an application, it must be ensured that the board layout fits all packages under consideration.

#### 4.1.1 Package Parameters

**Table 55** provides the thermal characteristics of the packages used in XMC4[12]00. The availability of different packages for different markings is listed in [Table 2](#).

**Table 55 Thermal Characteristics of the Packages**

Parameter	Symbol	Limit Values		Unit	Package Types
		Min.	Max.		
Exposed Die Pad Dimensions	$Ex \times Ey$ CC	–	$5.7 \times 5.7$	mm	PG-TQFP-64-19 PG-TQFP-64-21
		–	$5.2 \times 5.2$	mm	PG-VQFN-48-71
Thermal resistance Junction-Ambient	$R_{\Theta JA}$ CC	–	23.4	K/W	PG-TQFP-64-19 <sup>1)</sup> PG-TQFP-64-21 <sup>1)</sup>
		–	34.8	K/W	PG-VQFN-48-71 <sup>1)</sup>
Package thickness $1.0 \pm 0.05$ mm	–	–	1.2 Max	mm	PG-TQFP-64-19 PG-TQFP-64-21

1) Device mounted on a 4-layer JEDEC board (JESD 51-7) with thermal vias; exposed pad soldered.

**Note:** For electrical reasons, it is required to connect the exposed pad to the board ground  $V_{SS}$ , independent of EMC and thermal requirements.

#### 4.1.1 Thermal Considerations

When operating the XMC4[12]00 in a system, the total heat generated in the chip must be dissipated to the ambient environment to prevent overheating and the resulting thermal damage.

The maximum heat that can be dissipated depends on the package and its integration into the target board. The “Thermal resistance  $R_{\Theta JA}$ ” quantifies these parameters. The power dissipation must be limited so that the average junction temperature does not exceed 150°C.

The difference between junction temperature and ambient temperature is determined by

$$\Delta T = (P_{INT} + P_{IOSTAT} + P_{IODYN}) \times R_{\Theta JA}$$

The internal power consumption is defined as

$$P_{INT} = V_{DDP} \times I_{DDP} \text{ (switching current and leakage current).}$$

The static external power consumption caused by the output drivers is defined as

$$P_{IOSTAT} = \sum((V_{DDP} - V_{OH}) \times I_{OH}) + \sum(V_{OL} \times I_{OL})$$

The dynamic external power consumption caused by the output drivers ( $P_{IODYN}$ ) depends on the capacitive load connected to the respective pins and their switching frequencies.

## 4 Package and Reliability

If the total power dissipation for a given system configuration exceeds the defined limit, countermeasures must be taken to ensure proper system operation:

- Reduce  $V_{DDP}$ , if possible in the system
- Reduce the system frequency
- Reduce the number of output pins
- Reduce the load on active output drivers

### 4.2 Package Outlines

The availability of different packages for different devices types is listed in [Table 1](#), specific packages for different device markings are listed in [Table 2](#).

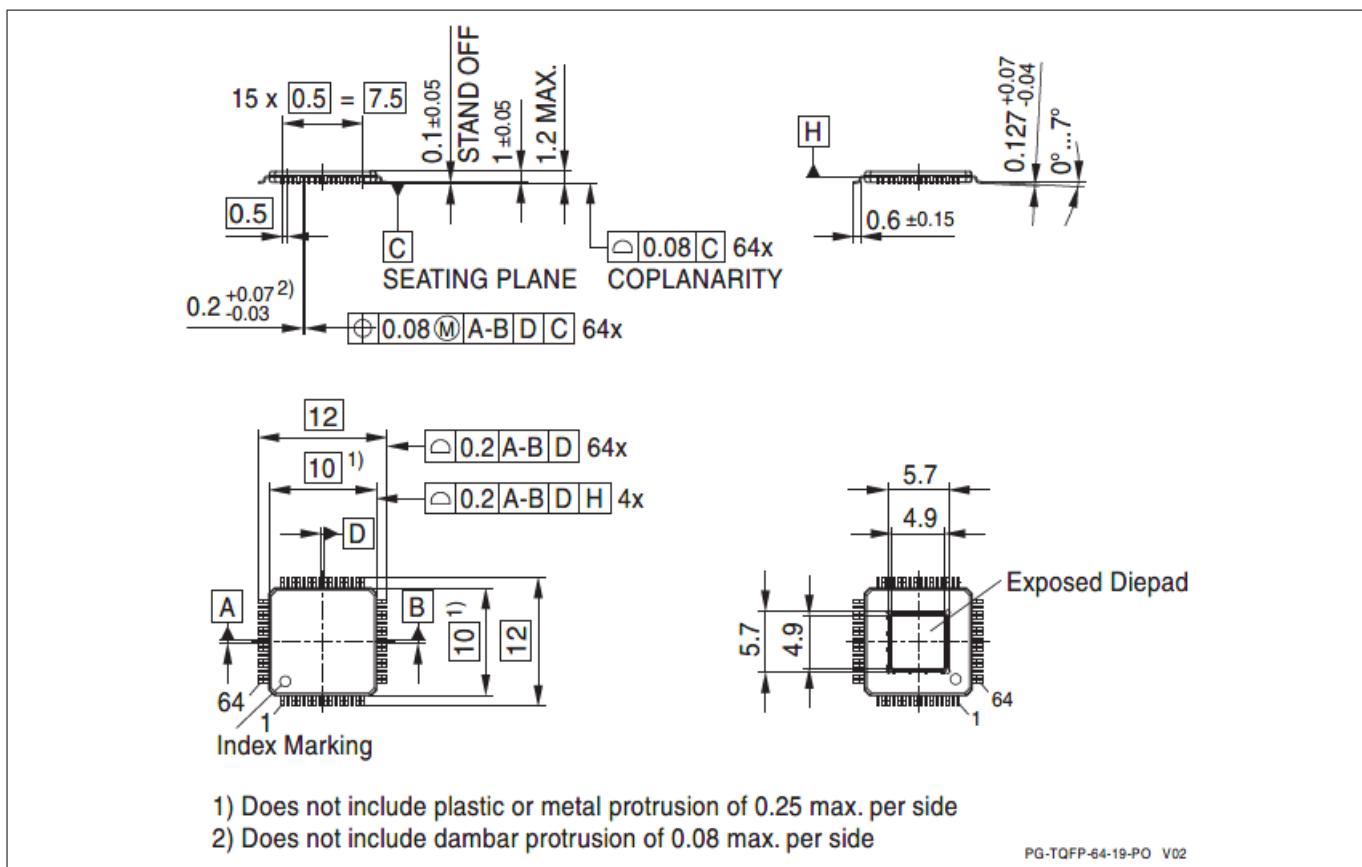
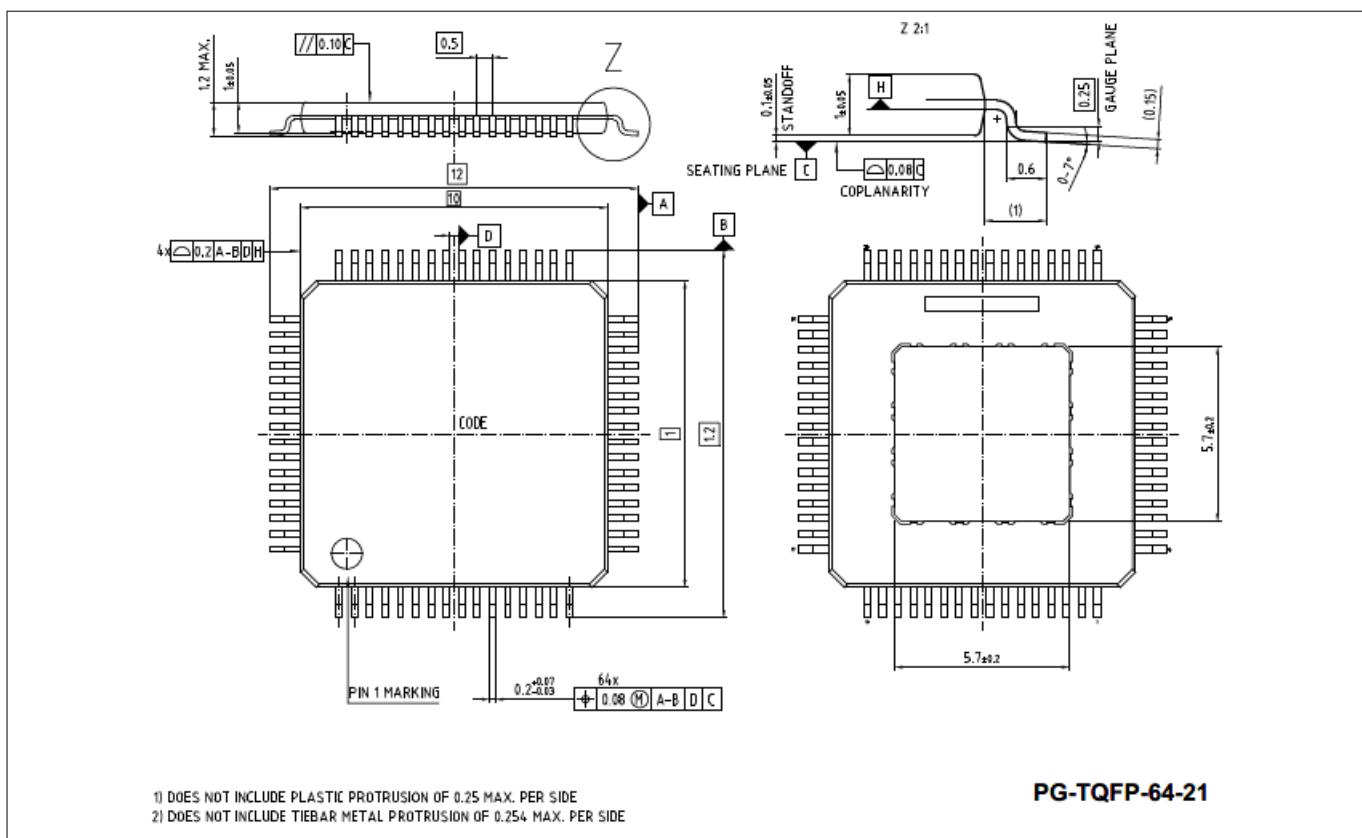


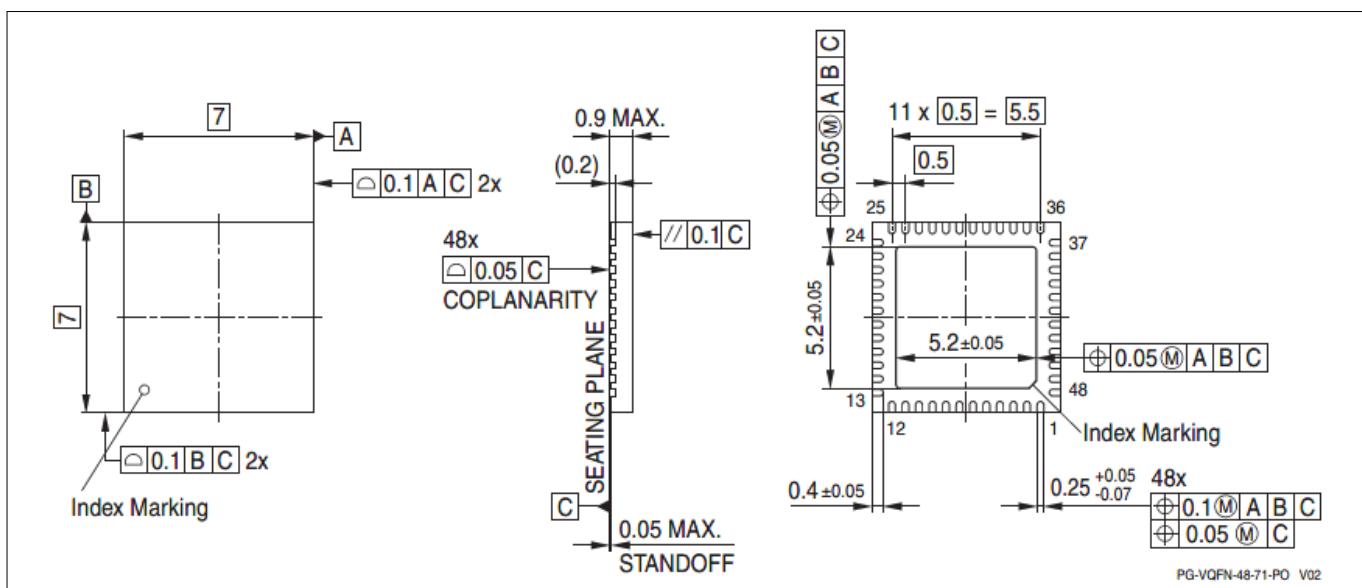
Figure 36

PG-TQFP-64-19 (Plastic Green Thin Profile Quad Flat Package)

#### 4 Package and Reliability



**Figure 37** PG-TQFP-64-21 (Plastic Green Thin Profile Quad Flat Package)



**Figure 38** PG-VQFN-48-71 (Plastic Green Very Thin Profile Flat Non Leaded Package)

All dimensions in mm.

You can find complete information about Infineon packages, packing and marking in our Infineon Internet Page "Packages": <http://www.infineon.com/packages>

## 5 Quality Declarations

### 5 Quality Declarations

The qualification of the XMC4[12]00 is executed according to the JEDEC standard JESD47H.

**Note:** For automotive applications refer to the Infineon automotive microcontrollers.

**Table 56 Quality Parameters**

<b>Parameter</b>	<b>Symbol</b>	<b>Values</b>			<b>Unit</b>	<b>Note/Test Condition</b>
		<b>Min.</b>	<b>Typ.</b>	<b>Max.</b>		
Operation lifetime	$t_{OP}$ CC	20	-	-	a	$T_J \leq 109^{\circ}\text{C}$ , device permanent on
ESD susceptibility according to Human Body Model (HBM)	$V_{HBM}$ SR	-	-	2000	V	EIA/JESD22-A114-B
ESD susceptibility according to Charged Device Model (CDM)	$V_{CDM}$ SR	-	-	500	V	Conforming to JESD22-C101-C
Moisture sensitivity level	MSL CC	-	-	3	-	JEDEC J-STD-020D
Soldering temperature	$T_{SDR}$ SR	-	-	260	°C	Profile according to JEDEC J-STD-020D

**Revision history**

## Revision history

<b>Document revision</b>	<b>Date</b>	<b>Description of changes</b>
V1.5	2023-04-01	<p><a href="#">Table 55</a> Deleted package details: PG-LQFP-64-19 and PG-VQFN-48-53. Added package details: PG-TQFP-64-21.</p> <p><a href="#">Package Outlines</a> Deleted Table 56 and 57.</p> <p><a href="#">Figure 37</a> Added package diagram: PG-TQFP-64-21. Deleted package diagram: PG-LQFP-64-19.</p>
V1.6	2024-11-28	Template update; no content update

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